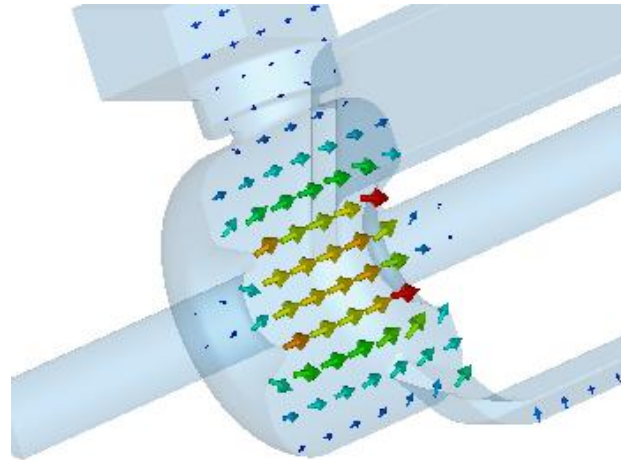
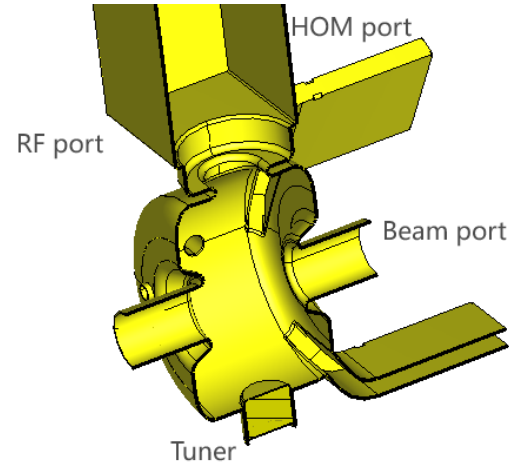
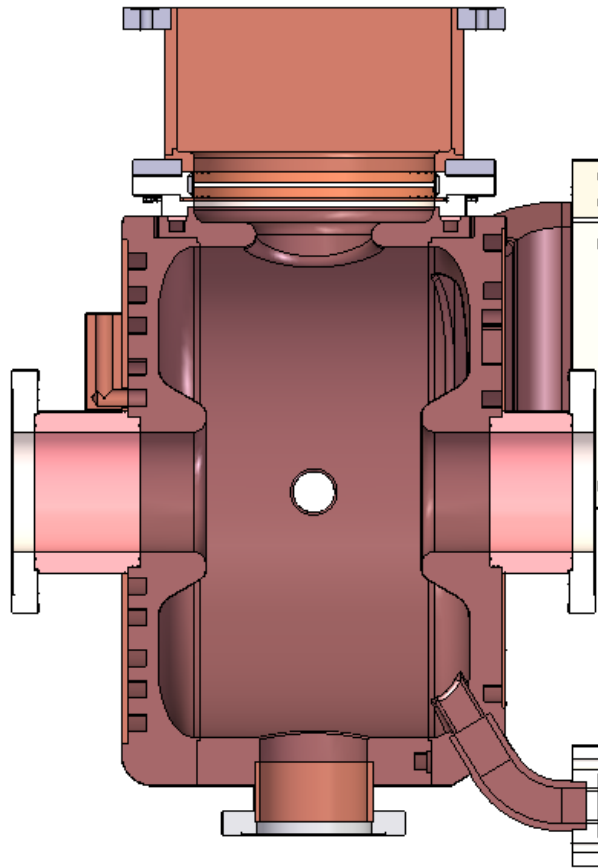


RF Cavity

Wencheng Fang
ISBA, Sept. 2nd, 2025, Shanghai

What's RF Cavity?



Boundary

- ☐ Copper
- ☐ Steel
- ☐ Niobium
- ☐ Others...

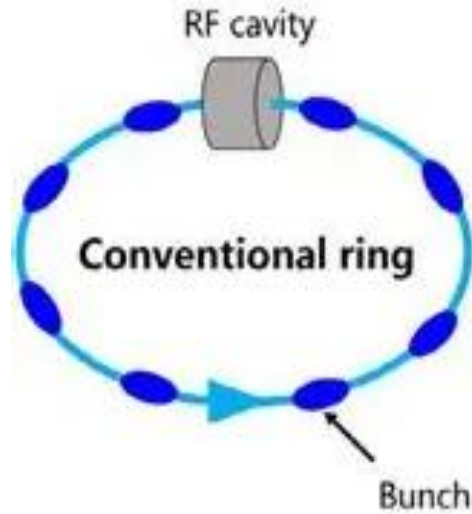
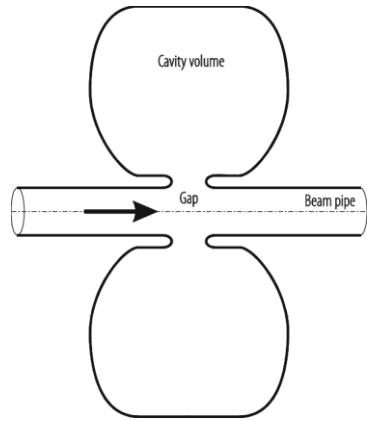
Space

- ☐ Vacuum
- ☐ Air
- ☐ Dielectric
- ☐ Others...

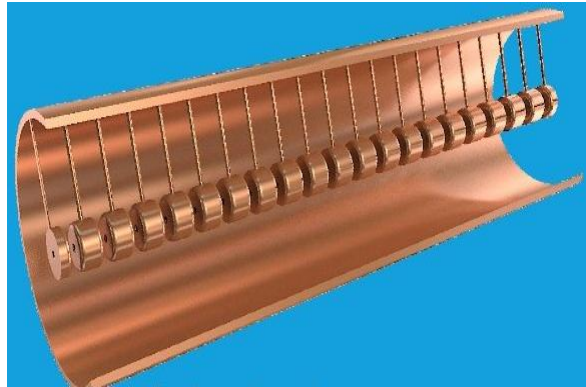
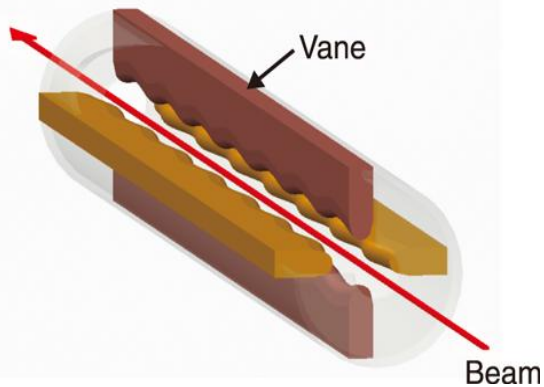
NC Copper boundary + **Vacuum** space in this course!

Typic RF cavities

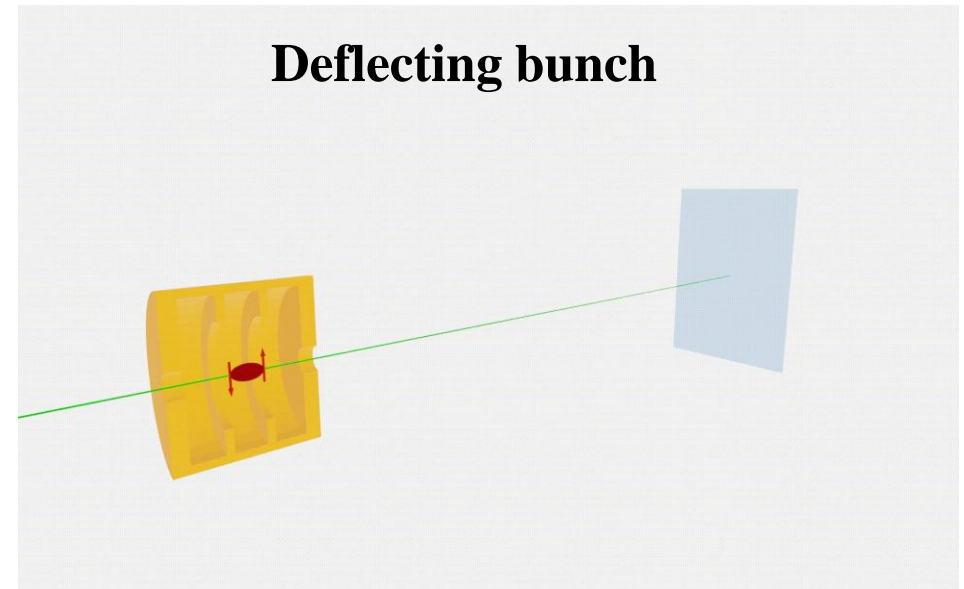
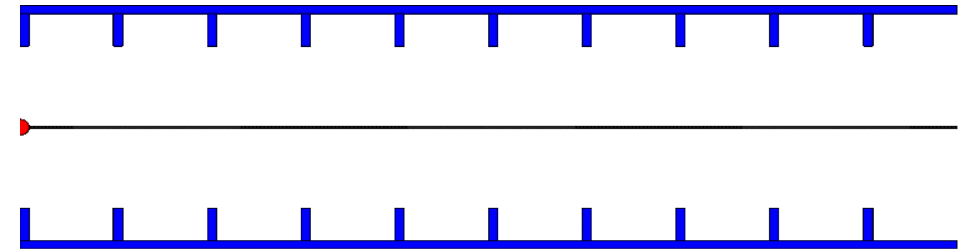
1. Single-cell RF cavity in electron ring



Focusing and acceleration



2. Multi-cell accelerating structure in linac



3. RFQ, DTL cavity in proton and ions injector

4. Multi-cell Deflecting cavity



中国科学院上海高等研究院
SHANGHAI ADVANCED RESEARCH INSTITUTE, CHINESE ACADEMY OF SCIENCES

- **Look back to basic theory**
- **NC accelerating structure**
- **Deflecting cavity**
- **NC RF cavity for light source**

Basic Theory: Maxwell Equations

Where:

J is the current density

E is the electric field intensity

D is the electric flux density

H is the magnetic intensity field

B is the magnetic flux density

p is the charge density

Faraday's Law

$$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t}$$

Ampère's Law

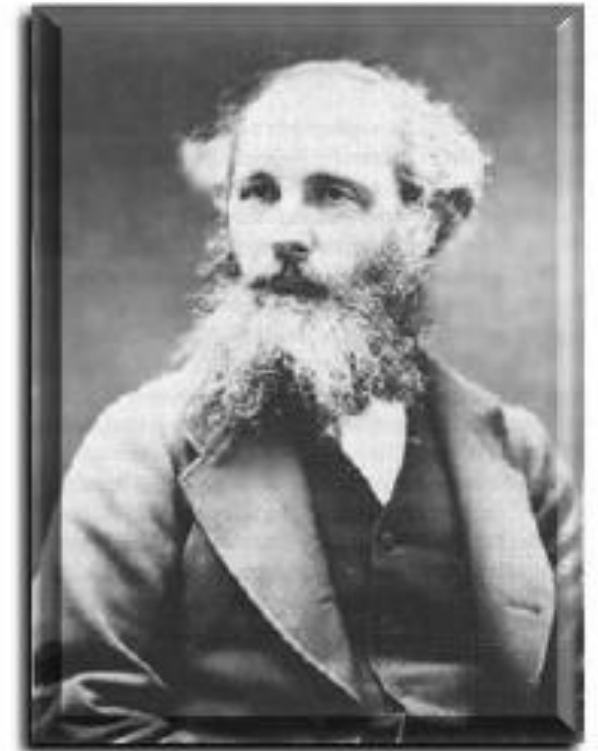
$$\nabla \times \mathbf{H} = \frac{\partial \mathbf{D}}{\partial t} + \mathbf{J}$$

Gauss' Law (Electricity)

$$\nabla \cdot \mathbf{D} = \rho$$

Gauss' Law (Magnetism)

$$\nabla \cdot \mathbf{B} = 0$$



James Clerk Maxwell

Look back to basic theory in RF cavity

$$\nabla \times \mathbf{E}(\mathbf{r}, t) = -\frac{\partial}{\partial t} \mathbf{B}(\mathbf{r}, t) \quad \left| \nabla \times \text{ taking the curl on both sides} \right.$$

$$\nabla \times \nabla \times \mathbf{E}(\mathbf{r}, t) = \nabla \times \left(-\frac{\partial}{\partial t} \mathbf{B}(\mathbf{r}, t) \right) \quad \left| \text{ exchanging curl and time derivative} \right.$$

$$\nabla \times \nabla \times \mathbf{E}(\mathbf{r}, t) = -\frac{\partial}{\partial t} (\nabla \times \mathbf{B}(\mathbf{r}, t)) \quad \left| \text{ applying the material law } \mathbf{B} = \mu \mathbf{H} \right.$$

$$\nabla \times \nabla \times \mathbf{E}(\mathbf{r}, t) = -\frac{\partial}{\partial t} (\nabla \times \mu \mathbf{H}(\mathbf{r}, t)) \quad \left| \text{ constant permeability in vacuum} \right.$$

$$\nabla \times \nabla \times \mathbf{E}(\mathbf{r}, t) = -\mu \frac{\partial}{\partial t} \nabla \times \mathbf{H}(\mathbf{r}, t) \quad \left| \text{ using Ampère's law} \right.$$

$$\nabla \times \nabla \times \mathbf{E}(\mathbf{r}, t) = -\mu \frac{\partial}{\partial t} \left(\frac{\partial}{\partial t} \mathbf{D}(\mathbf{r}, t) + \mathbf{J}(\mathbf{r}, t) \right) \quad \left| \text{ deriving expression in brackets} \right.$$

$$= -\mu \frac{\partial^2}{\partial t^2} \mathbf{D}(\mathbf{r}, t) - \mu \frac{\partial}{\partial t} \mathbf{J}(\mathbf{r}, t)$$

Wave equation

$$\nabla \times \nabla \times \mathbf{E}(\mathbf{r}, t) = -\mu \frac{\partial^2}{\partial t^2} \mathbf{D}(\mathbf{r}, t) - \mu \frac{\partial}{\partial t} \mathbf{J}(\mathbf{r}, t) \quad \left| \quad \text{applying the material law } \mathbf{D} = \epsilon \mathbf{E} \right.$$

$$= -\epsilon \mu \frac{\partial^2}{\partial t^2} \mathbf{E}(\mathbf{r}, t) - \mu \frac{\partial}{\partial t} \mathbf{J}(\mathbf{r}, t)$$

$$\nabla \times \nabla \times \mathbf{E}(\mathbf{r}, t) + \epsilon \mu \frac{\partial^2}{\partial t^2} \mathbf{E}(\mathbf{r}, t) = -\mu \frac{\partial}{\partial t} \mathbf{J}(\mathbf{r}, t) \quad \text{curl-curl equation}$$

$$\nabla \left(\underbrace{\nabla \cdot \mathbf{E}(\mathbf{r}, t)}_{\frac{\rho(\mathbf{r}, t)}{\epsilon}} \right) - \nabla^2 \mathbf{E}(\mathbf{r}, t) + \epsilon \mu \frac{\partial^2}{\partial t^2} \mathbf{E}(\mathbf{r}, t) = -\mu \frac{\partial}{\partial t} \mathbf{J}(\mathbf{r}, t) \quad \left| \quad \text{for charge-free in the vacuum} \right.$$

$$\nabla^2 \mathbf{E}(\mathbf{r}, t) - \epsilon \mu \frac{\partial^2}{\partial t^2} \mathbf{E}(\mathbf{r}, t) = \mu \frac{\partial}{\partial t} \mathbf{J}(\mathbf{r}, t) \quad \text{wave equation (with excitation)}$$

Wave equation

Wave equation with excitation

$$\nabla^2 \mathbf{E}(\mathbf{r}, t) - \varepsilon\mu \frac{\partial^2}{\partial t^2} \mathbf{E}(\mathbf{r}, t) = \mu \frac{\partial}{\partial t} \mathbf{J}(\mathbf{r}, t)$$

Reminding: **vacuum in the RF cavity**, far from the sources, excitation vanishes

$$\nabla^2 \mathbf{E}(\mathbf{r}, t) - \varepsilon\mu \frac{\partial^2}{\partial t^2} \mathbf{E}(\mathbf{r}, t) = \mathbf{0} \quad \text{with speed of light} \quad c = \frac{1}{\sqrt{\varepsilon\mu}}$$

RF field works on the constant frequency ω in the RF cavity, then wave equations is:

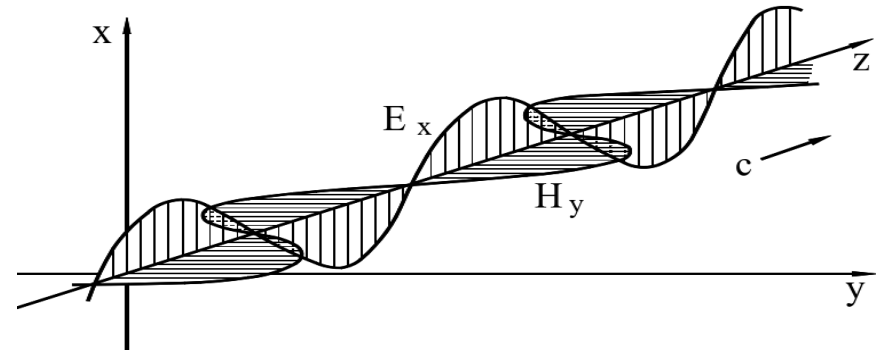
$$\nabla^2 \vec{E} + k^2 \vec{E} = \mathbf{0}$$

$$\nabla^2 \vec{H} + k^2 \vec{H} = \mathbf{0}$$

- Take-home
- Derivation

Wave number
Wave vector

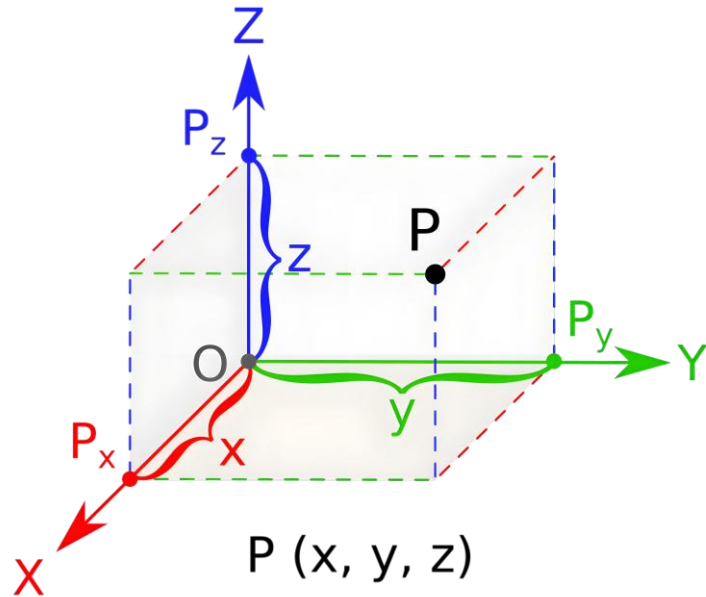
$$k^2 = \omega^2 \varepsilon\mu$$



Plane Wave propagating in +z-Direction in free space

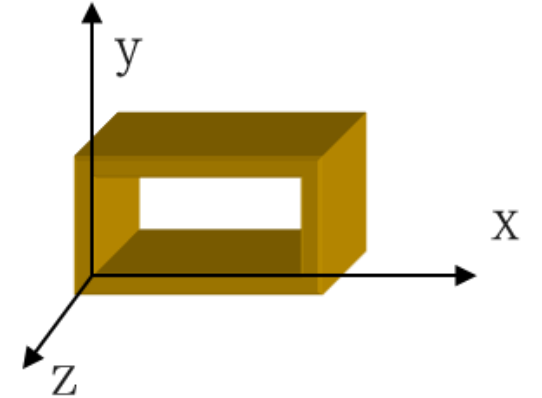
RF field in rectangular waveguide

Cartesian coordinate system



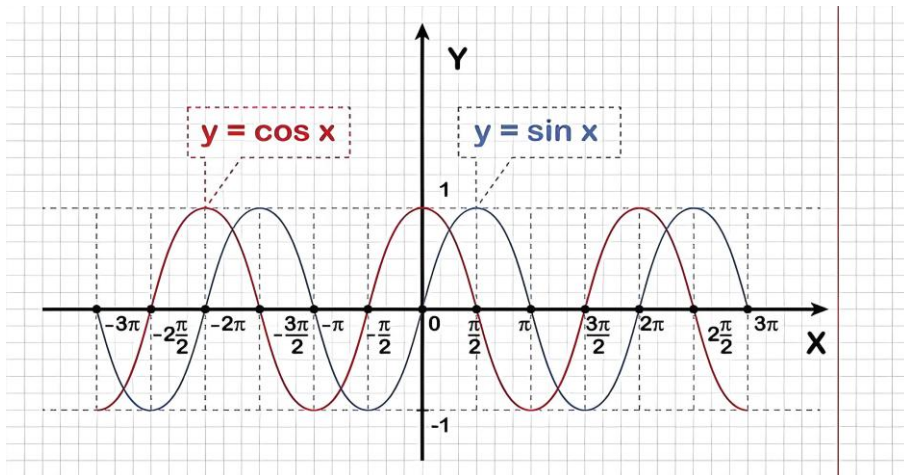
$$E(x, y, z),$$

$$H(x, y, z)$$



TE_{mnp} mode

TM_{mnp} mode



$$\underline{E}_x = -\underline{C}^H \frac{n\pi}{b} \cos\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) e^{\pm ik_z z},$$

$$\underline{E}_y = \underline{C}^H \frac{m\pi}{a} \sin\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) e^{\pm ik_z z},$$

$$\underline{E}_z = 0,$$

$$\underline{H}_x = \pm \frac{k_z}{\omega\mu} \underline{E}_y,$$

$$\underline{H}_y = \mp \frac{k_z}{\omega\mu} \underline{E}_x,$$

$$\underline{H}_z = -\underline{C}^H \frac{k^2 - k_z^2}{i\omega\mu} \cos\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) e^{\pm ik_z z}.$$

$$\underline{H}_x = \underline{C}^E \frac{n\pi}{b} \sin\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) e^{\pm ik_z z},$$

$$\underline{H}_y = -\underline{C}^E \frac{m\pi}{a} \cos\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) e^{\pm ik_z z},$$

$$\underline{H}_z = 0,$$

$$\underline{E}_x = \mp \frac{k_z}{\omega\varepsilon} \underline{H}_y,$$

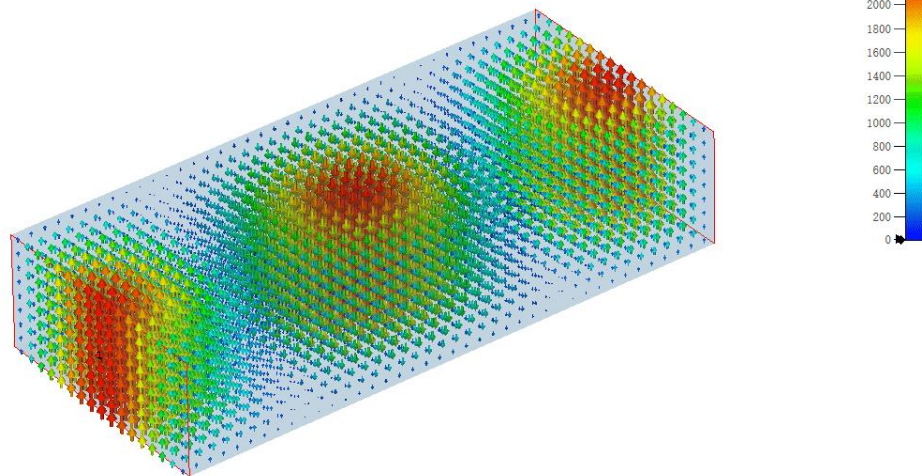
$$\underline{E}_y = \pm \frac{k_z}{\omega\varepsilon} \underline{H}_x,$$

$$\underline{E}_z = \underline{C}^E \frac{k^2 - k_z^2}{i\omega\varepsilon} \sin\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) e^{\pm ik_z z}.$$

Cosine and Sine in all directions: X, Y, Z.

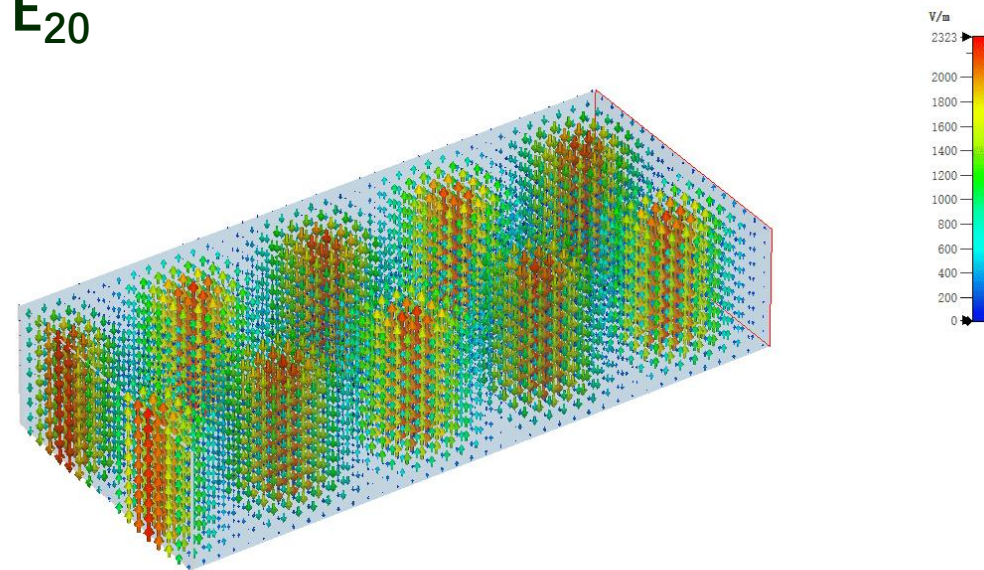
RF field in rectangular waveguides

● TE_{10}



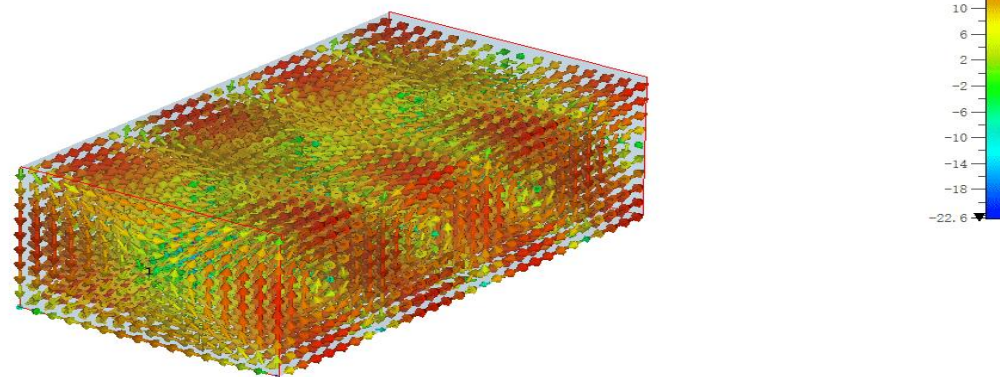
e-field (f=9200) [1]
Frequency 9200 MHz
Phase 0°
Maximum (Plot) 2278.38 V/m

● TE_{20}



e-field (f=1.91e+4) [1(2)]
Frequency 19100 MHz
Phase 0°
Maximum (Plot) 2323.13 V/m

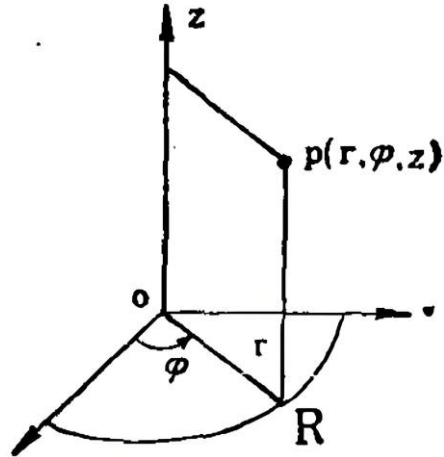
● TE_{11}



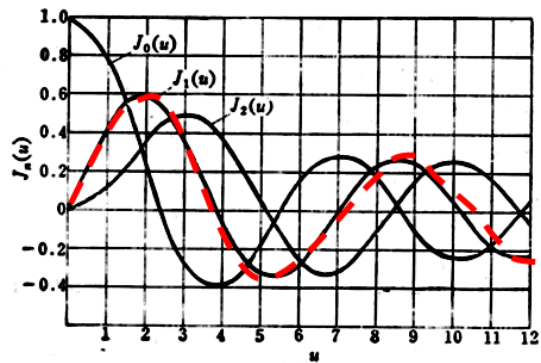
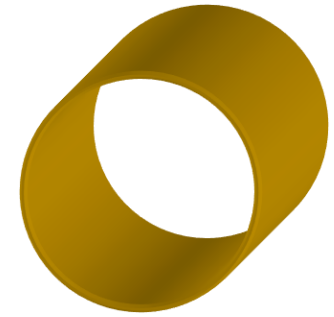
h-field (f=2.5e+4) [1(3)]
Frequency 25000 MHz
Phase 0°
Maximum (Plot) 17.3689 dB(A/m)

RF field in Circular waveguides

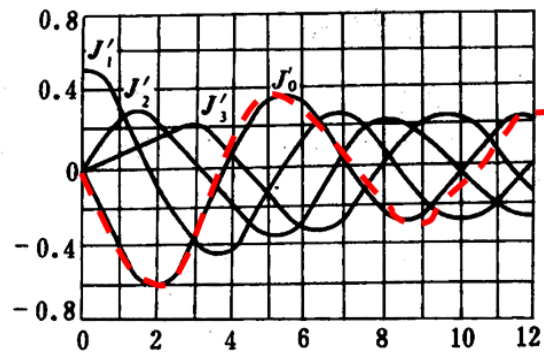
Cylindrical coordinate system



$$E(r, \varphi, z), \\ H(r, \varphi, z)$$



Bessel Functions $J_n(k_c r)$



Derivative functions $J'_n(k_c r)$

TE_{mnp} mode

$$\begin{cases} E_z = 0 \\ E_r = \frac{jn\omega_0\mu_0}{k_c^2 r} H_m J_n(k_c r) \sin[n(\phi - \phi_0)] e^{-jk_z z} \\ E_\phi = \frac{jn\omega_0\mu_0}{k_c} H_m J'_n(k_c r) \cos[n(\phi - \phi_0)] e^{-jk_z z} \\ H_z = H_m J_n(k_c r) \cos[n(\phi - \phi_0)] e^{-jk_z z} \\ H_r = -\frac{jk_z}{k_c} H_m J'_n(k_c r) \cos[n(\phi - \phi_0)] e^{-jk_z z} \\ H_\phi = \frac{jnk_z}{k_c^2 r} H_m J_n(k_c r) \sin[n(\phi - \phi_0)] e^{-jk_z z} \end{cases}$$

TM_{mnp} mode

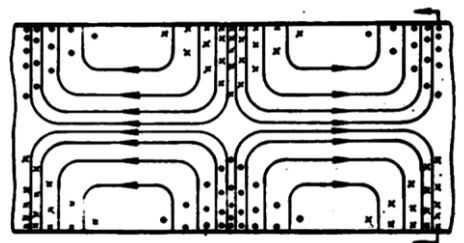
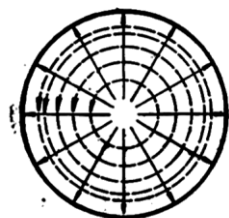
$$\begin{cases} E_z = E_m J_n(k_c r) \cos[n(\phi - \phi_0)] e^{-jk_z z} \\ E_r = -\frac{jk_z}{k_c} E_m J'_n(k_c r) \cos[n(\phi - \phi_0)] e^{-jk_z z} \\ E_\phi = \frac{jnk_z}{k_c^2 r} E_m J_n(k_c r) \sin[n(\phi - \phi_0)] e^{-jk_z z} \\ H_z = 0 \\ H_r = -\frac{jn\omega_0\epsilon_0}{k_c^2 r} E_m J_n(k_c r) \sin[n(\phi - \phi_0)] e^{-jk_z z} \\ H_\phi = -\frac{jn\omega_0\epsilon_0}{k_c} E_m J'_n(k_c r) \cos[n(\phi - \phi_0)] e^{-jk_z z} \end{cases}$$

- ❑ Cosine and Sine in φ, z directions
- ❑ However Bessel in r directions

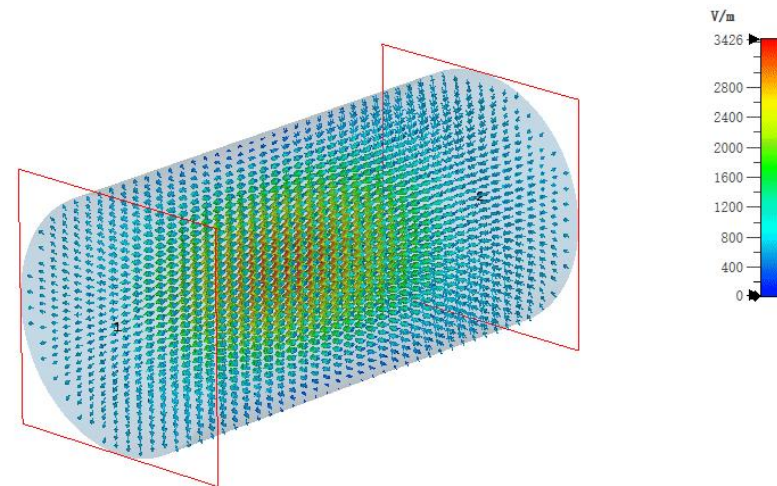


RF field in Circular waveguides

● TM_{01}

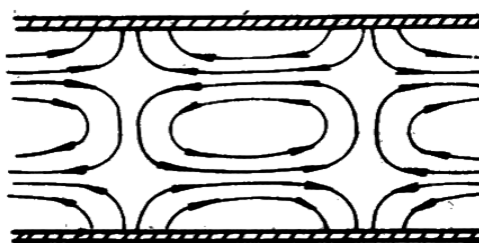
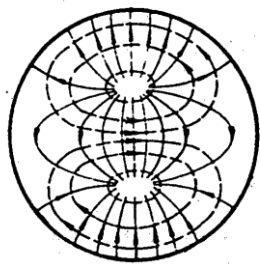


Acceleration Mode

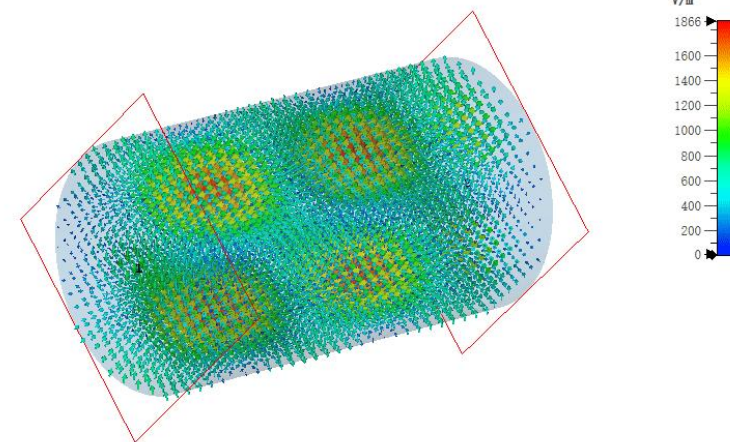


e-field (f=1.2e+4) [1(3)]
Frequency 12000 MHz
Phase 0°
Maximum (Plot) 3425.93 V/m

● HEM_{11}



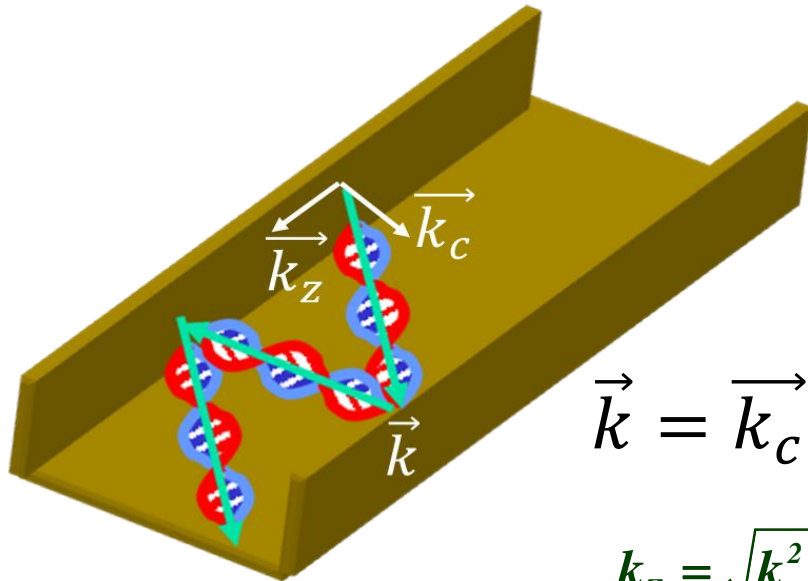
Deflecting Mode



e-field (f=1.4e+4) [1(7)]
Frequency 14000 MHz
Phase 0°
Maximum (Plot) 1865.61 V/m

Cut-off frequency and dispersion

Wave is reflected many times during propagation in the waveguide.



$$\vec{k} = \vec{k}_c + \vec{k}_z$$

$$k_z = \sqrt{k^2 - k_c^2}$$

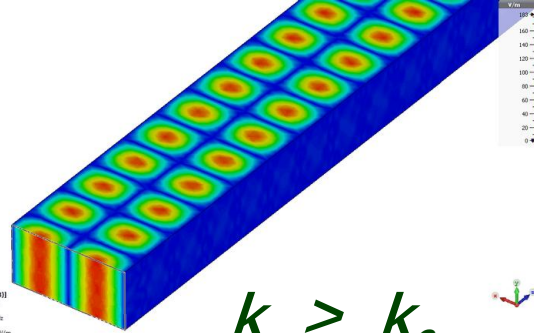
$$\nabla^2 \vec{E} + k^2 \vec{E} = 0$$

$$\nabla^2 \vec{H} + k^2 \vec{H} = 0 \quad \rightarrow \quad \begin{aligned} E_z &= E_z(x, y) e^{\mp j k_z z} \\ H_z &= H_z(x, y) e^{\mp j k_z z} \end{aligned}$$

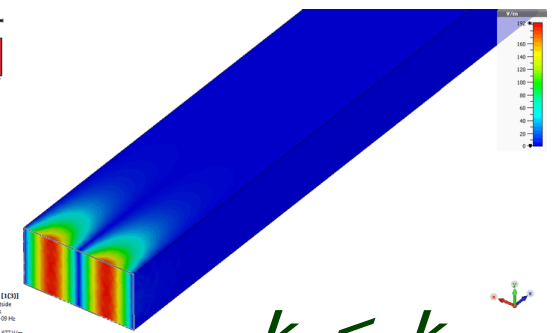
$$\nabla_T^2 E_z(x, y) + k_c^2 E_z(x, y) = 0$$

$$\nabla_T^2 H_z(x, y) + k_c^2 H_z(x, y) = 0$$

$$k_c^2 = k^2 - k_z^2$$



$$k > k_c$$



$$k < k_c$$

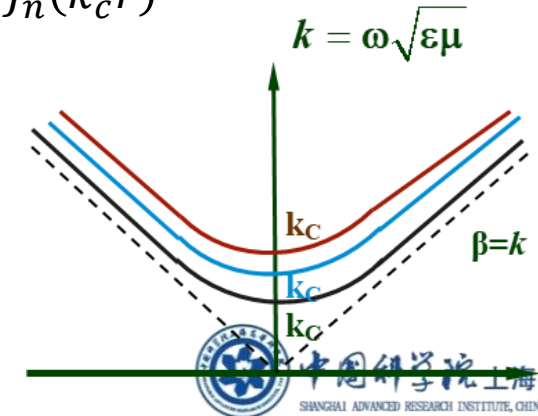
$k_c = 2\pi f_c$ is cut-off frequency

Rect.

$$k_{c(mn)} = \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}$$

Circ.

$$k_c \text{ in } J_n(k_c r)$$



Some types of RF cavity

- **Accelerating structures**
- Deflecting RF cavity
- NC Single-cell RF cavity

Accelerating structures for linac

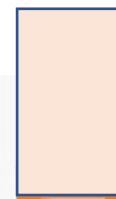
$$E_p = q \cdot E_z \cos(\omega t - \beta z) \cdot L$$



E_p : Energy gain in the structure
 q : Charge of particle
 E_z : Accelerating gradient
 ω : Frequency of RF structure
 β : v_p/c , phase velocity of RF field.
 L : Length of accelerating structure

- Synchronization acceleration
- Cascaded for linac

Input coupler



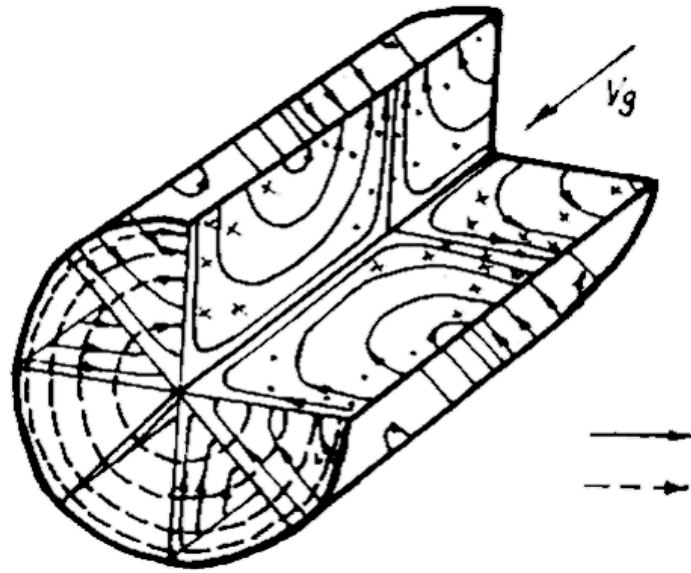
- Disk-loaded accelerating structure
- Travelling wave for most linac

Output coupler



Linac

Why disk-loaded? Not constant circular waveguide?



TM₀₁

Periodic structure, Floquet theorem



SSRF

$$E_z = E_0 J_0(k_c r) \cdot \cos(\omega t - \beta z) \quad \beta = \frac{\omega}{V_p}$$

$$k_c^2 = \left(\frac{\omega}{c}\right)^2 - \left(\frac{\omega}{V_p}\right)^2 > 0 \quad V_p > c$$

- Take-home
- In physics?

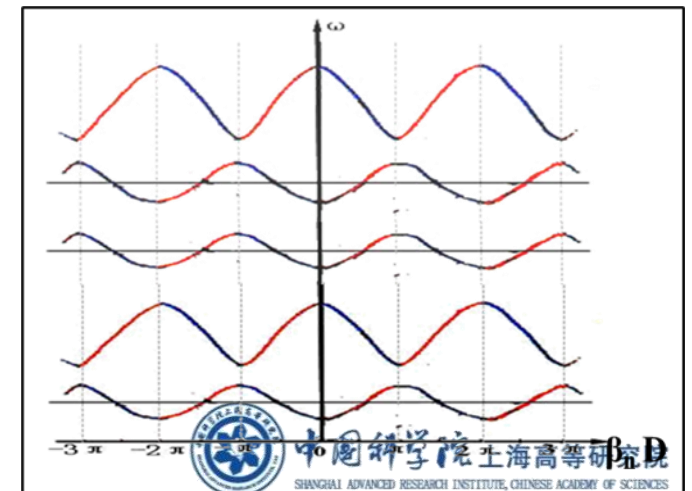
In order to propagate RF field and synchronize particle and RF phase together, constant circular waveguide is not able to be used for acceleration directly, and disk-loaded structure can slow the phase velocity to match the light speed c .

$$E_z = j \sum_{n=-\infty}^{\infty} E_n J_0(\chi_n r) e^{j(\omega t - \beta_n z)}$$

$$E_r = - \sum_{n=-\infty}^{\infty} \frac{E_n \beta_n J_1(\chi_n r)}{\chi_n} e^{j(\omega t - \beta_n z)}$$

$$H_\theta = -\omega \epsilon_0 \sum_{n=-\infty}^{\infty} \frac{E_n J_1(\chi_n r)}{\chi_n} e^{j(\omega t - \beta_n z)}$$

$$E_\theta = H_r = H_z = 0$$



High gradient is key issue on R&D

❑ 1928 Fowler-Nordheim $I_F = \frac{1.54 \times 10^{-6} \times 10^{4.52\varphi^{-0.5}} A_e \beta^2 E^2}{\varphi} \exp\left(-\frac{6.53 \times 10^9 \varphi^{1.5}}{\beta E}\right)$

❑ 1957 W. D. KILPATRICK $WE^2 e^{-K_1/E} = K_2$ **DC, Low frequency**

❑ 1983 W. Peter Modified Kilpatrick formula $f = (62/E) e^{17E}$

❑ 1989 Wang, Juwen Field-emission $I_{FE} = \frac{5.7 \times 10^{-12} \times 10^{4.52\varphi^{-0.5}} A_e (\beta E)^{2.5}}{\varphi^{1.75}} e^{-K\varphi^{1.5}/\beta E}$

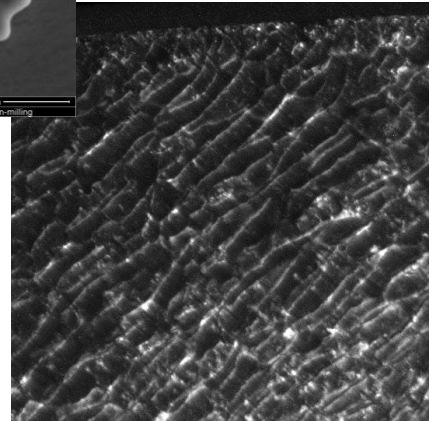
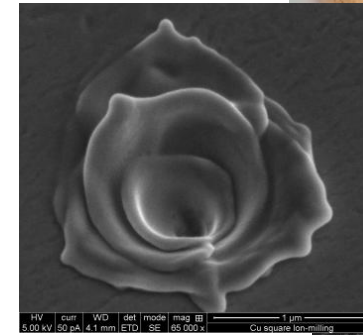
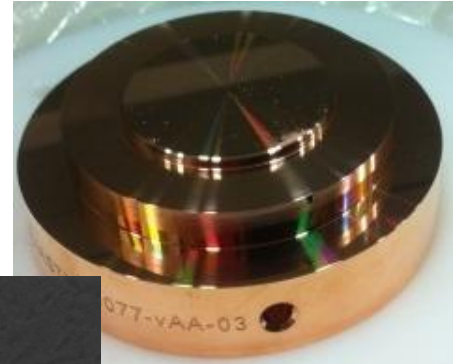
❑ 2009 A. Grudiev Modified Poynting vector $S_c = Re\{\bar{S}\} + g_c \cdot Im\{\bar{S}\}$

❑ 2011 Z. Insepov Surface damage $MTBF = BJ^n e^{-E_a/kT}$ **RF, Microwave
X/C/S-band**

❑ 2012 K. Nordlund Dislocation enthalpy $R_{BD} = a' c_0 E_i^{-E_i/k_B T} E_0^{E_0 \Delta V/k_B T}$
 $= a e^{\varepsilon_0 \Delta V/k_B T}$

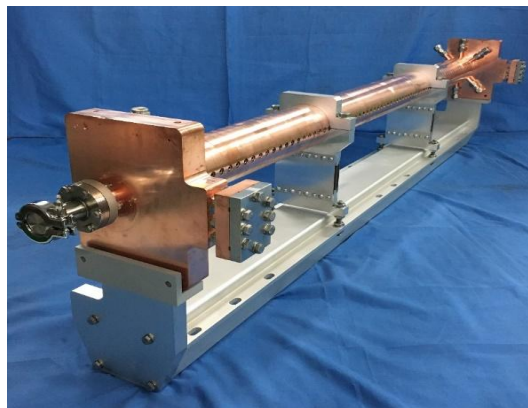
❑ 2018 Eliyahu Zvi Engelberg Fluctuation of moving dislocation $\tau \sim \exp\left\{\gamma \left(1 - \frac{E}{E_0}\right)\right\}$

❑ 2022 Zhou Liuyuan Modified dislocation fluctuation $R_{BD} \approx C \exp\left\{\gamma \left(1 - \sqrt{\frac{\sigma_L}{\sigma_0}}\right)\right\} \sim \exp(\xi \sigma_L)$



Typical accelerating structures for linac

	S-band	C-band	X-band
Structure	SLAC/PAL-FEL	SPring-8/SXFEL	NLC/GLC
RF Frequency	2856 MHz	5712 MHz	11424 MHz
Structure Length	3 m	1.8 m	0.6 m
Filling time	830 ns	286 ns	105 ns
Shunt impedance	53 ~ 60 MΩ/m	53.1 MΩ/m	48.8 ~ 77.8 MΩ/m
Operational Gradient	20 MV/m	40 MV/m	65-80 MV/m



How to design the accelerating structure

- Aperture range a/λ – wakefields
- Optimization for high RF efficiency
- Attenuation factor
- Total length
- Group velocity range
- High field suppression
- Beam loading calculation

Longitudinal $SWR : W(s) = \frac{Z_0 c}{\pi a^2} e^{-1.16(s/mm)^{0.55}}$

Transverse $SWR : W(s) = \frac{2Z_0 c}{\pi a^4} (1 - e^{-0.89(s/mm)^{0.87}})$

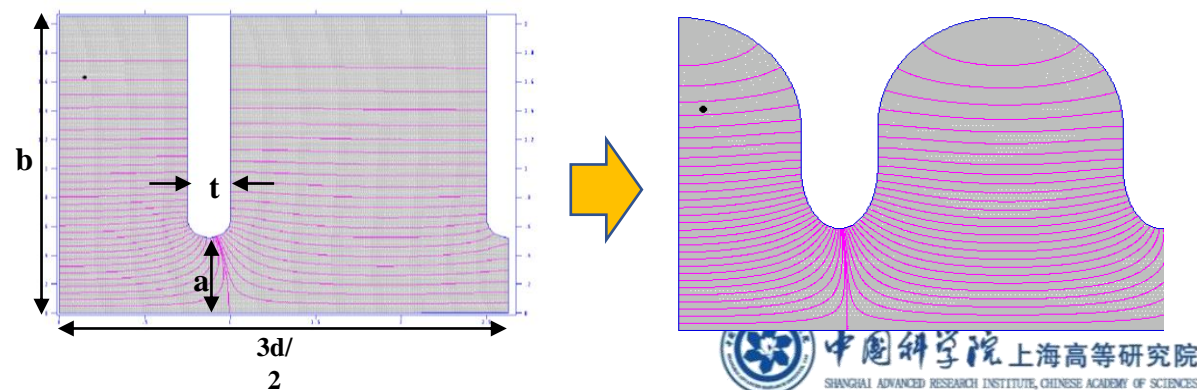
Shunt impedance $R_s = \frac{E^2}{P_l}$

Group velocity $v_g = \frac{\omega L}{Q} \{ [1 - (1 - e^{-2\tau})z / L] / (1 - e^{-2\tau}) \}$

Peak field $S_c = \text{Re}\{\bar{S}\} + g_c \cdot \text{Im}\{\bar{S}\}$, and more...

For FEL and light source:

- Good beam quality: large aperture a/λ
- Higher impedance: high mode, $3\pi/4$, $4\pi/5$, $5\pi/6$
- Lower field distribution: round cell, elliptical iris



Coupler design

1. Coupler Problem and Resolutions

- Pulse Heating Reduction
 - Fat lips
 - Waveguide couplers
- Symmetry feed for correction of dipole modes
- Racetrack cavity for correction of quadruple modes

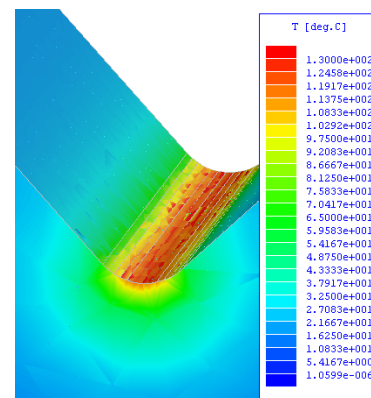
2. Some Coupler Types

- Iris coupling
- Mode convertor coupling
- Compact coupler

3. Coupler simulation

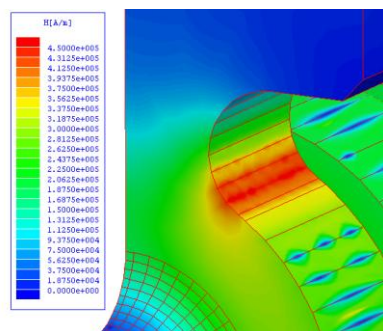
CST
HFSS

Multi-physics simulation

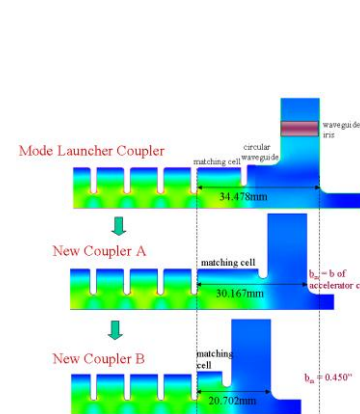


$$\Delta T = H_t^2 \sqrt{T_p} \frac{R_s}{\sqrt{\pi \rho c \kappa}}$$

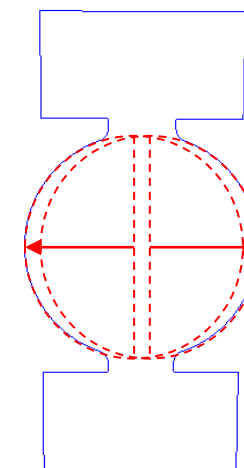
Magnetic field $\rightarrow H_t$
 Pulse width $\rightarrow T_p$
 Surface resistivity $\rightarrow R_s$
 Thermal conductivity $\rightarrow \kappa$
 Specific heat $\rightarrow c$



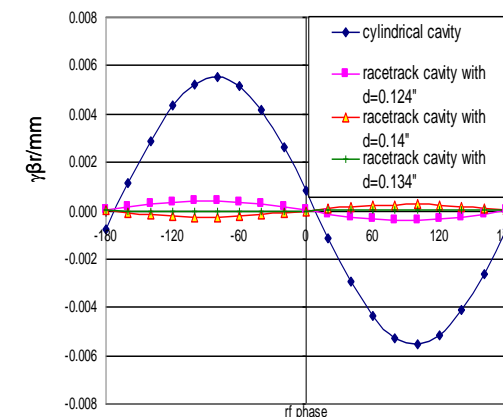
Iris coupling
High pulse heating
More symmetric
Low energy (S-band)



Mode launcher
low pulse heating
asymmetric
High energy (C/X-band)



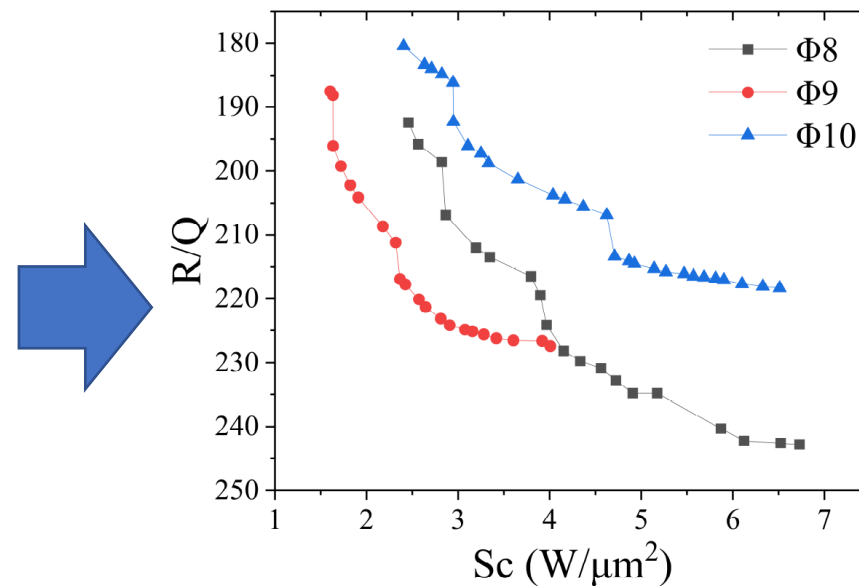
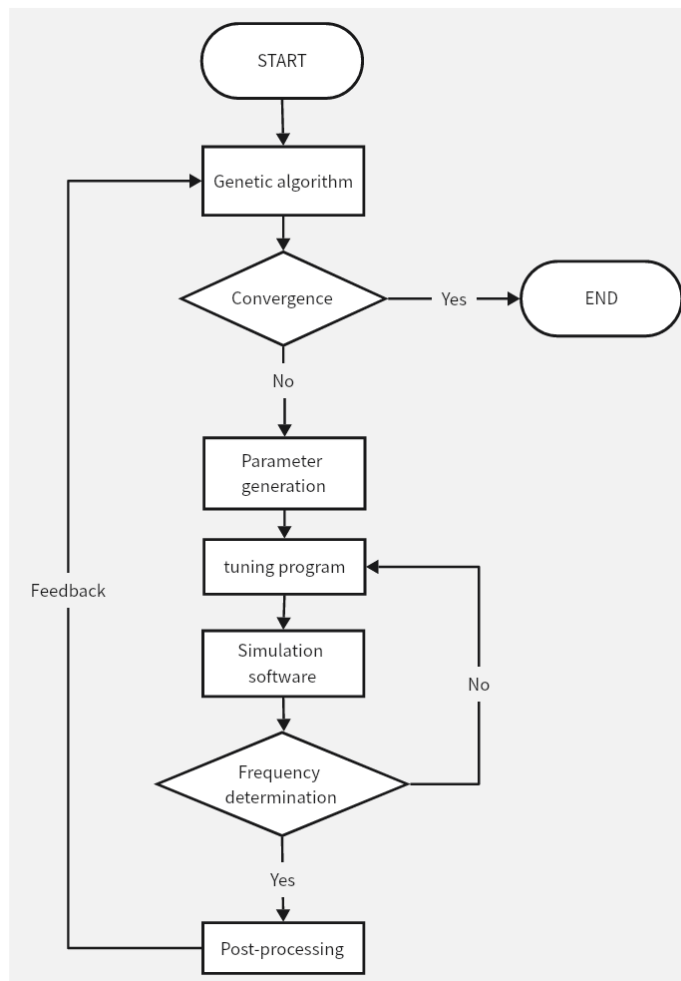
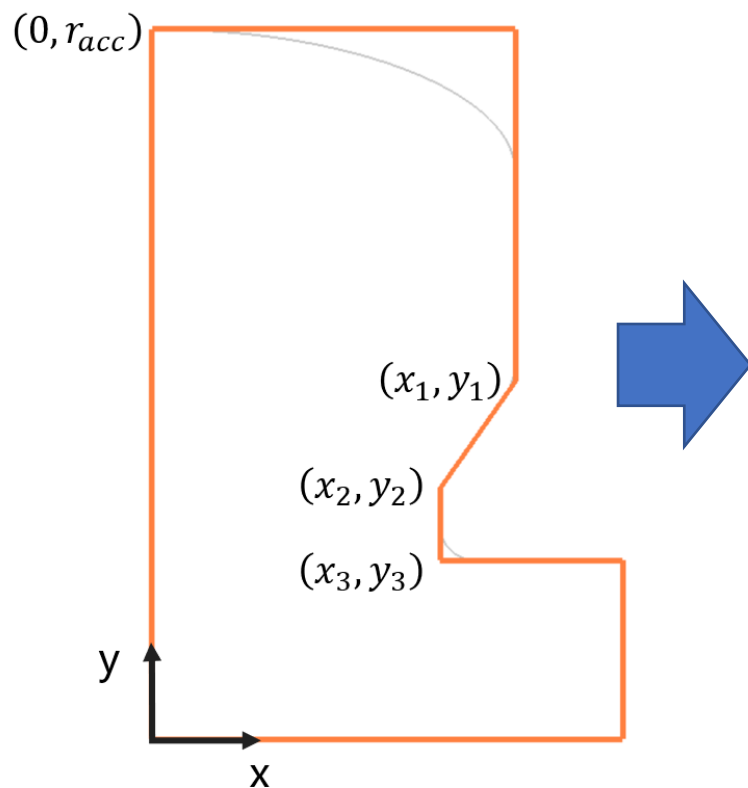
Racetrack coupler



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Genetic algorithm used for optimization

- Many size and angle parameters as input, and many targets should be optimized as well
- Genetic algorithm is used to improve the efficiency of high gradient AC optimization

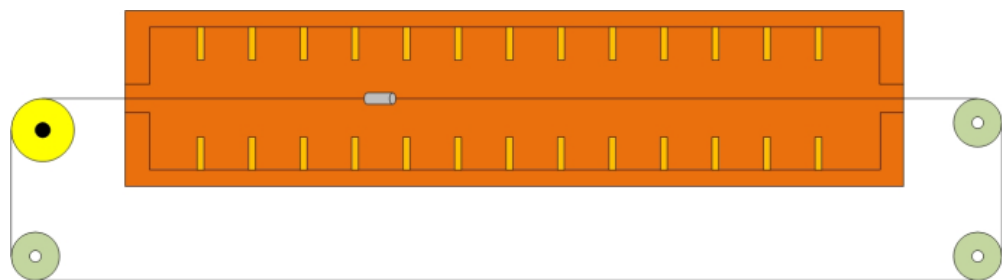
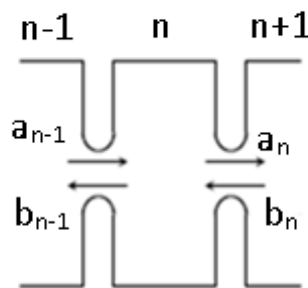


Bead-pulled RF measurement and tuning

$$2P_i(\Gamma_p - \Gamma_a) = -j\omega[k_e E_a^2 - k_m H_a^2]$$

$$\Delta S_{11} = S_{11_p} - S_{11_a} = -\frac{j\omega k_e}{P_i} E_z^2$$

$$S_n e^{j\theta_n} = \frac{b_n e^{j\phi_n} - b_{n+1} e^{j(\phi_{n+1} - 2\pi/3)}}{a_n e^{j\psi_n}} = A + jB$$

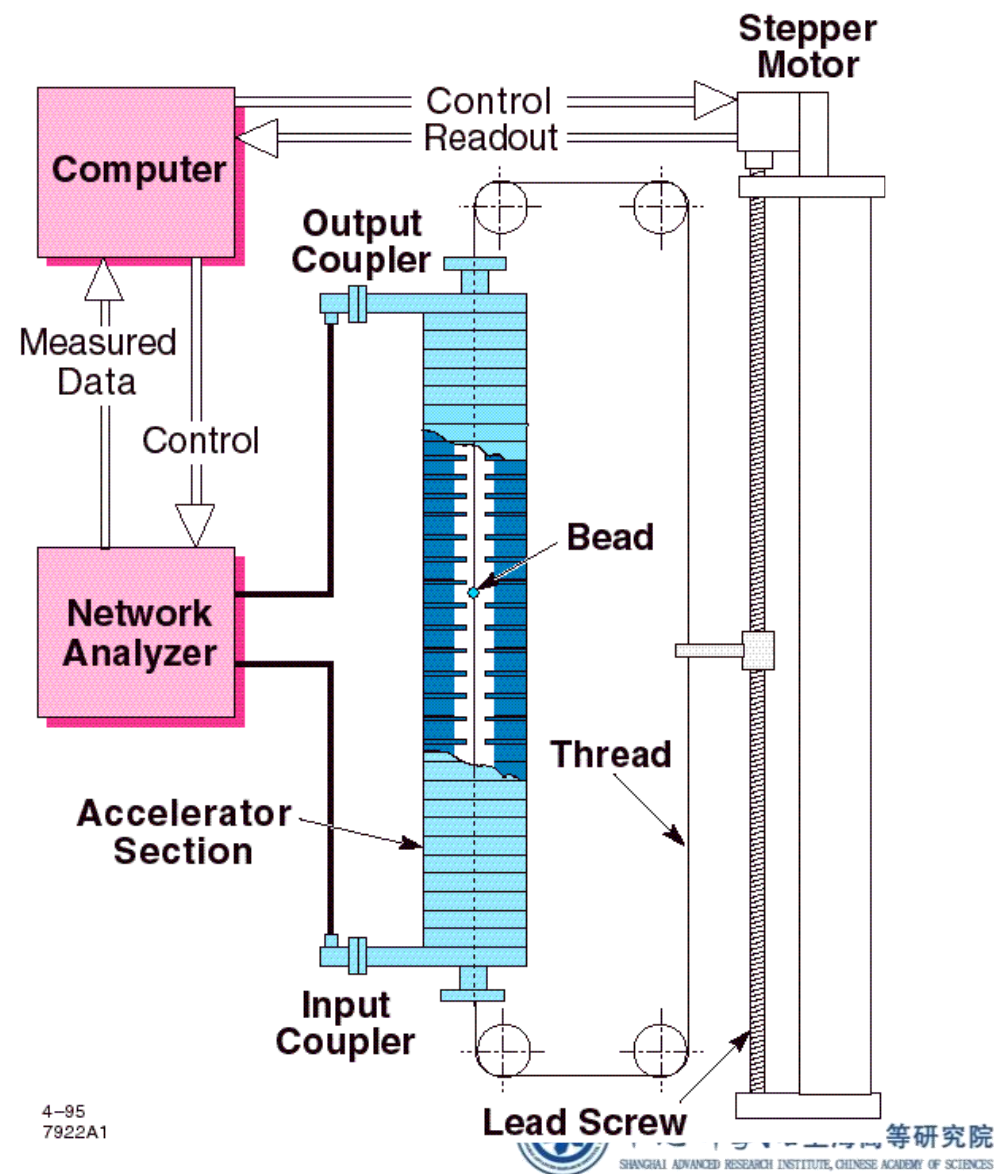


步进电机



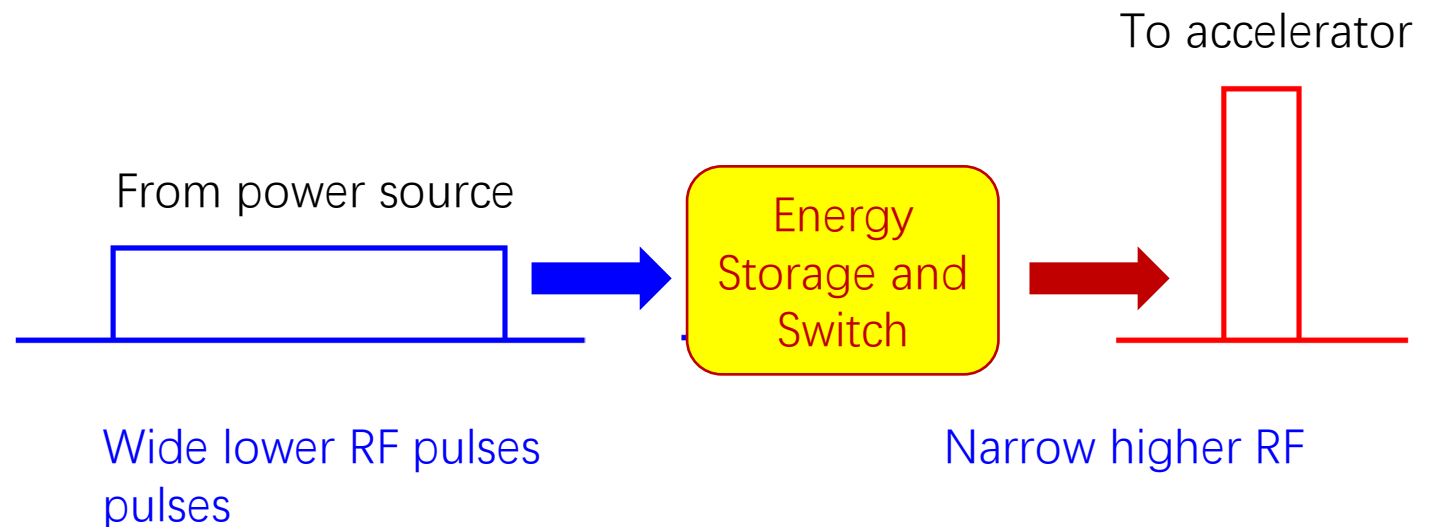
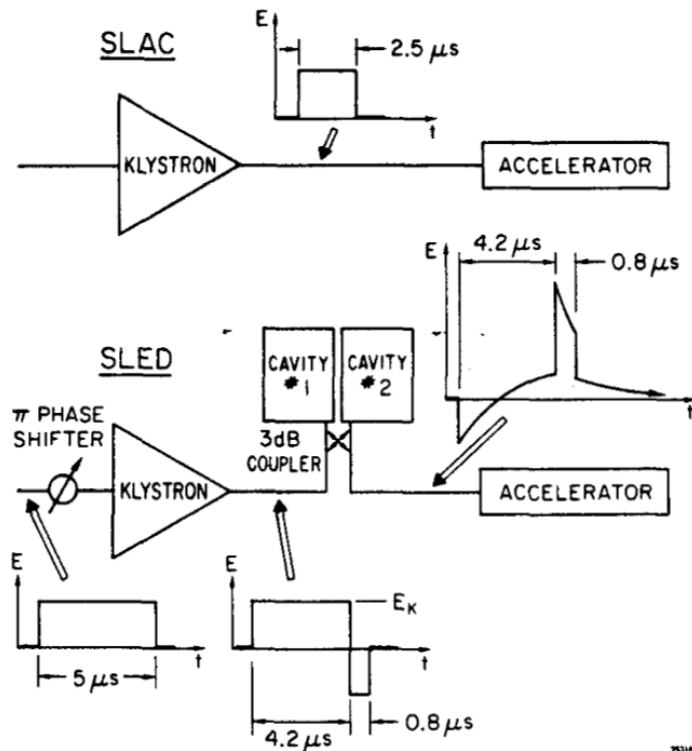
滑轮

微扰体



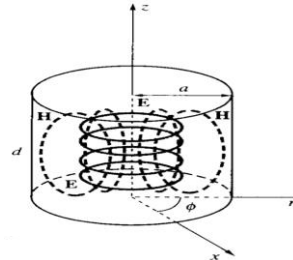
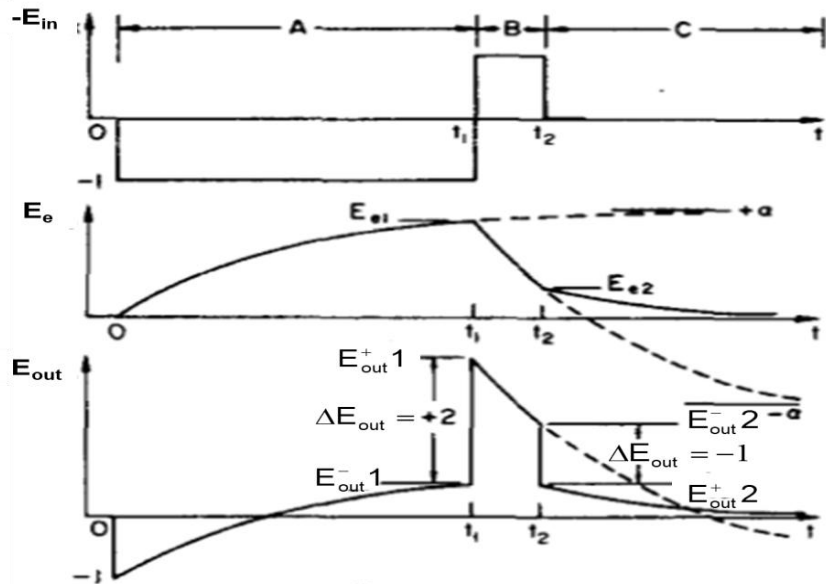
Pulse Compressor to improve high gradient

- Extremely high RF power is needed for high gradient accelerators
- High peak RF power limitation for pulsed sources

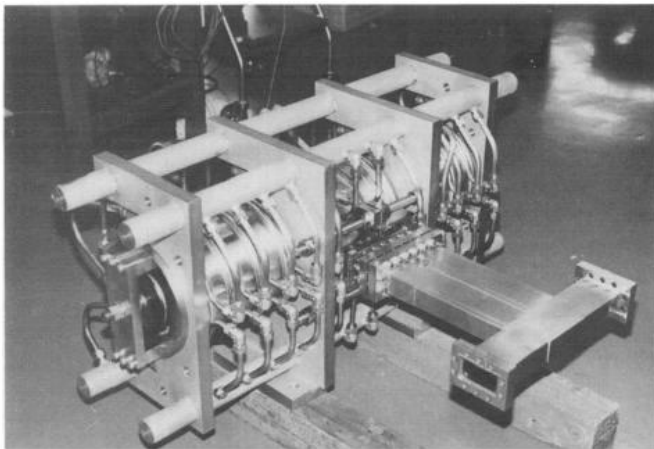
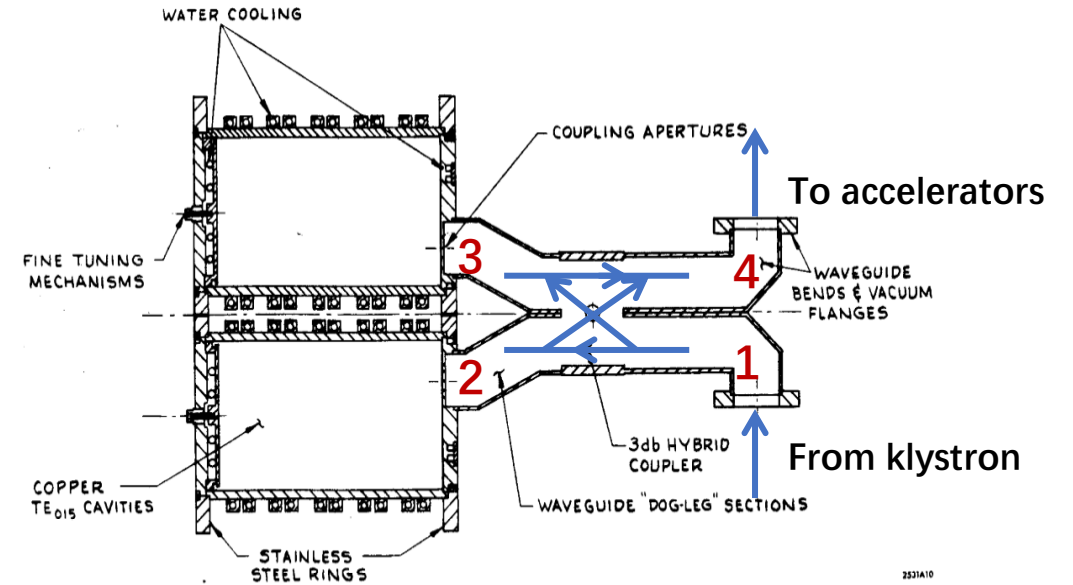
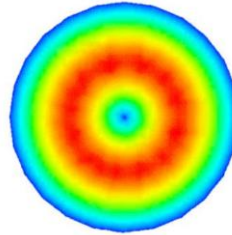


SLED

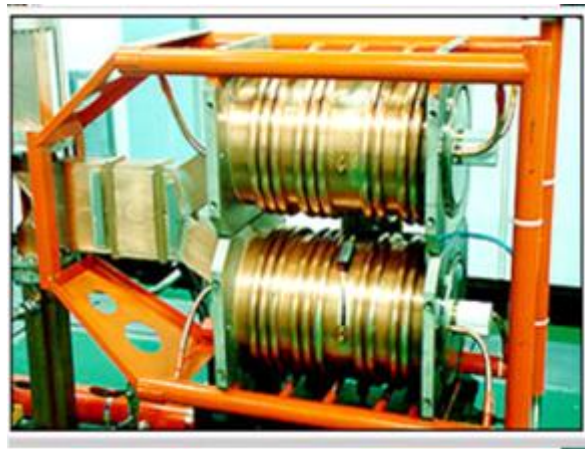
Waveforms for SLED System



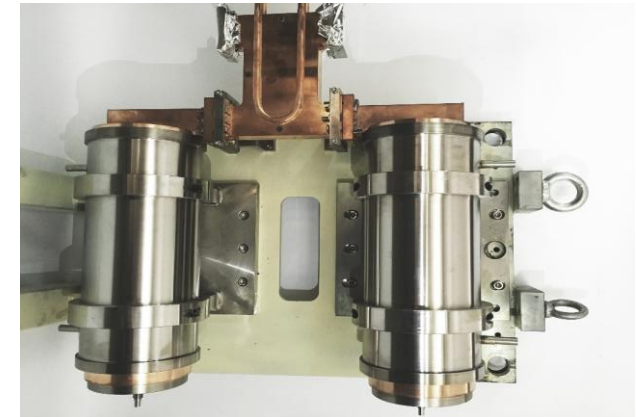
TE_{01N} field pattern.



S-band, KEK, ATF&KEKB, Japan 1992.

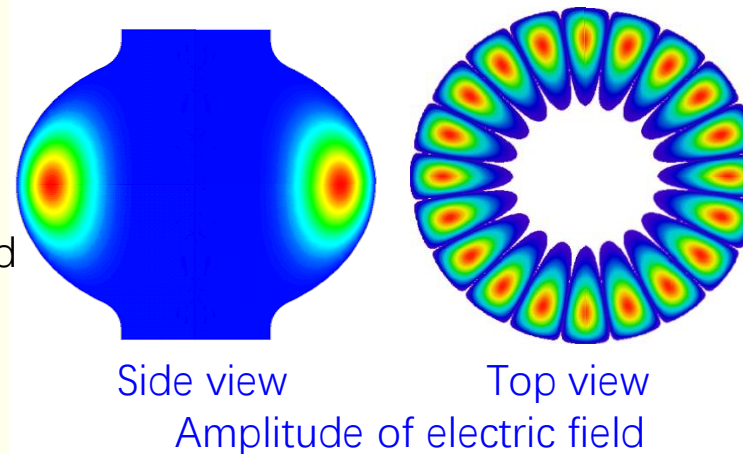
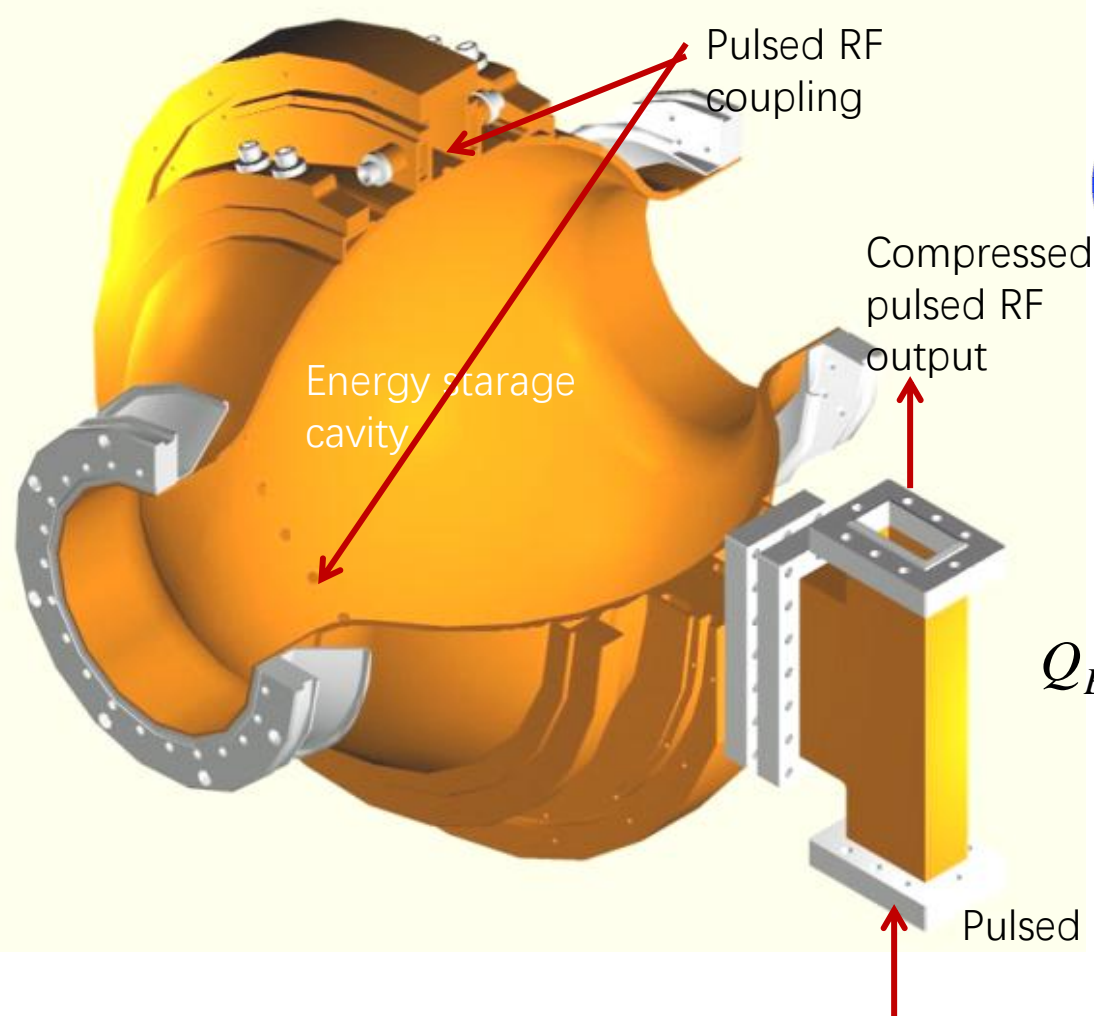


S-band, Pohang, Korea, 1994

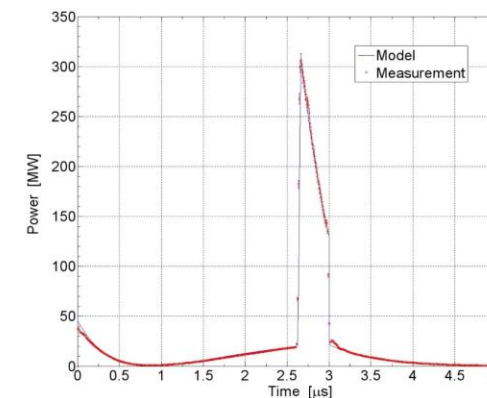
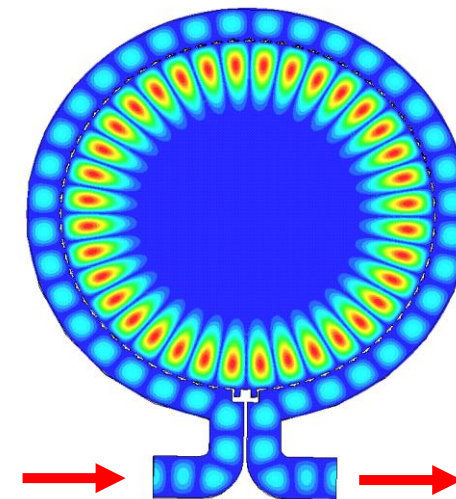
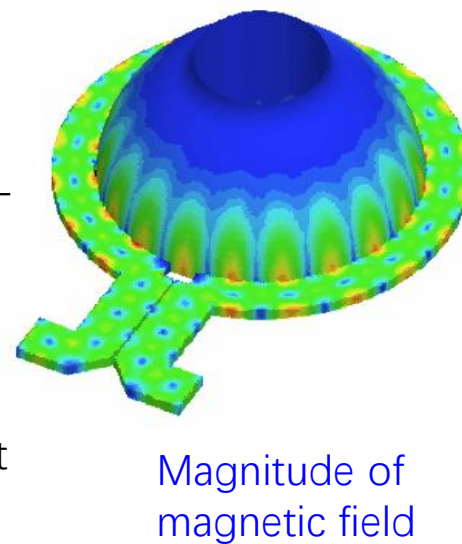


C-band, SXFEL, Shanghai, 2016高等研究院
NESE ACADEMY OF SCIENCES

Barrel Open Cavity (BOC)

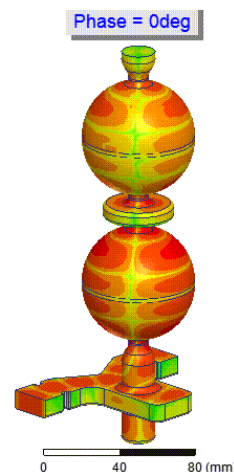
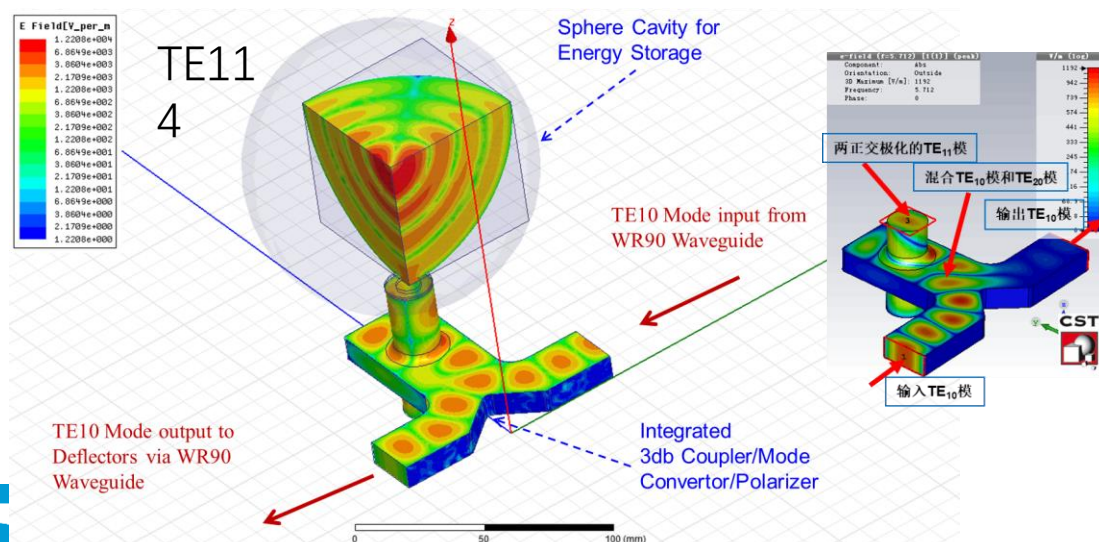
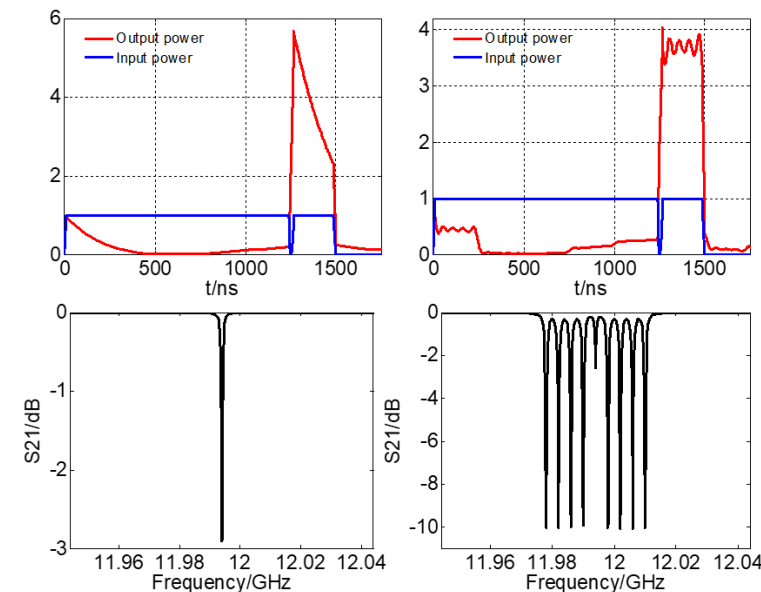
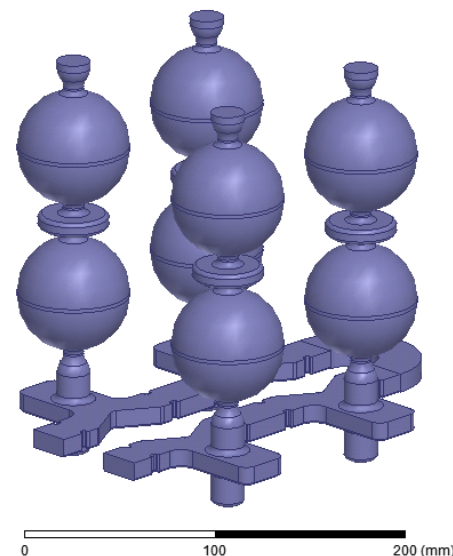
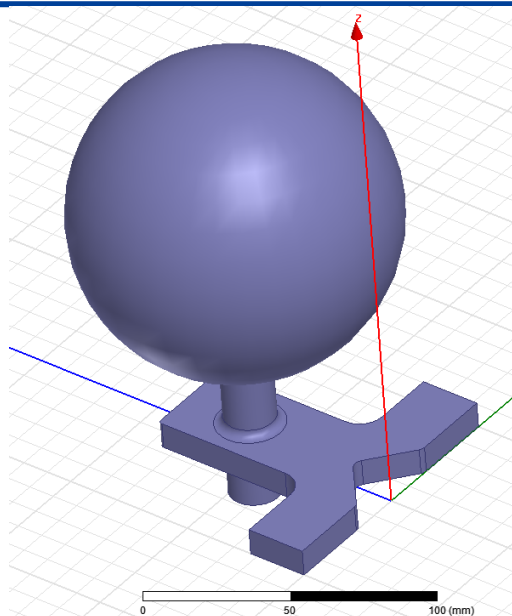


$$Q_E = \frac{a}{\sigma_s}$$

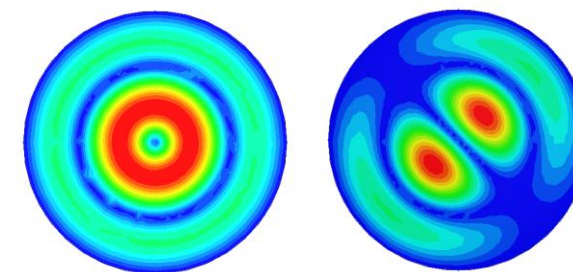


An example of Whispering Gallery Mode cavity:
 $TM_{10,1,1}$, $Q_0 = 197800$, $f = 2.9985$ GHz, $\beta = 6$

New spherical Pulse compressor (Flat-top output)



- Cavity operation mode TE012



RF system of high gradient accelerating structure

Accelerating structure



Pulse compressor



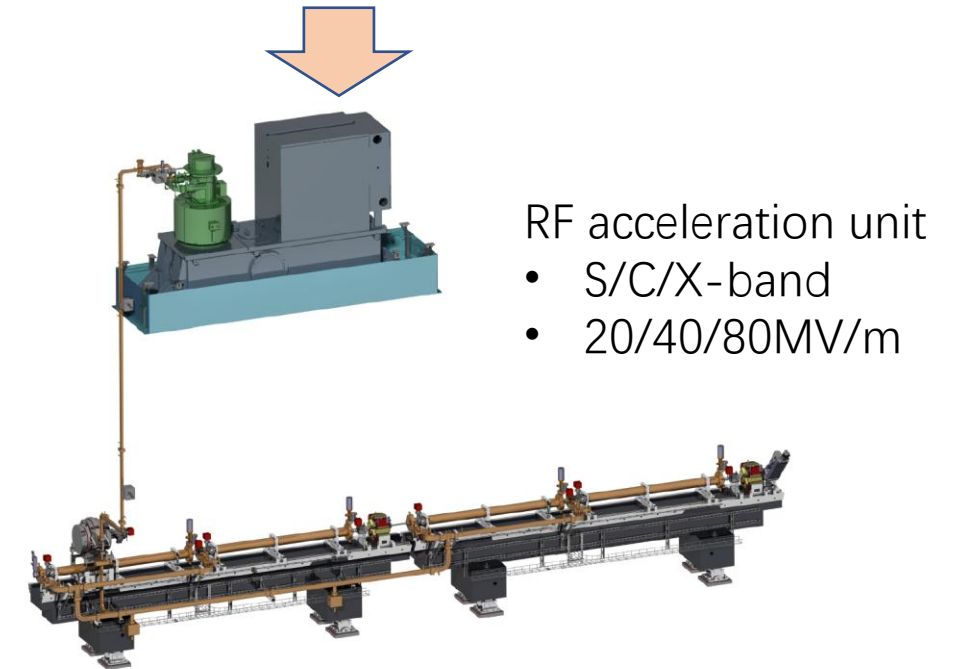
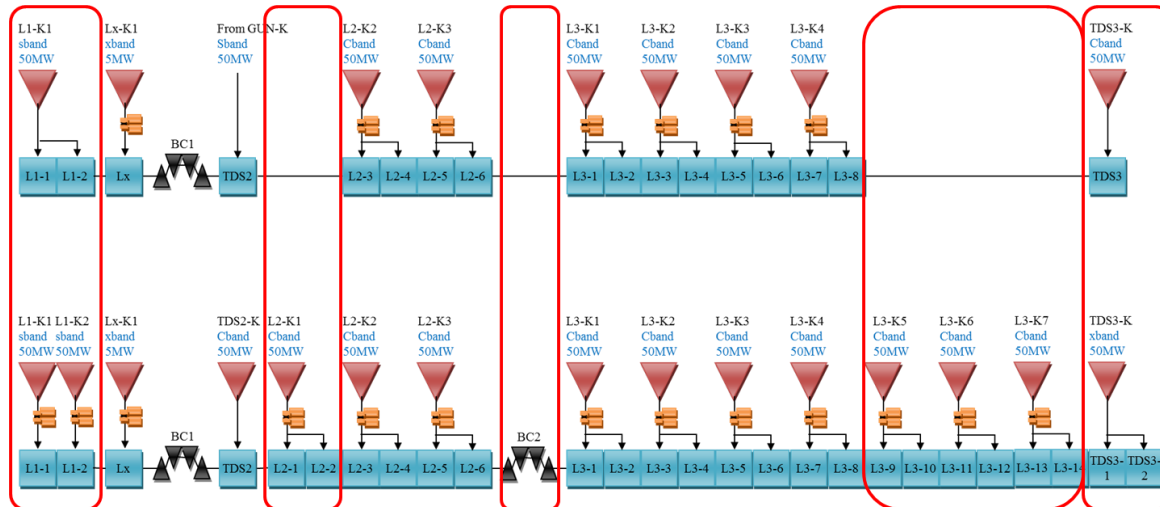
Waveguide system



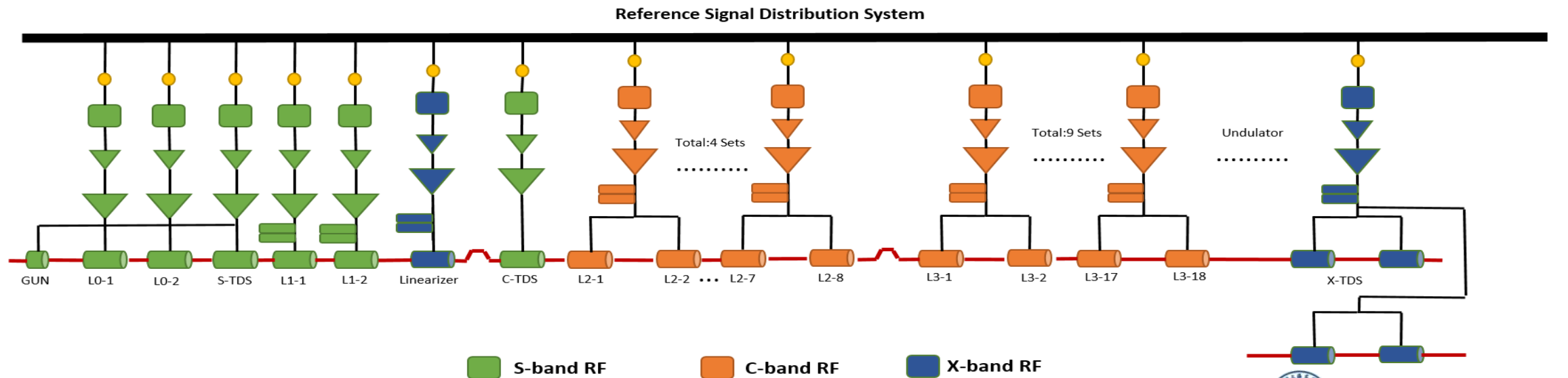
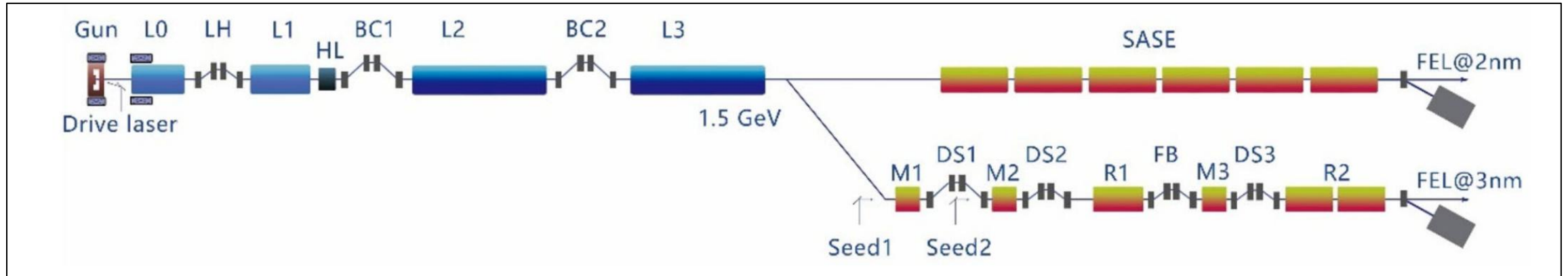
Klystron



Low-level RF control



Linac example: NC RF system at SXFEL



S-band accelerating structures at SXFEL

❑ Single port-> dual port, race track -> dual-port, dual mode.

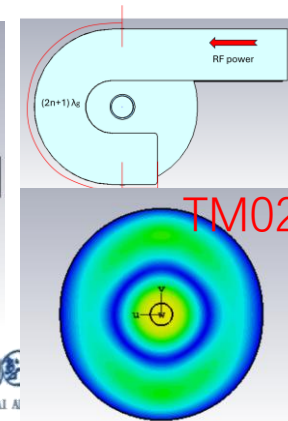
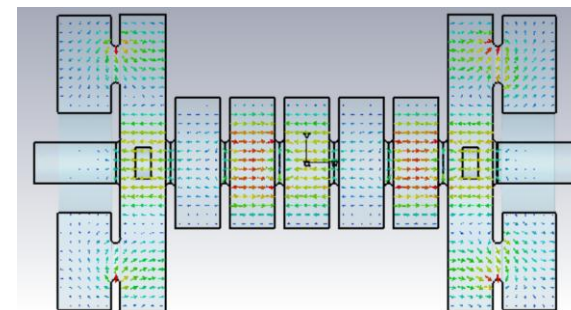
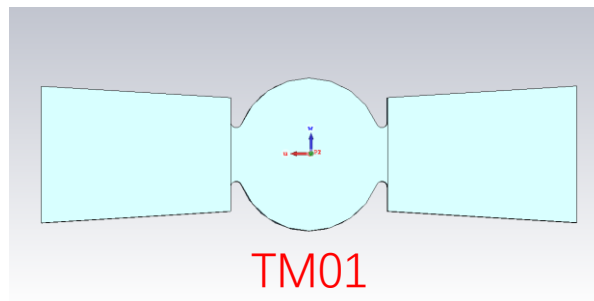
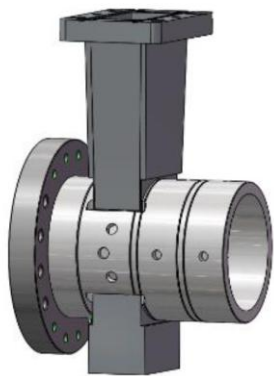
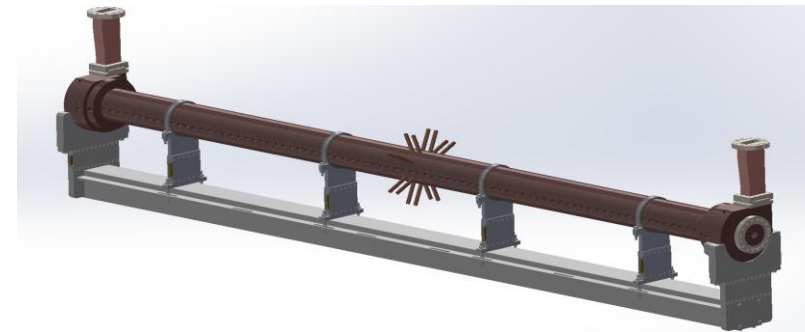
Single-port, side coupling
Outer water cooling
Multi-pole field
Max gradient: 17MV/m



Dual-port, side coupling
Racetrack coupler
Assembling asymmetrically
Inner water cooling
Max gradient: 23MV/m

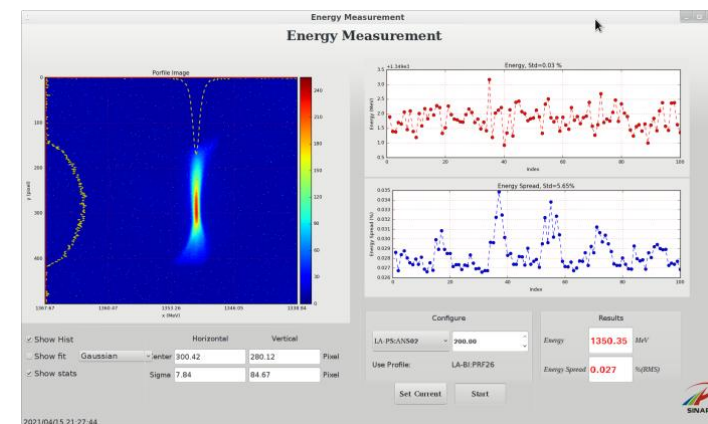
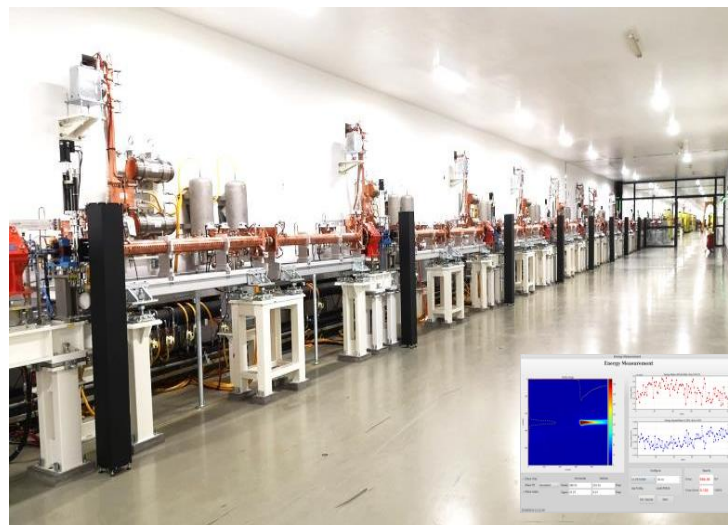
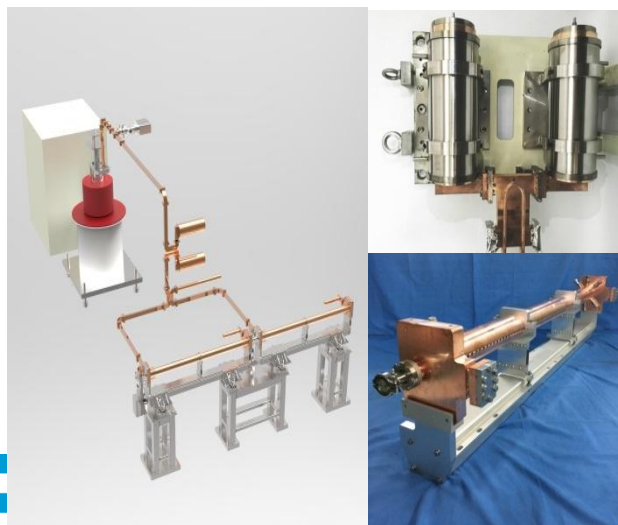
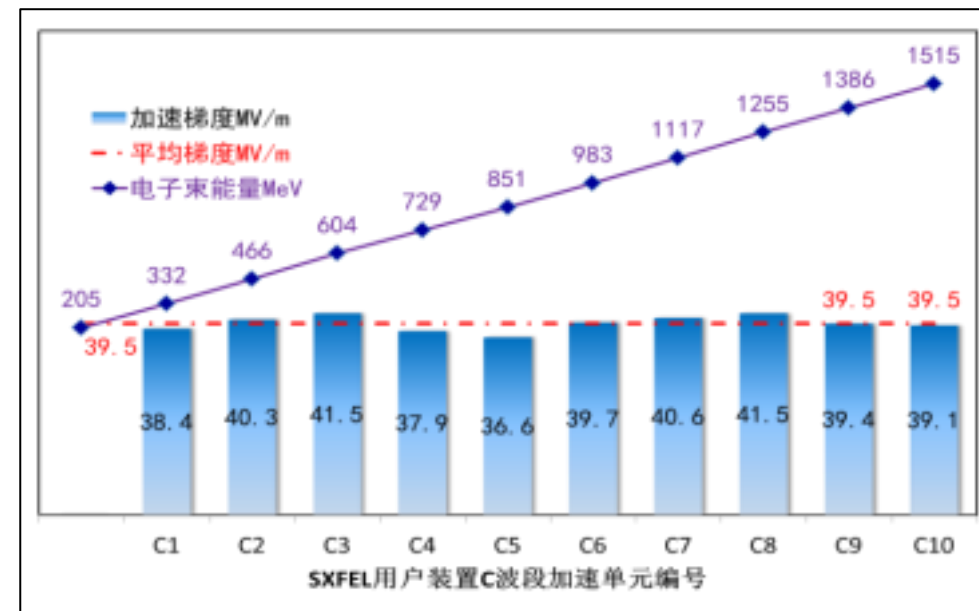
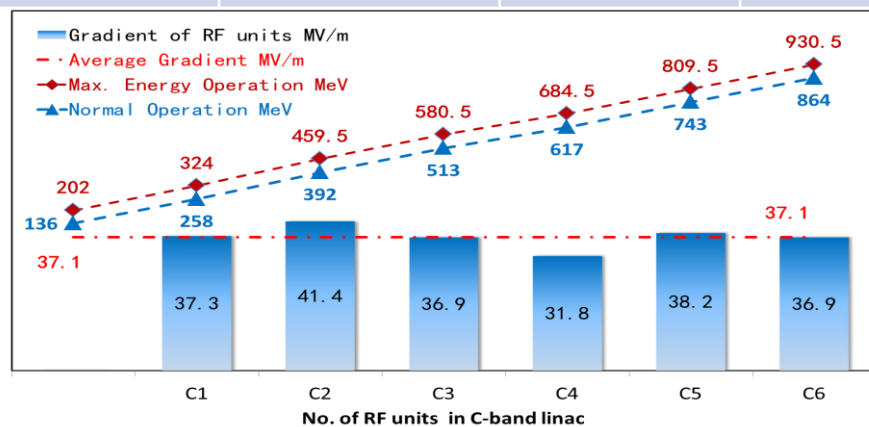


Dual feed, one-port
Racetrack coupler
Perfect symmetry
Inner water cooling
Max gradient: 23MV/m



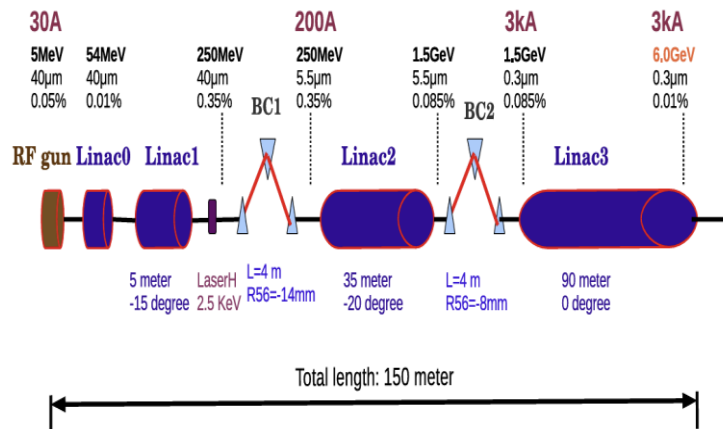
C-band accelerating structures at SXFEL

SWISSFEL*	SACLA**	SXFEL-TF	SXFEL-UF
28 MV/m	35 MV/m	37.1 MV/m	39.5 MV/m

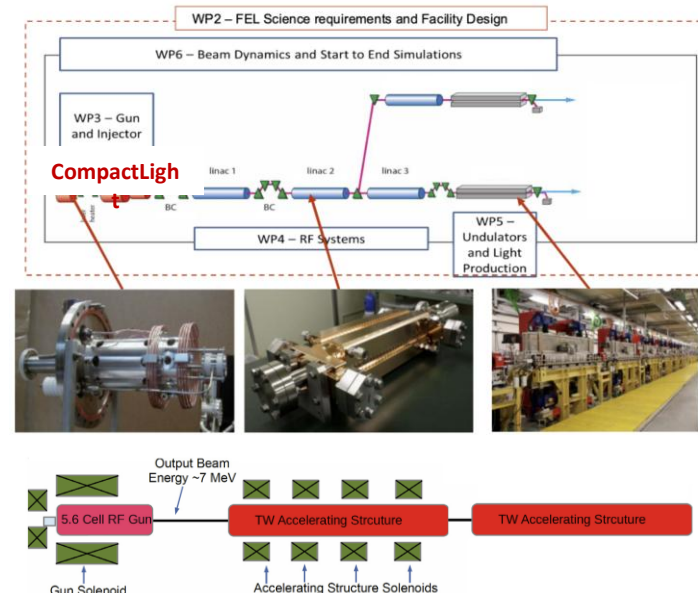


Linac example 2: Compact XFELs based on X-band technology

- A 6-GeV hard XFEL based on X-band photoinjector and linac is proposed by SLAC in 2012.
- The CompactLight design study, launched by a team of 22 International Laboratories and two Industries in January 2018. The main linac is based on very high-gradient X-band accelerating structures.
- SXFEL, LCLS, FERMI ELETTRA, Swiss FEL and PAL XFEL used X-band accelerating structure and deflecting structure for linac and diagnostic.

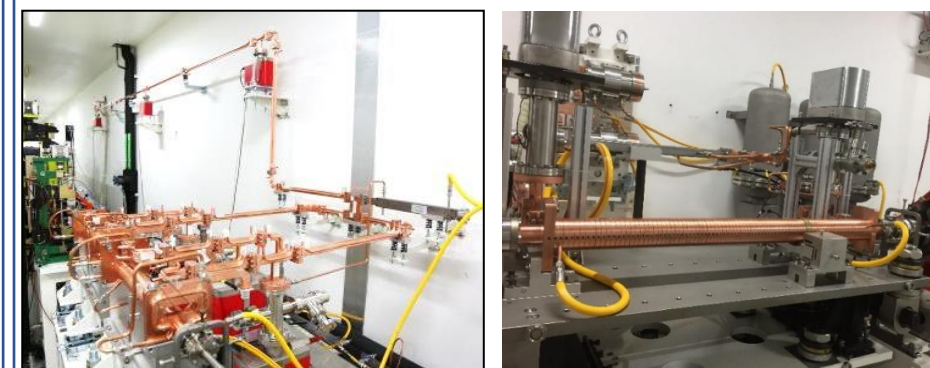


A 6 GeV Hard XFEL based on X-band at SLAC
Y. P. Sun, et al. PRAB 15 (2012): 030703



CompactLight based on X-band RF gun and linac

D. González-Iglesias, et al., NIM, A 1014 (2021) 165709.



X-band accelerating structure and deflecting structure applied on SXFEL for linac and diagnostic.

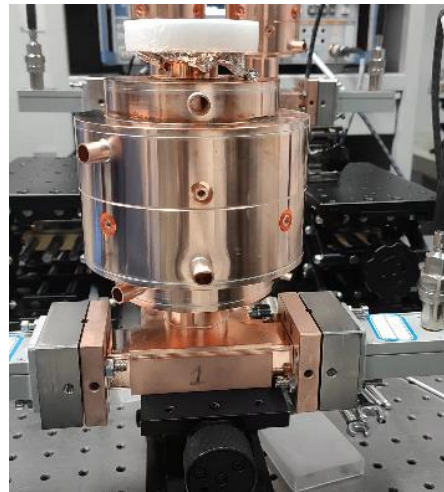
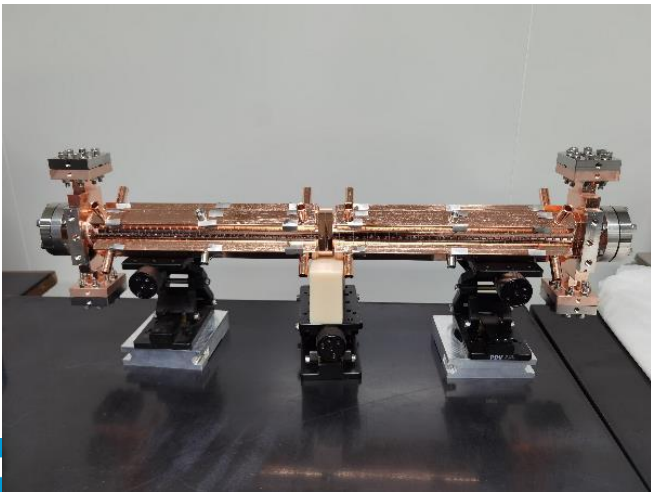
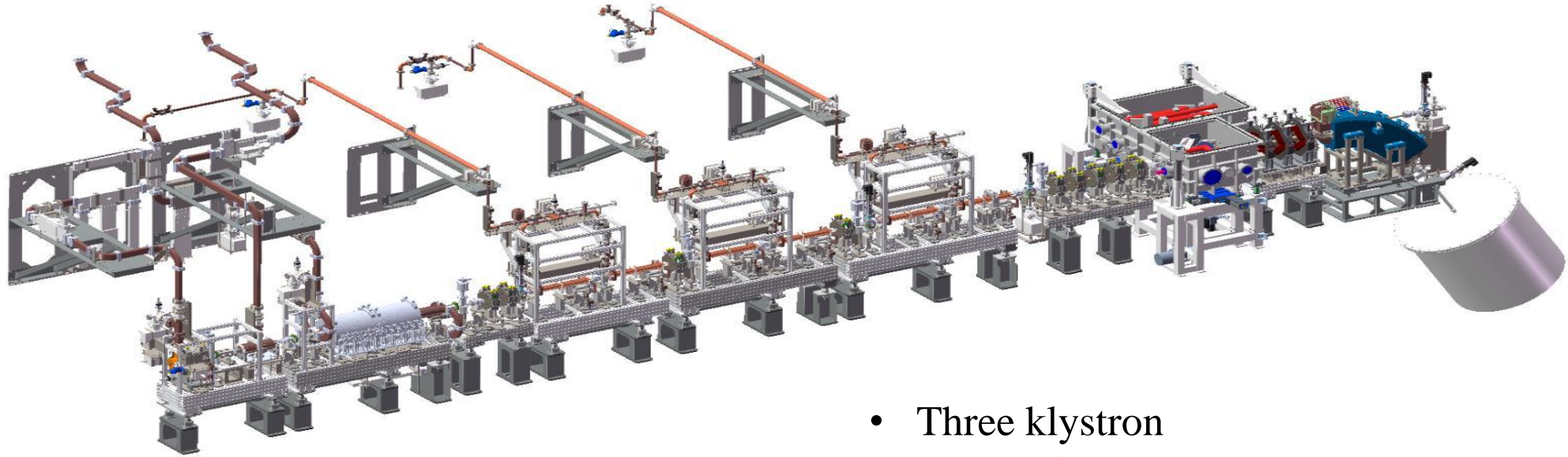
X. X. Huang, et al., NIM, A 854 (2017) 45-52.

J. H. Tan, et al., NIM, A 930 (2019) 210-219.



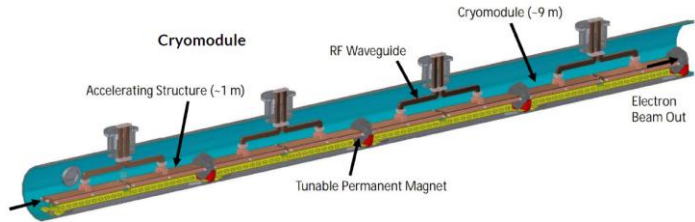
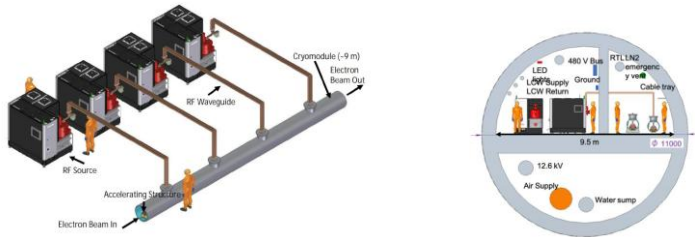
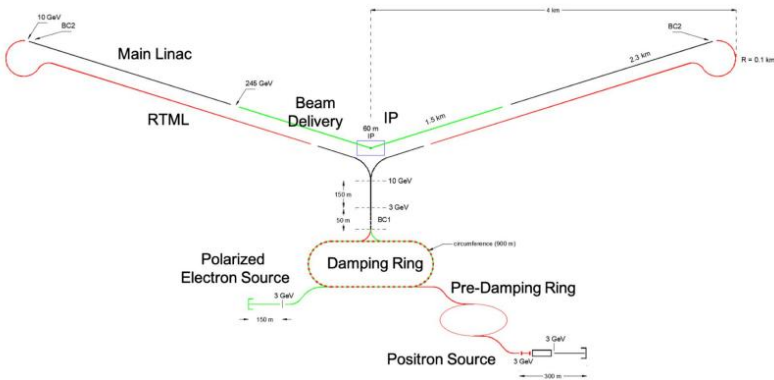
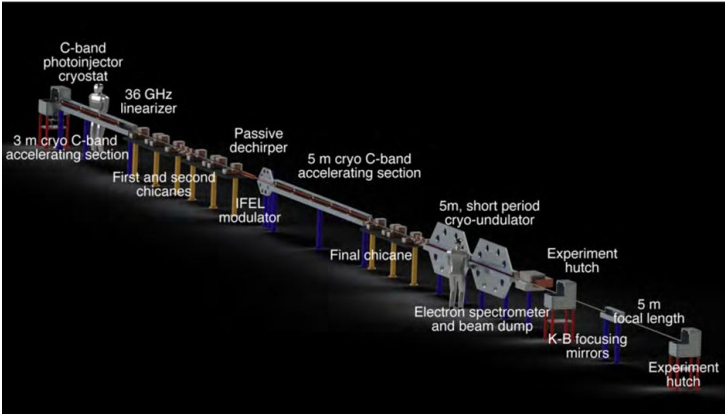
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Linac example 3: X-band linac for VIGAS at Tsinghua University

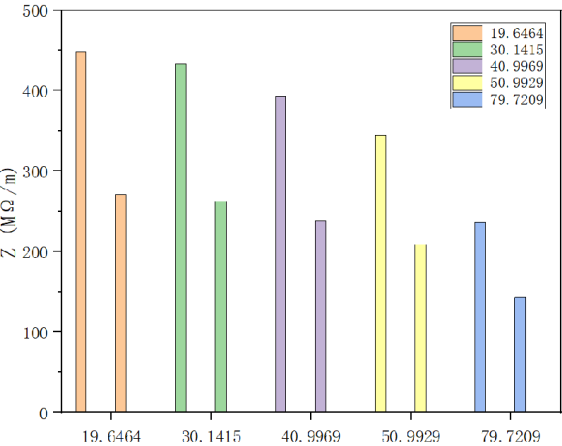


- Three klystron
 - Each 50 MW, 1.5us
- Three pulse compressor (SLED I type)
- Six X band high gradient structures
 - Average gradient ≥ 80 MV/m
 - Energy gain per structure > 50 MeV
 - Filling time < 150 ns
- 91 MW at Xacc w/ PC gain factor as 4.5

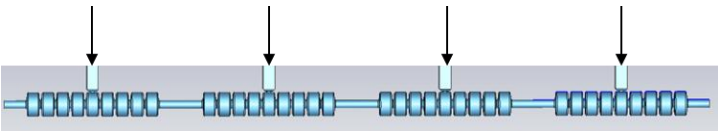
Next: Cryogenic high gradient accelerating structures



Cryogenic for ultra compact FEL



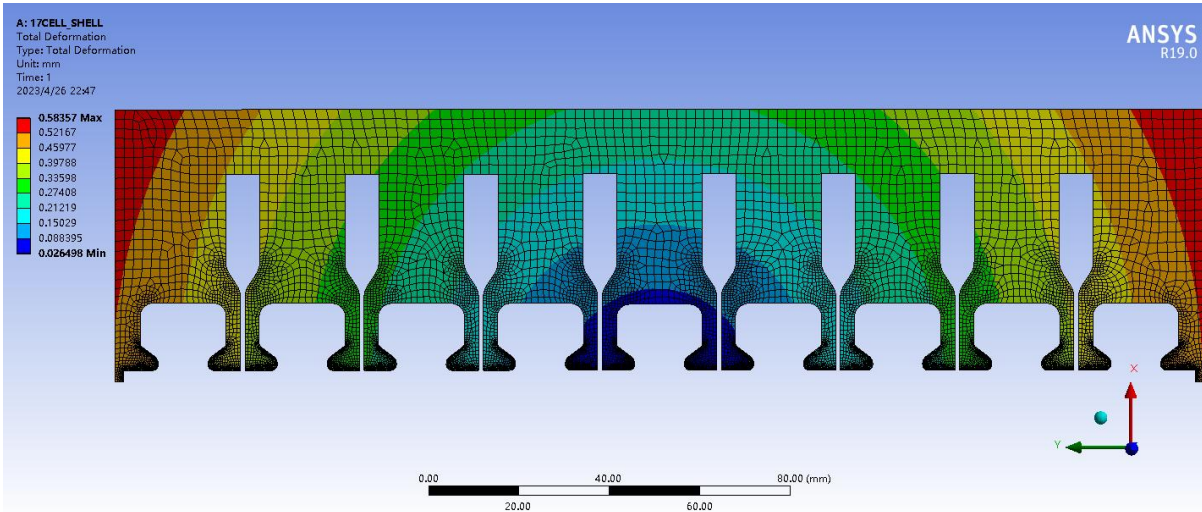
- C3: Cooled Copper Collider
- total length: 33 km->8 km, 250/550GeV



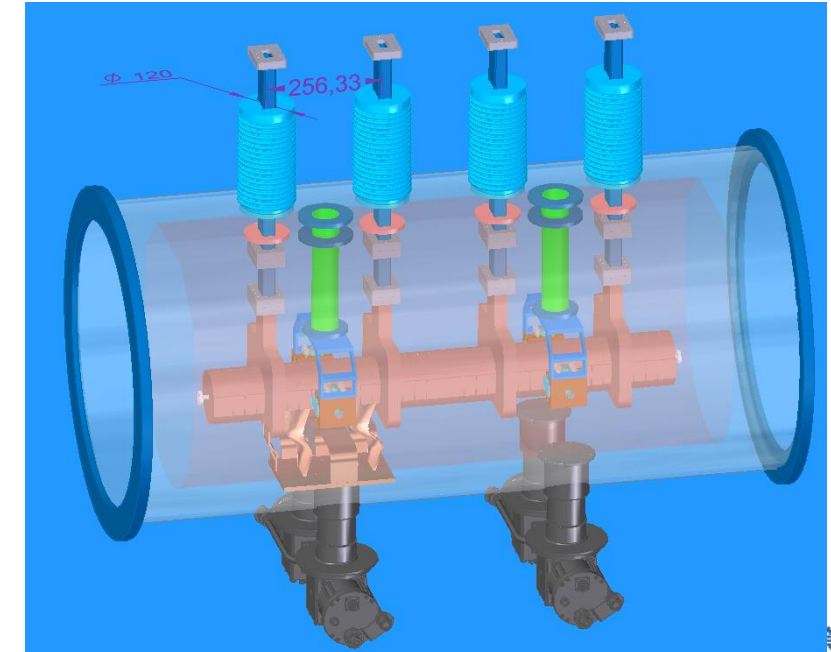
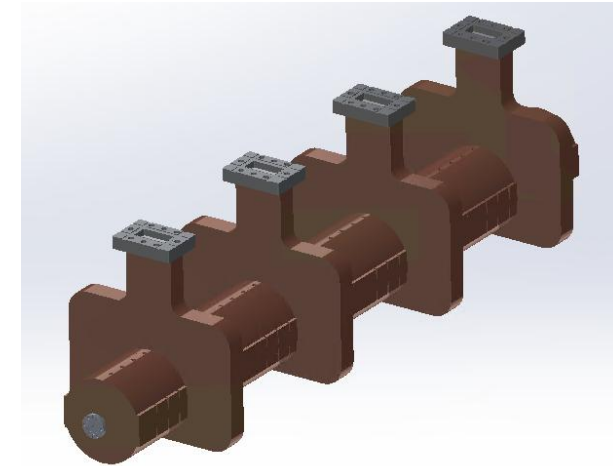
Parameter	Value	Unit
Frequency	5712	MHz
Temperature	20-50	K
Shunt impedance@20K	0.0037	Ω
Gradient E_{acc}	65/80	MV/m
Material	Copper	
RRR	500	
Cooling	Helium	

- C-band cryogenic module for C3
- 460M Ω/m
- 20-80K (27K for UC-XFEL; 80K for C3)
- Peak electrical field: 140MV/m

Cryogenic RF cavity



- ❑ The structure dynamic thermal load is 80W@40K.
- ❑ Static thermal load 20W/40K.
- ❑ Plus 50%, total thermal load 150W@40K.
- ❑ The number of chillers is 4.
- ❑ For each chiller, the length and thickness of the cooling belt of the structure are calculated according to 75W@30K
- ❑ The fabrication of first cryogenic system prototype will carried out recently.



Some types of RF cavity

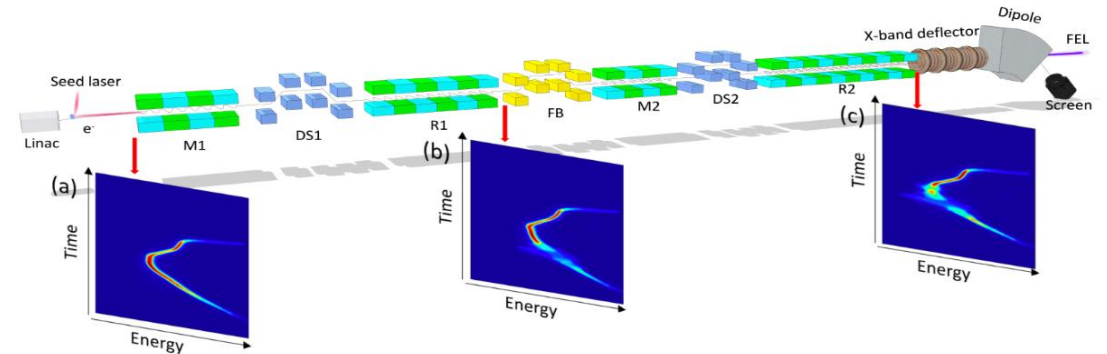
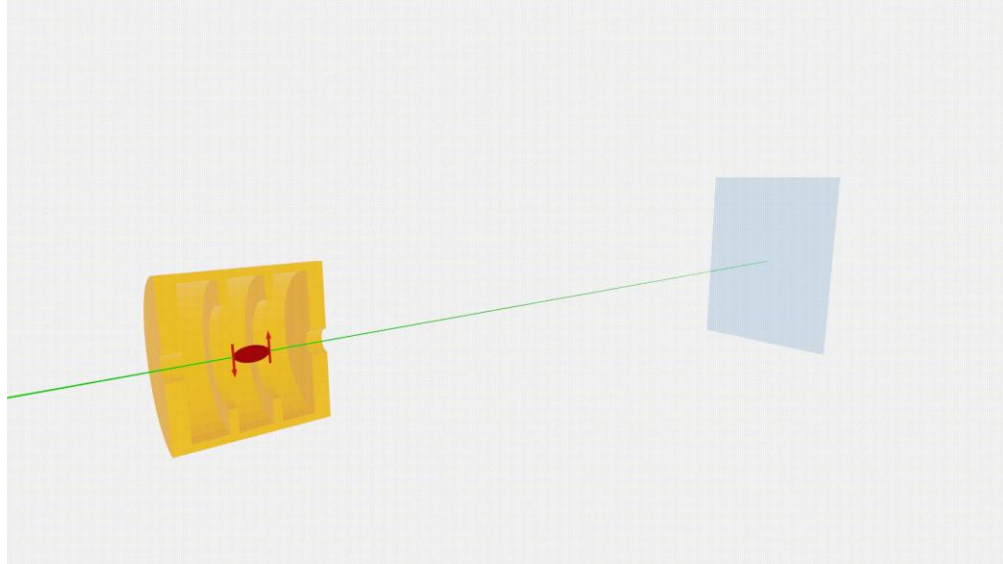
- Accelerating structures
- Deflecting RF cavity**
- NC Single-cell RF cavity

Deflecting RF cavity for beam diagnostics

Principal of operation of the TM11 transverse deflecting RF cavity to measure bunch length, diagnose beam-laser synchronize and real-time lasing spectrum on a profile monitor.

$$eV_{\perp} \approx n \frac{\lambda}{2\pi c \Delta t} \sqrt{\frac{\epsilon_N E m c^2}{\beta_d}}$$

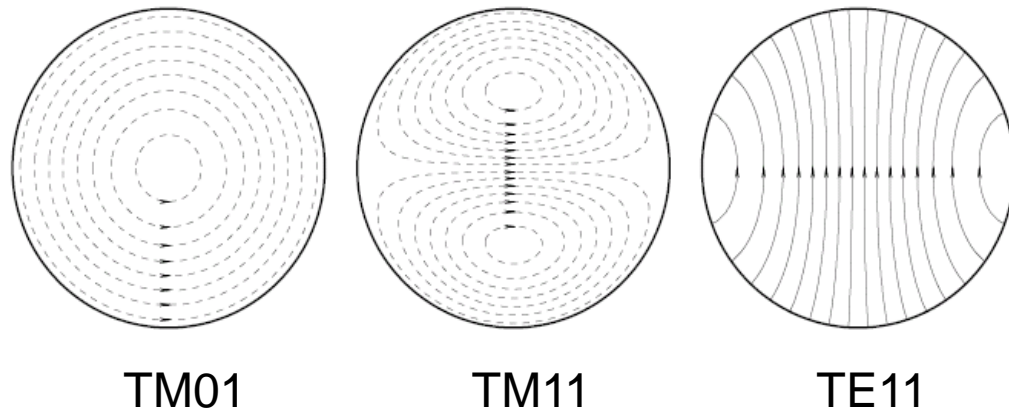
V_{\perp} : Deflecting voltage
 Δt : time or length resolution
 $1/\lambda$: frequency of cavity
 ϵ : Emittance
 E : beam energy
 β_d : twiss parameter



- ❑ Beam-laser synchronization: 10fs
- ❑ Real-time FEL lasing diagnostics

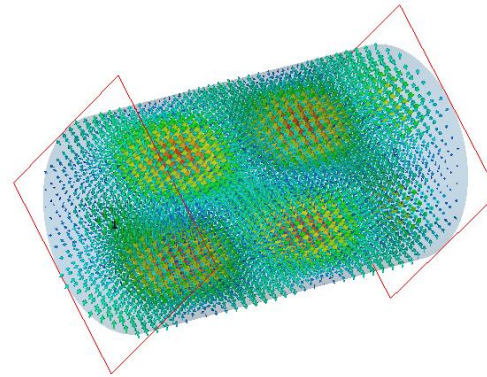
RF design of deflecting cavity

RF mode in the deflector is mixed with **TM11** and **TE11**, called **HEM11**. TM11 is located in the middle of cell, and TE11 is located on the iris tip. The structure is constant impedance.



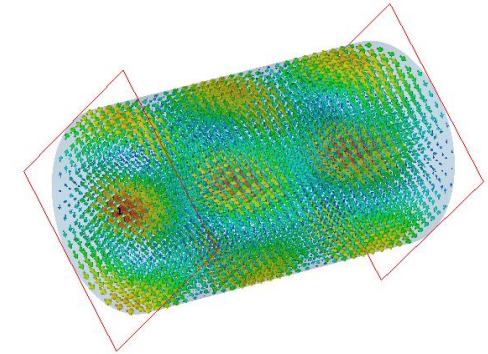
1) [1 (7)]
14000 MHz
0°
1865.61 V/m

HEM11 E-Field



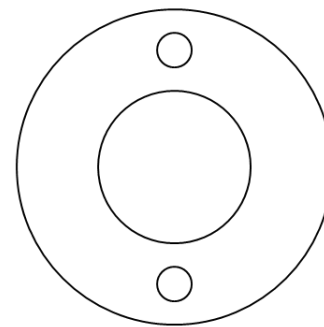
V/m
1800
1600
1400
1200
1000
800
600
400
200
0

HEM11 H-Field

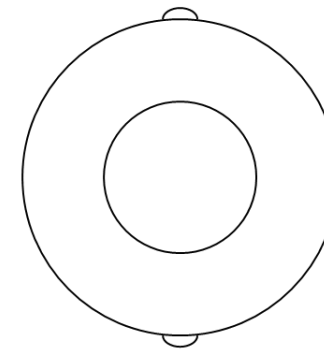


A/m
4.98
4.4
4
3.6
3.2
2.8
2.4
2
1.6
1.2
0.8
0.4
0

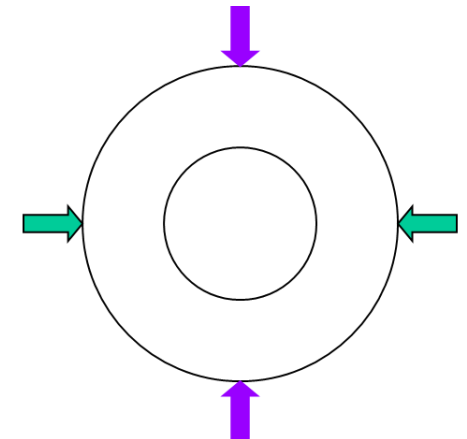
≈1.4e+4 [1 (7)]
14000 MHz
0°
(Plot) 4.98211 A/m



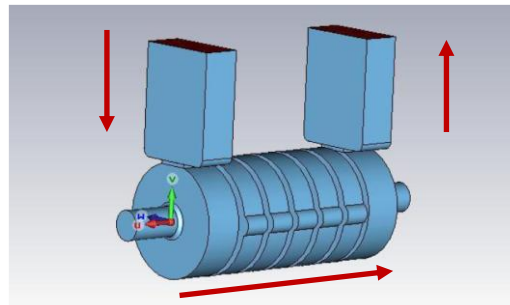
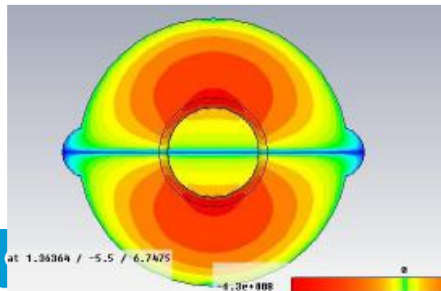
Two holes
(LOLA Structures)



Two caved-ins
on cell ID surfaces

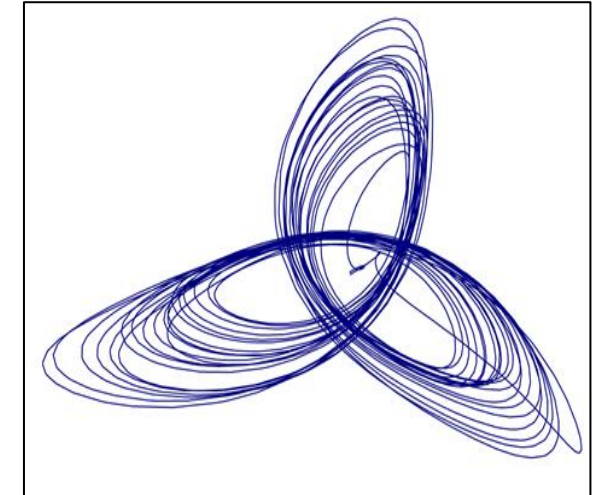
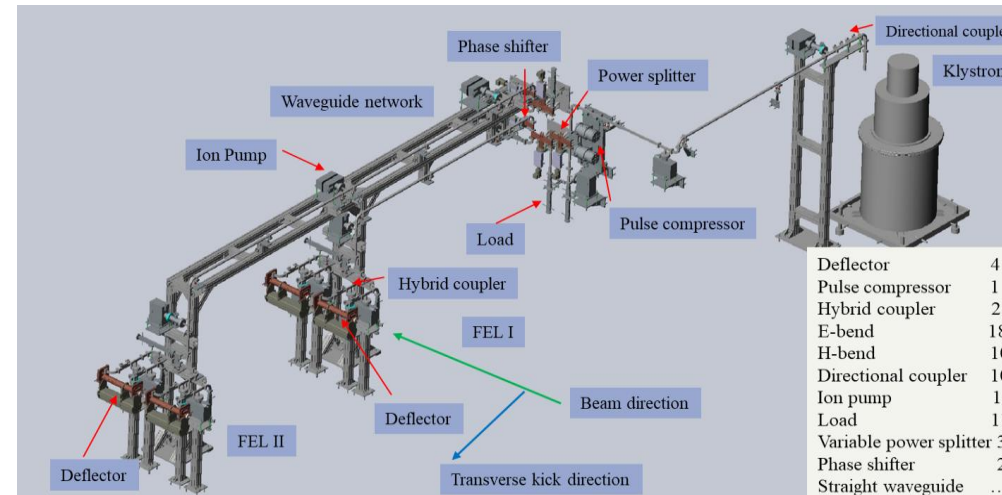
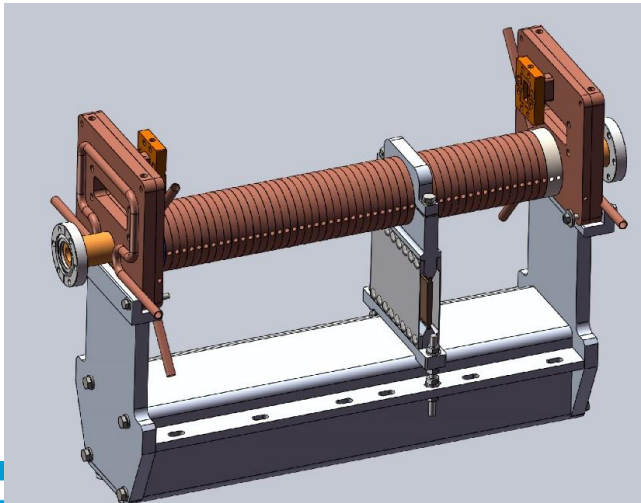
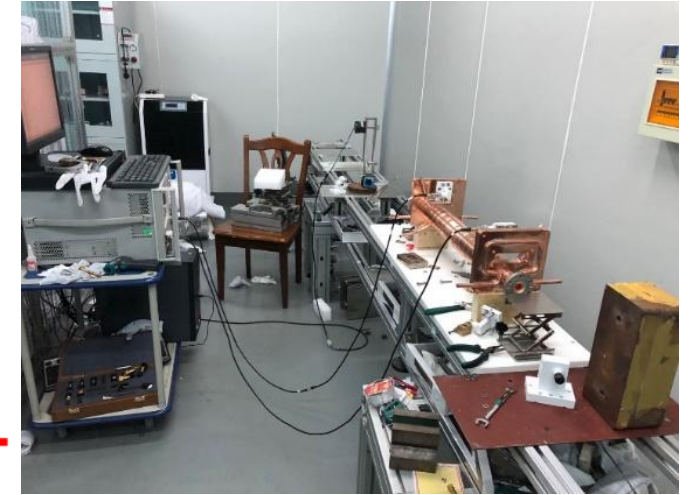
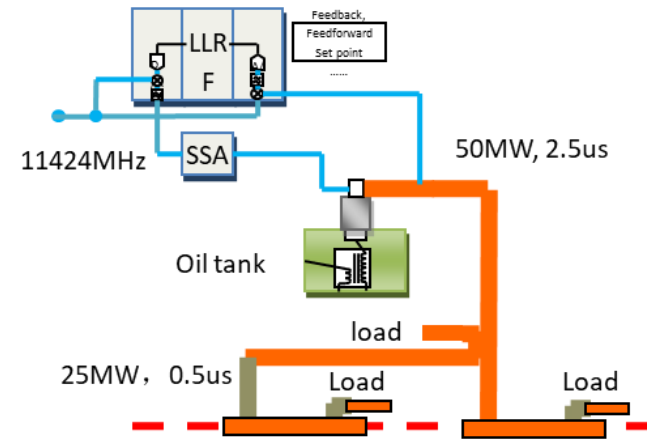


Deforming using
two more tuning holes

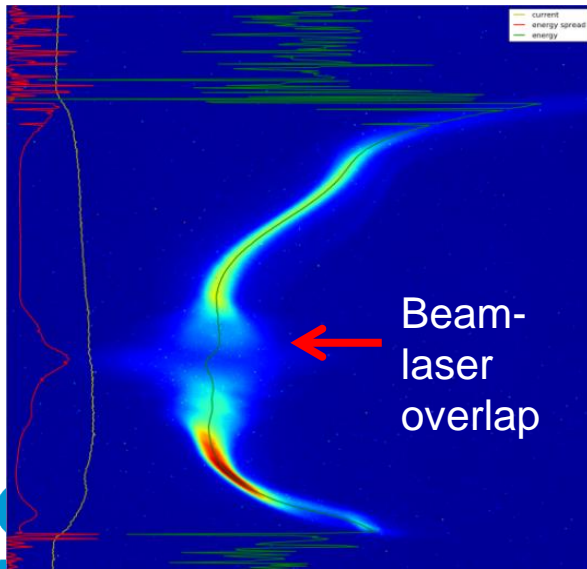
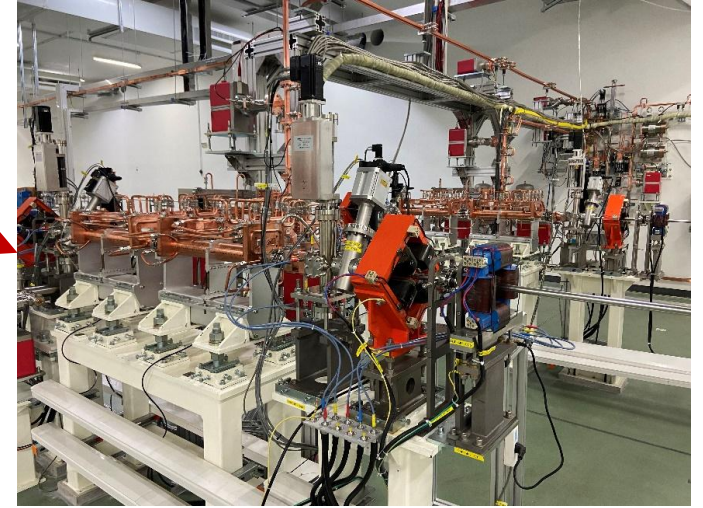
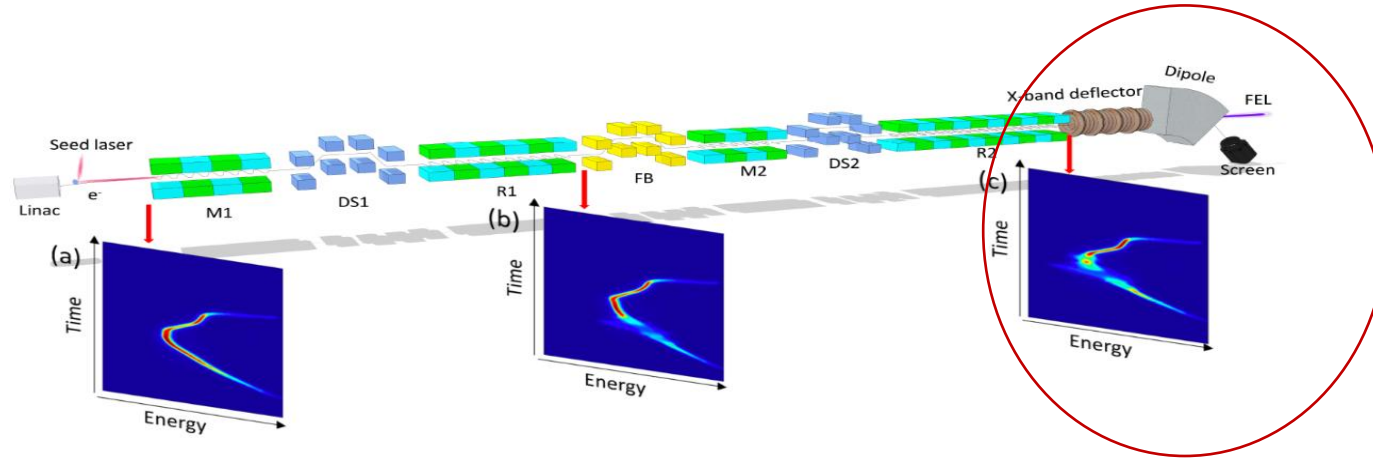


X-band deflecting cavities in SXFEL

- ❑ Dual-coupler design for new deflector
- ❑ One power unit for SXFEL
- ❑ Two deflector units for SXFEL

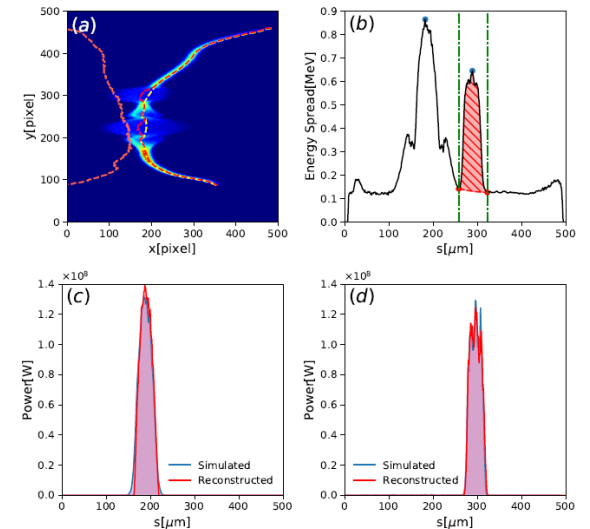
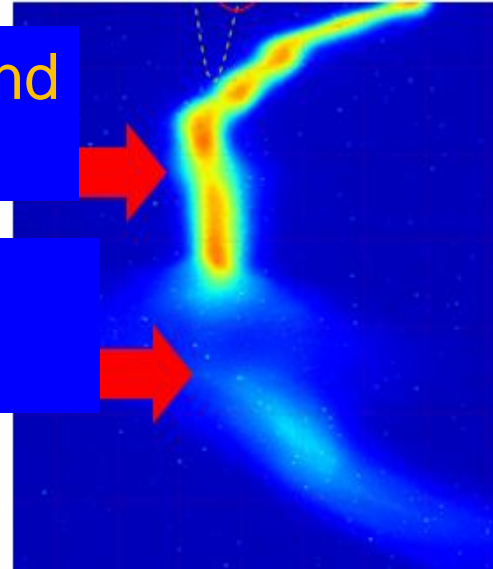


Beam-laser overlap and lasing feature real-time diagnostics at SXFEL



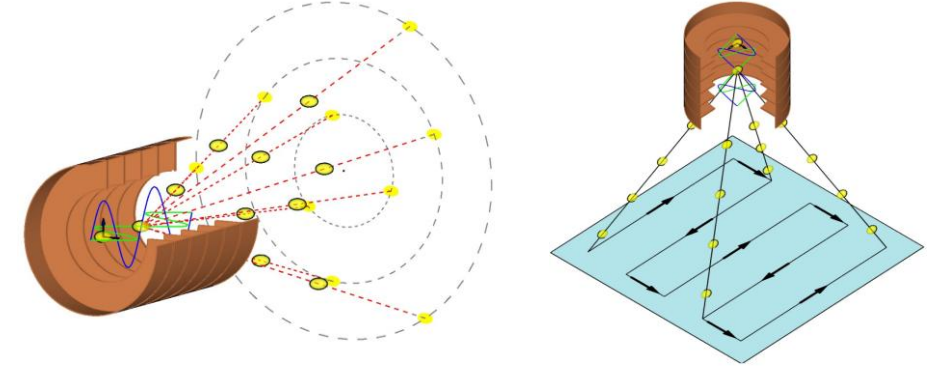
Second stage

First stage

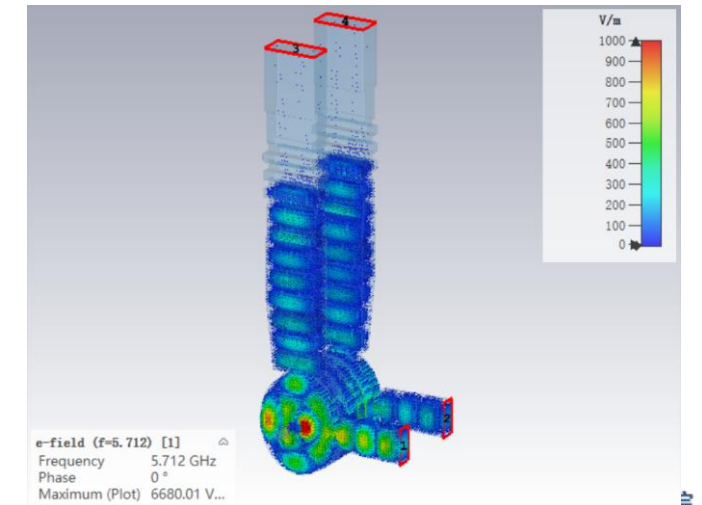
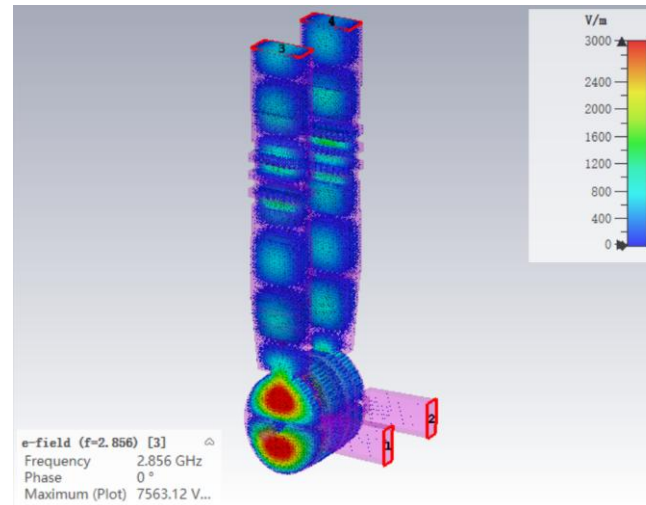
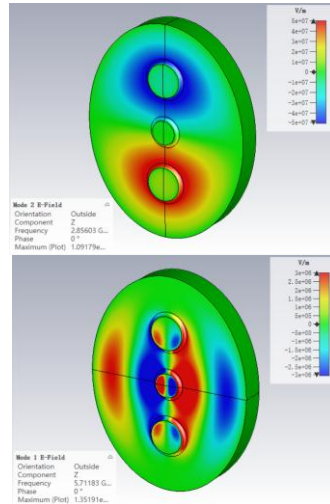
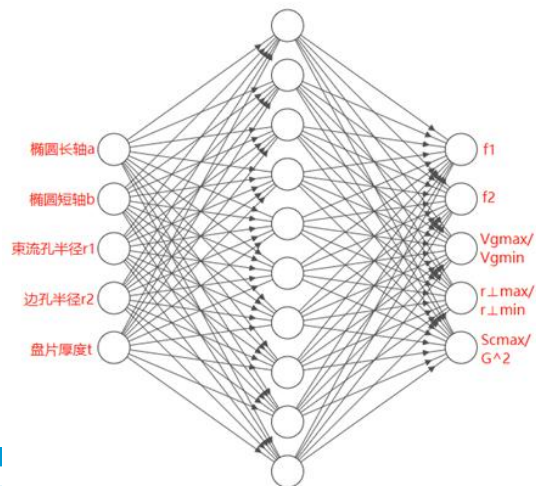


New Dual-mode deflector for variable polarization

1. HEM01 on 2856 MHz, HEM02 on 5712 MHz
2. 6-D (x, x', y, y', z, z') Tomography, FLASH scanning
3. Neural network and MOGA for optimize the RF structure
4. Two-port for coupling two modes independently.



PRAB Editor's suggestion



Some types of RF cavity

- Accelerating structures
- Deflecting RF cavity
- NC Single-cell RF cavity

Normal conducting single-cell RF cavity for light source

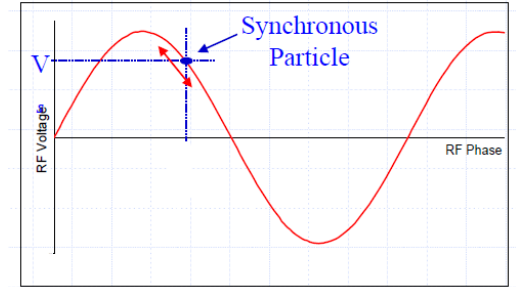
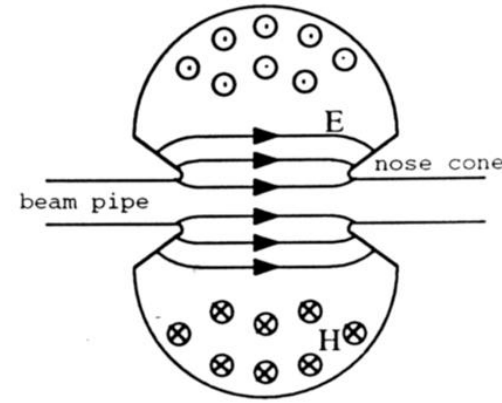
ID : The insertion device generates various characteristic types of SR.

BM : The bending magnet bends the beam and generates SR.

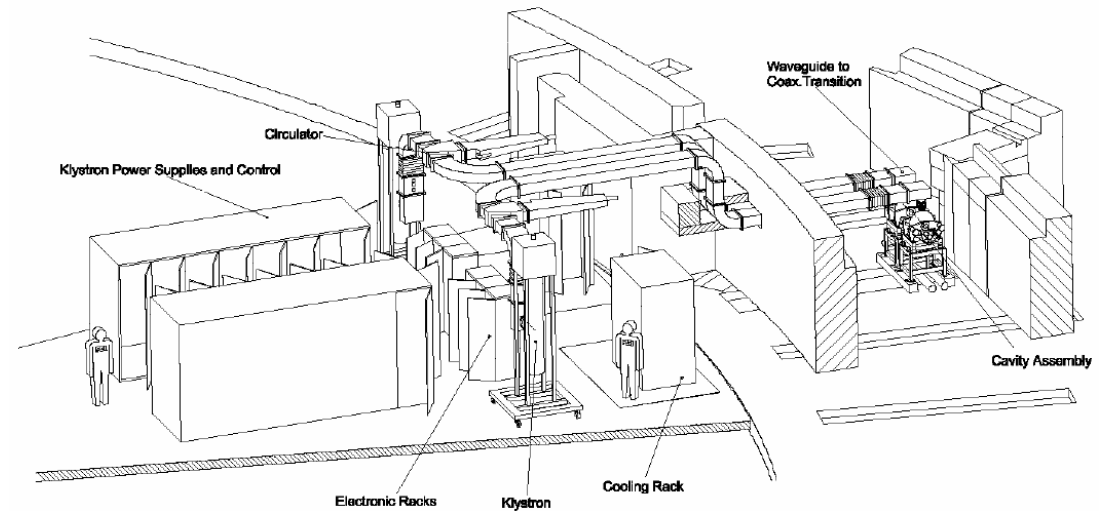
rf accelerating cavity : The cavity replenishes the stored beam with energy lost by the generation of SR.

Vacuum Duct : The pressure in the duct is kept below 10^{-10} Torr in order to reduce beam decay caused by collisions with residual gas.

QM : The quadrupole magnet works as a lens to focus the beam.



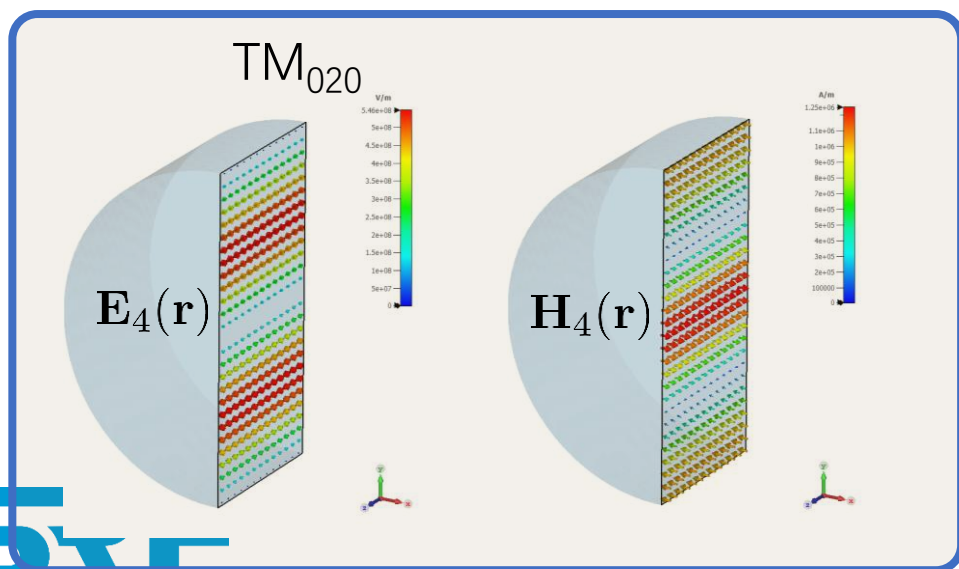
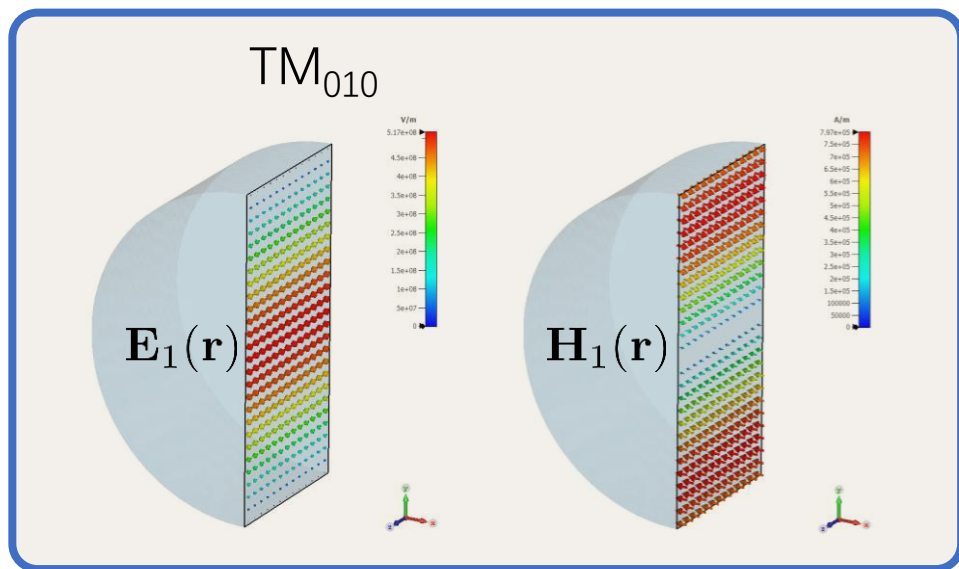
$$V_{rf} = V_c \sin(\omega_{rf}t + \phi)$$



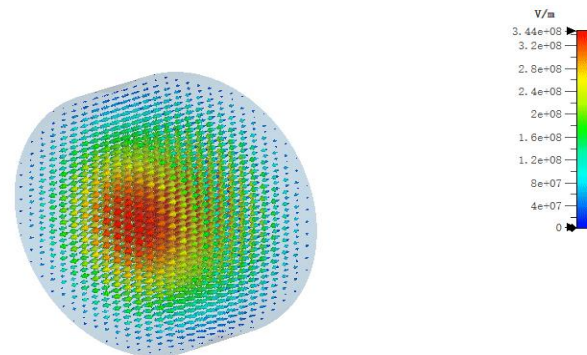
Function of RF cavity in light source

- ❑ Delivers the energy to electrons for compensating the energy loss due to Synchrotron radiation and interactions with beam chamber impedance, or ramping the electron beam to higher energy;
- ❑ Establishes RF voltage to capture and focus the electrons into bunches;
- ❑ Controls beam parameters, such as bunch length, beam lifetime, etc.;
- ❑ Provides damping effects to the electron motions by synchrotron radiation and RF acceleration.

Standing wave “Pillbox” model of single-cell RF cavity



TM010
E-Field



$$E_z = E_0 J_0(k_c r) \cdot \cos(\omega t - \beta z)$$

TW+

+

$$E_z = E_0 J_0(k_c r) \cdot \cos(\omega t + \beta z)$$

TW-

||

$$E_z = E_0 J_0(k_c r) \cdot \sin(\omega t)$$

SW

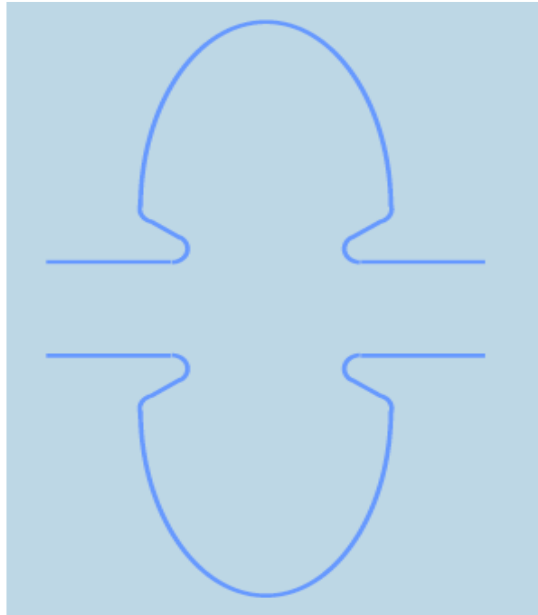
TM010
H-Field



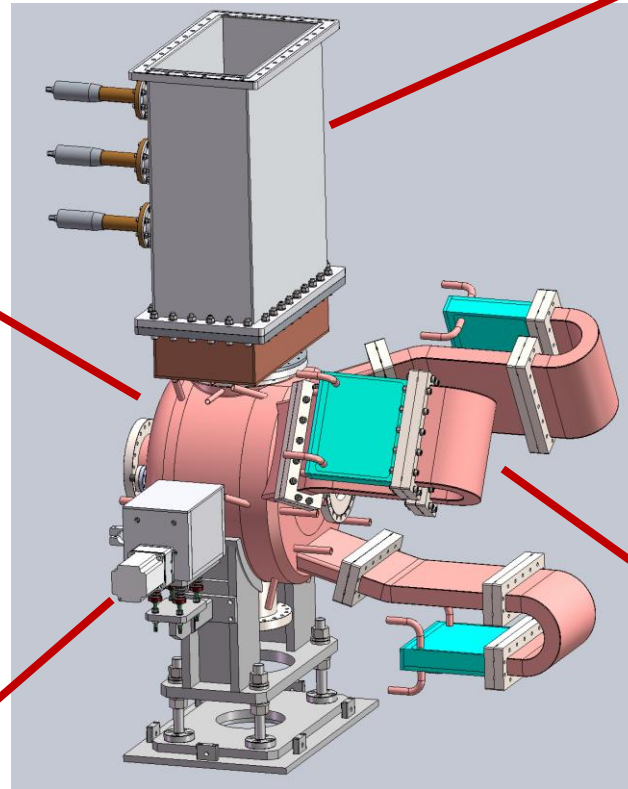
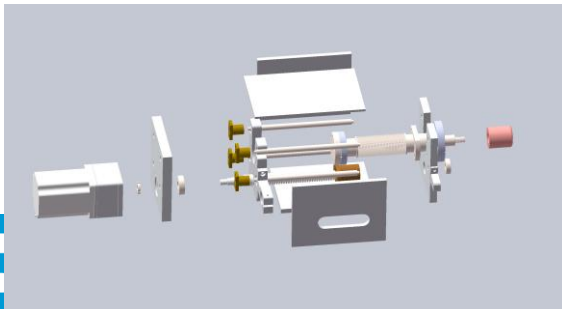
Mode 1 H-Field
Frequency 7649.52 MHz
Phase 0°
Maximum (Plot) 540312 A/m

Some key components of RF cavity

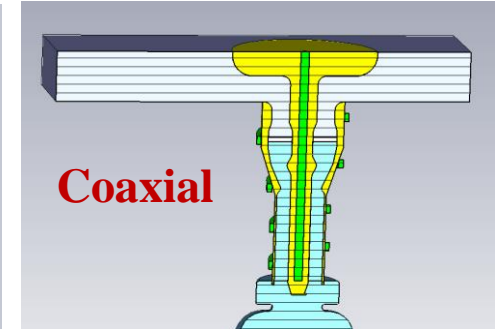
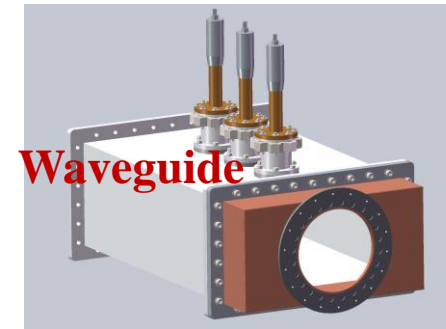
Main cavity, high R/Q



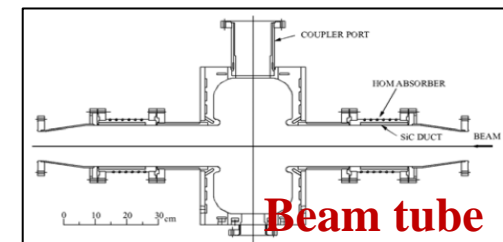
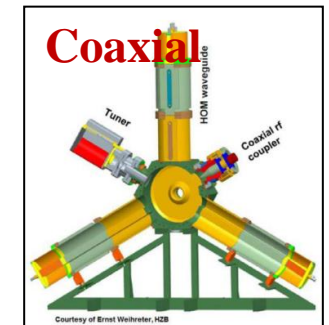
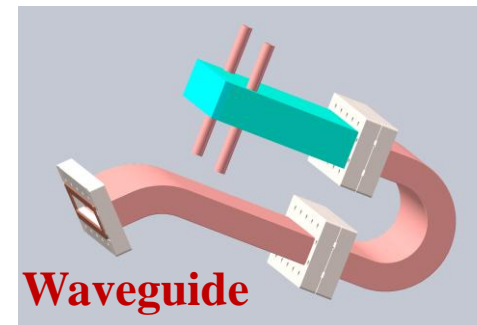
Tuner on frequency



Coupler, power input



HOM damper, three types



Some key parameters of RF cavity

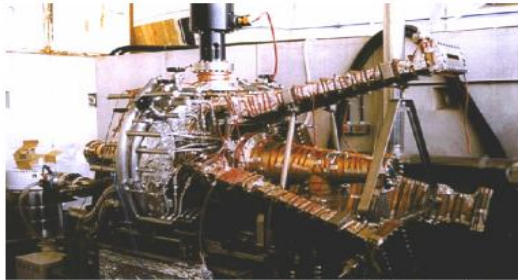
Shunt impedance $R = \frac{V_c^2}{P_c} = \frac{(E_0 \cdot L)^2}{P_c}$

Effective Shunt impedance $R_s = \frac{V_e^2}{P_c} = \frac{(E_0 \cdot T \cdot L)^2}{P_c} = R \cdot T^2$ T : Transition factor

Quality factor $Q_0 = \frac{\omega U}{P_c}$

Normalized impedance $R/Q = \frac{V_c^2}{\omega U}$

Some typic single-cell RF cavities

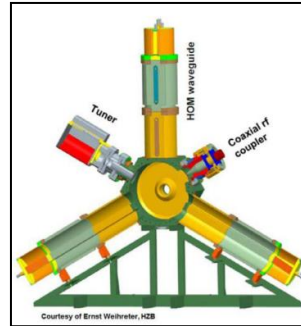


Daphne cavity, 368.2 MHz, 250 kV, 2 M Ω , L=1.9m



KEK ARES cavity, 509 MHz, 500 kV, 1.7 M Ω

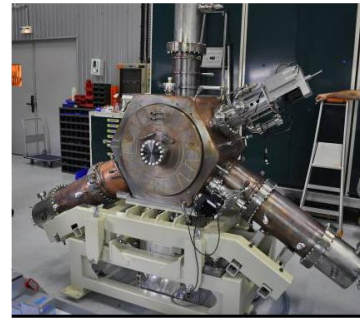
BESSY-II 500MHz



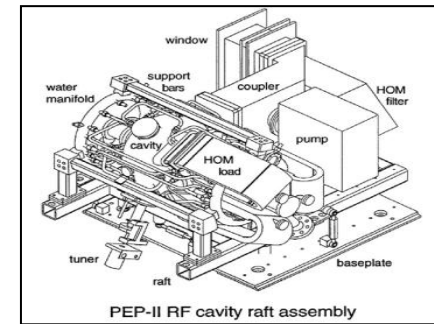
Courtesy of Ernst Wehner, HZB



ESRF 352.2 MHz

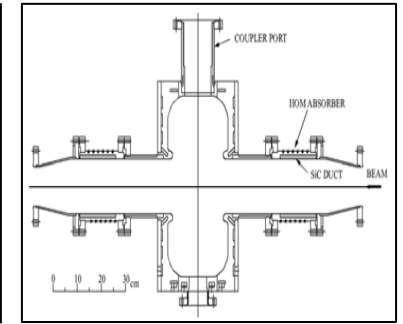


PEP-II 476MHz

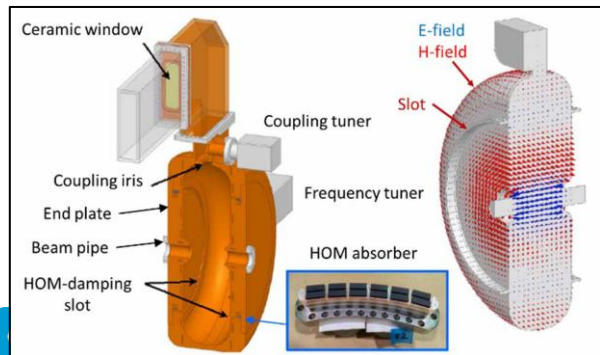


PEP-II RF cavity raft assembly

KEK-PF 500MHz



Spring-8 TM₀₂₀

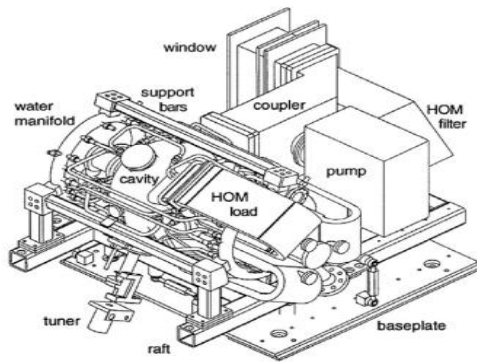
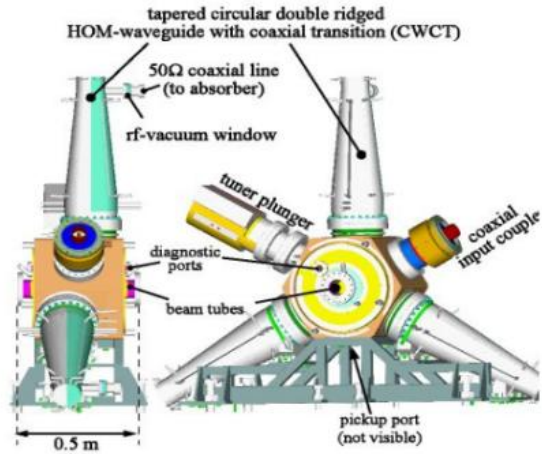
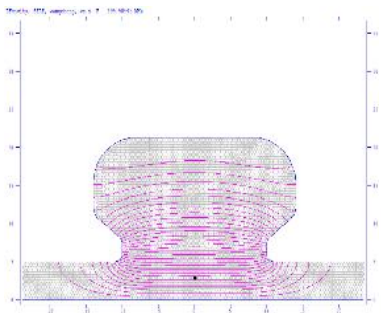


NC Cavities	f_0 MHz	V_{cy} kV	R_s M Ω	Q_0	P_{cy} kW	L m	$f_{HOM \parallel}$ MHz	Max. R_{\parallel} k Ω	$f_{HOM \perp}$ MHz	Max. R_{\perp} k Ω /m
PEP II	476.	850.	3.8	32400	103.	~1.5	1295.	1.83	1420.	144.
DAPHNE	368.2	250.	2.	33000	16.	1.9	863.	259.	-	-
ARES	509.	500.	1.75	118000	72.	~1.1	696.	1.35	989.	10.
VEPP2000	172.1	120.	0.23	8200	29.	0.95	246.0	0.4		<10.
DUKE-2	178.5	730	3.46	39000	77	3.16	-	-	-	-
KEK-PF	500.	785	3.45	39500	90.	1.4	791.	1000.	792.	5100.
ASP/Toshiba	500.	750	3.8	40400	75.	1.0	790.	25.	803.	8500.
BESSY	500.	735.	3.4	29600	80.	0.5	670.	11.	1072.	54.



Example on new R&D at SARI: BESSY + PEP-II

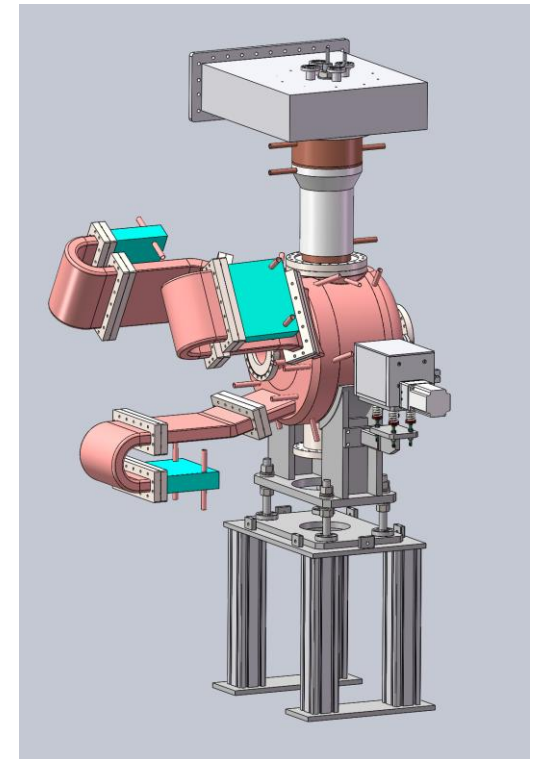
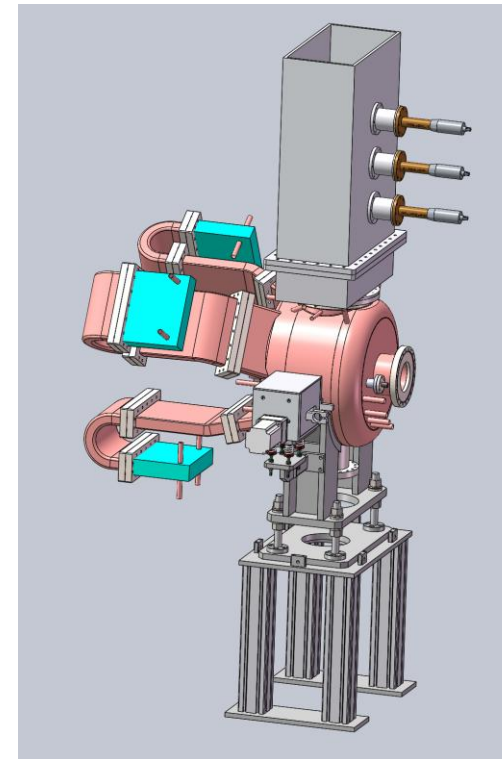
Bessy Cavity



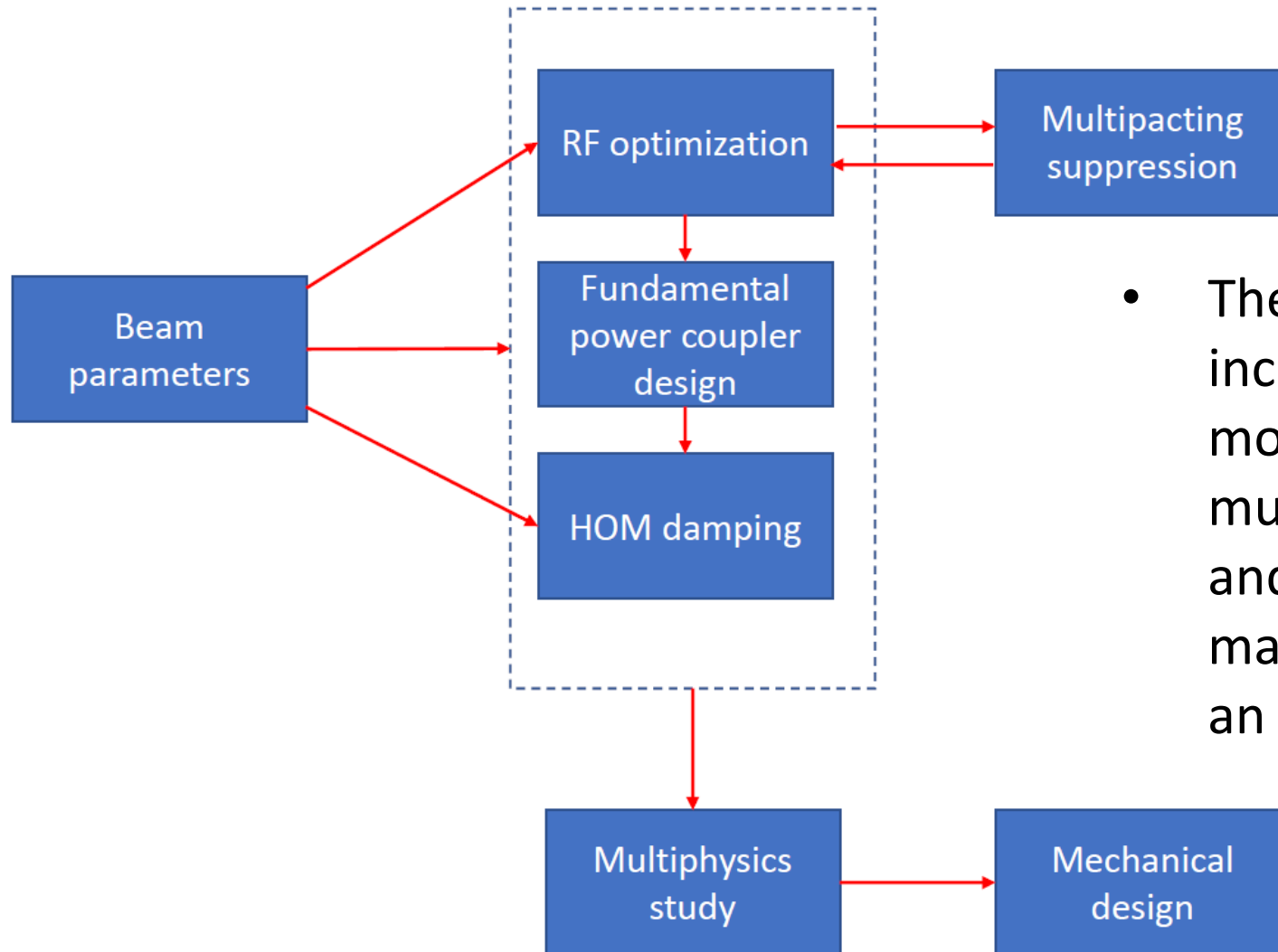
PEP-II RF cavity raft assembly

PEP-II HOMers

Bessy Cavity + PEP-II HOM, two types of coupler

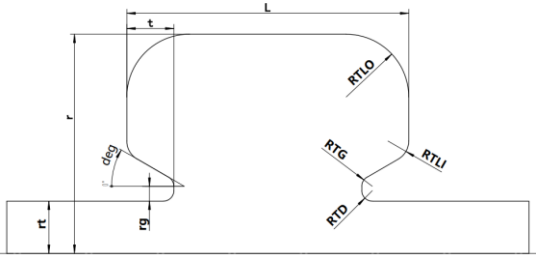


Design of the RF cavity



- The design of the RF cavity includes the fundamental mode optimization, the multipacting suppression, and the HOM-damping and many other progress, and is an iterative process.

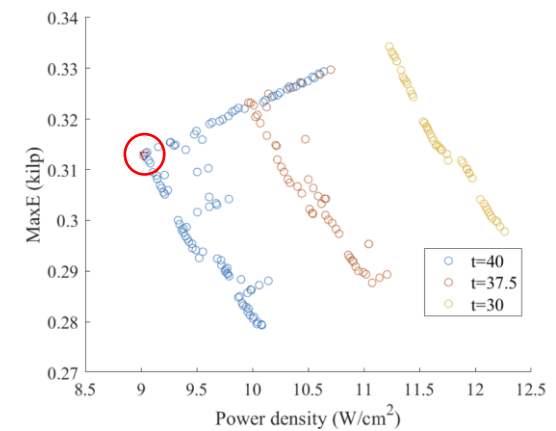
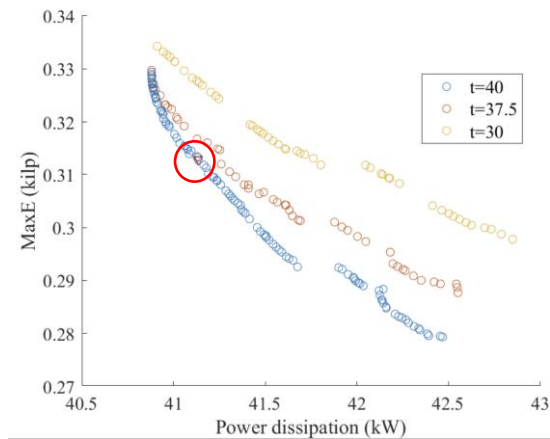
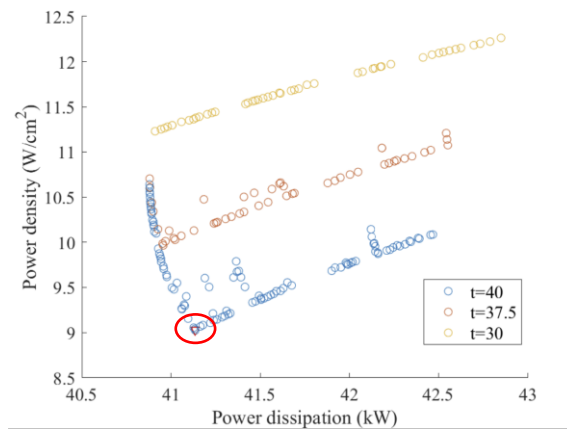
Design of the RF cavity — MOGA



Multi-objective evolutionary algorithm for the cavity optimization

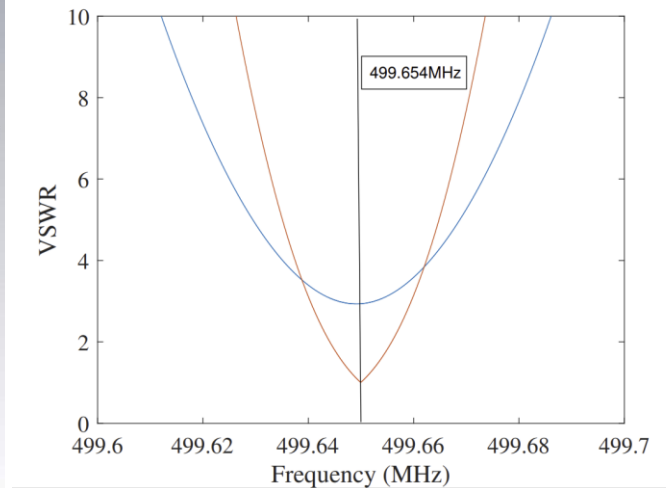
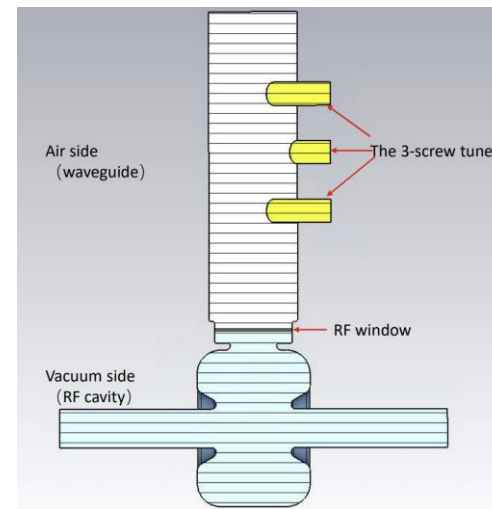
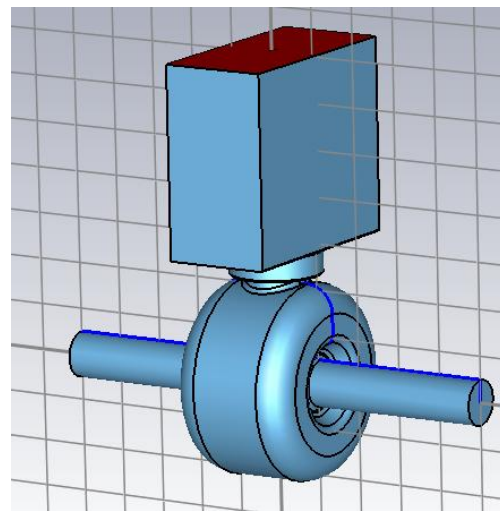
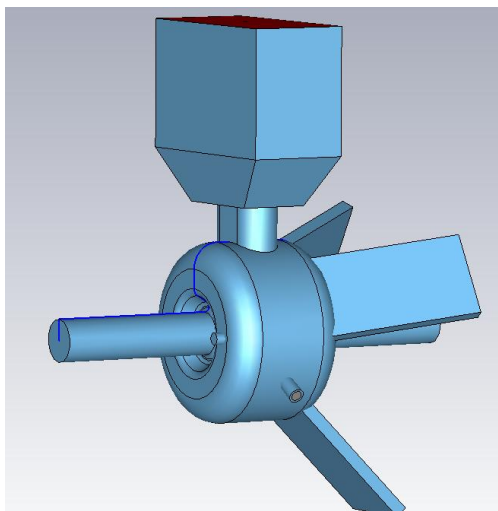
Objective	Constraint
With $V \cdot T = 600\text{kV}$	RF frequency 499.65~0.2MHz
Minimize P	Cavity radius < 30cm
Minimize Pden	Beam tube = 100mm
Minimize E	

	Optimized	Origin
Frequency	499.654MHz	499.654MHz
Voltage	600kV	
R/Q	114	96.79
Transit time factor	0.746	0.847
Shunt impedance	4.37M Ω	3.34M Ω
Power loss	41.19kW	53.55kW
Max Power density	9W/cm ²	16W/cm ²
MaxE	6.2MV/m	7.5MV/m

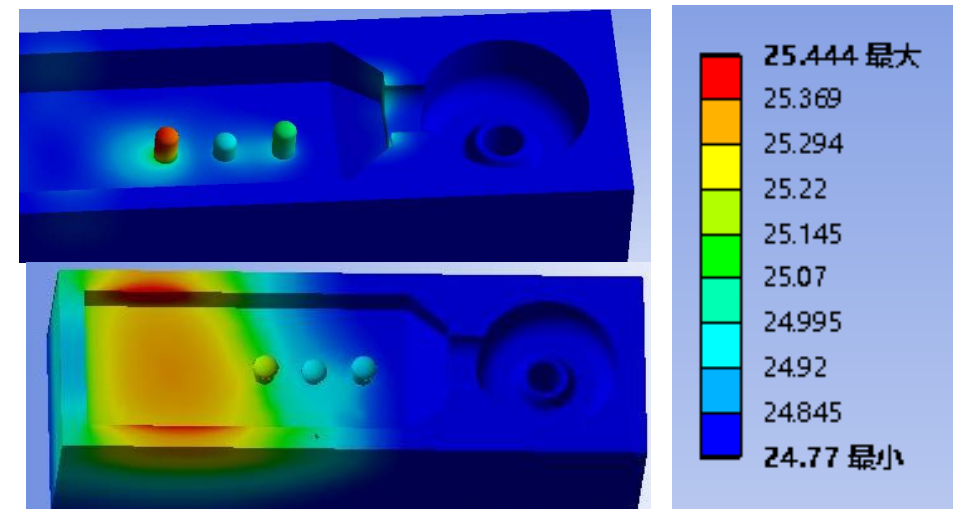


- Positive between power dissipation and power density
- Maximum electric field and power density are mutually exclusive.

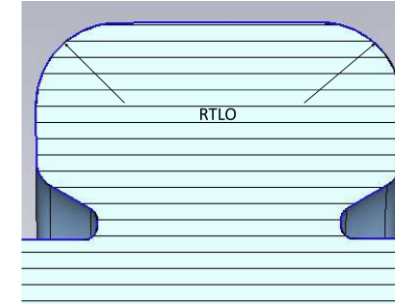
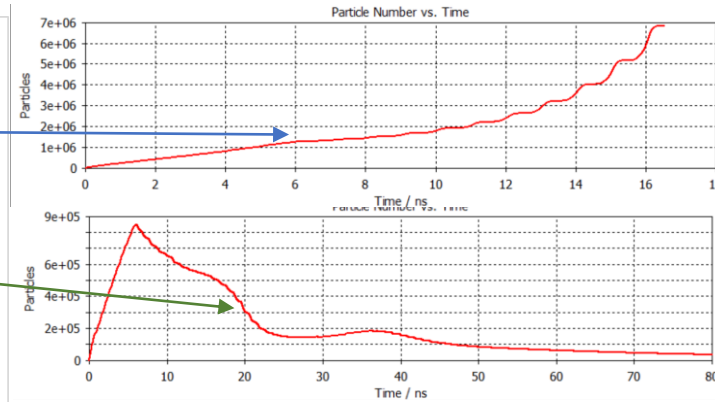
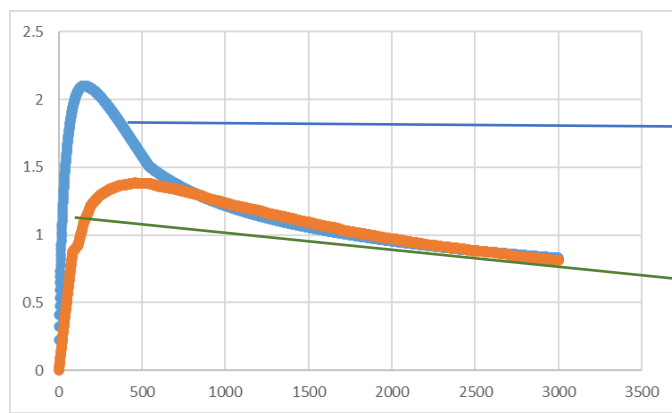
Design of the coupler



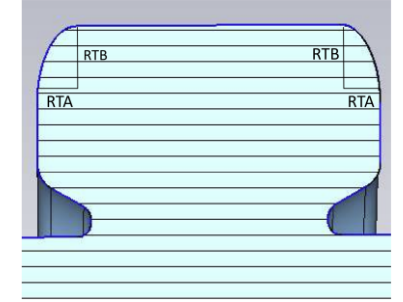
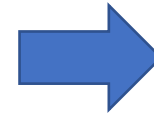
- Rectangular waveguide coupler, with circular ceramic RF window.
- Preserve the possibility of the coaxial coupler (iteration design)
- The function of adjustable coupling is realised by adopting a 3-screw tuner
- Simulation shows that the temperature rises little in the 3-screw tuner.



Multipacting

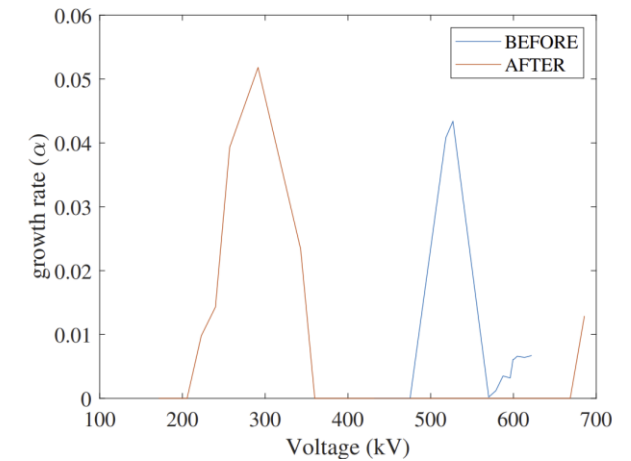
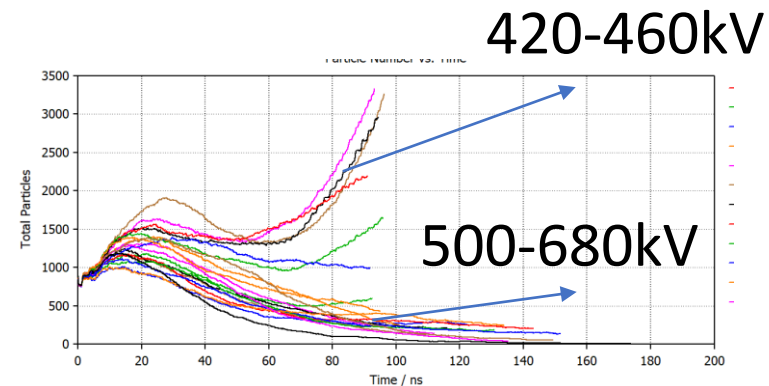
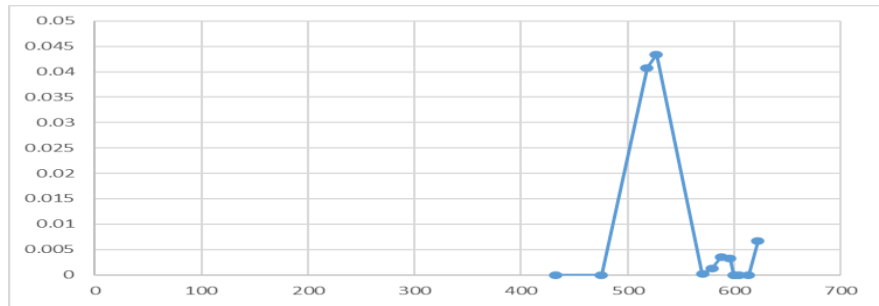


Round

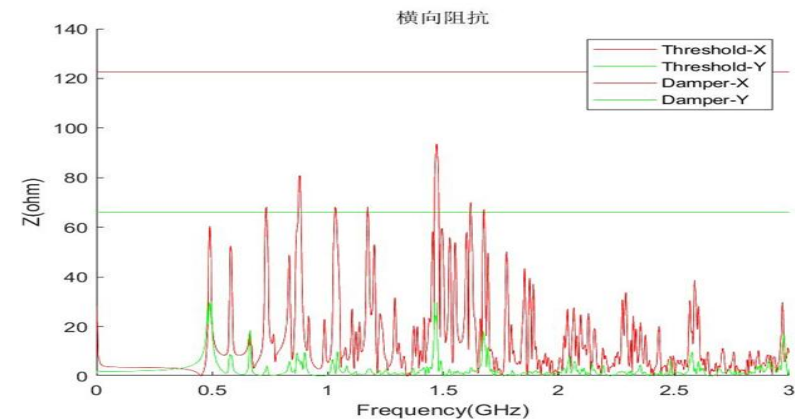
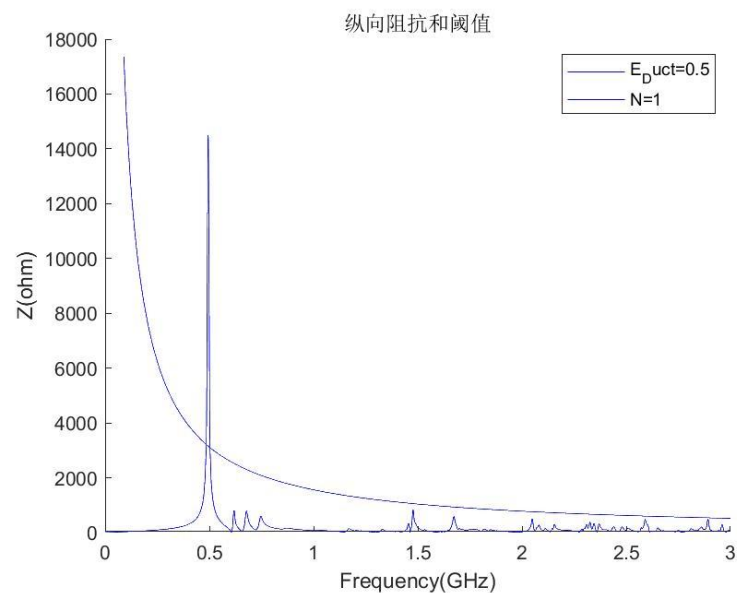


Ellipse

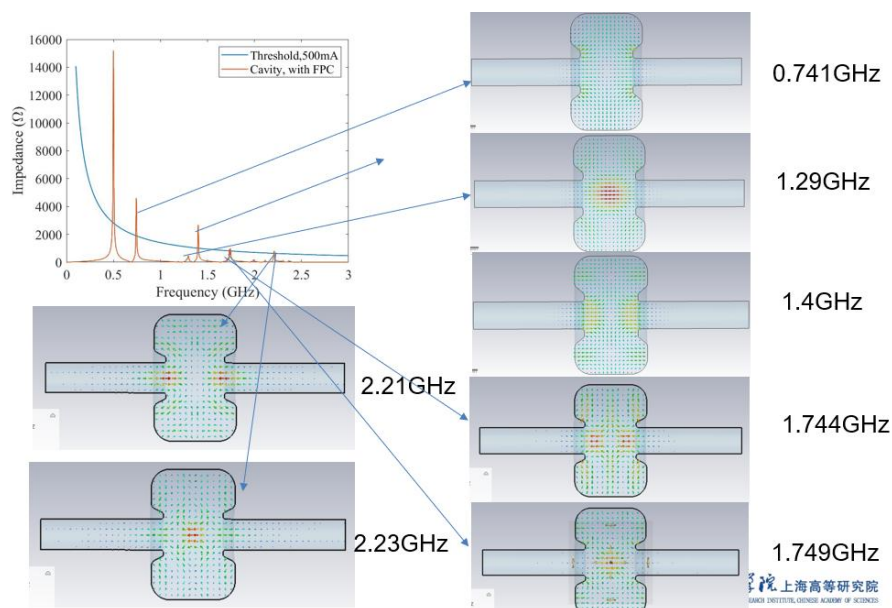
- Simulation of the multipacting in the CW cavity
- SEY in CST particle studio
- Modify the geometry of the cavity to move the multipact area



HOM-damping

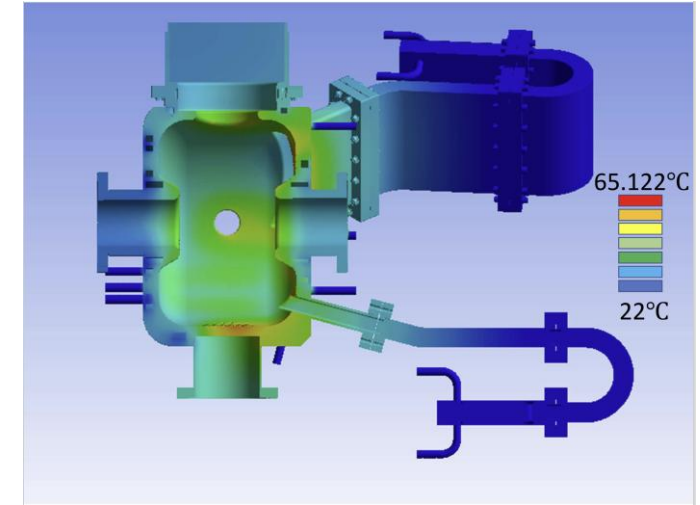
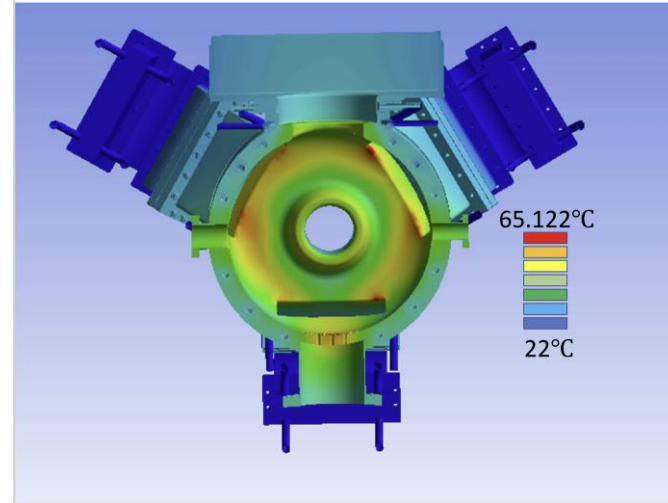
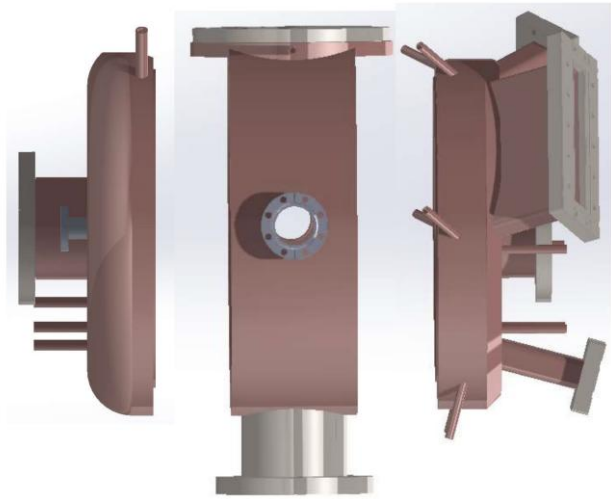


$$Z_{longitudinal}^{threshold} = \frac{1}{N f_{HOM}} \frac{2EQ_s}{eI\alpha\tau_L} \quad Z_{transverse}^{threshold} = \frac{1}{N f_{rev}} \frac{2E}{eI\beta_{trans}\tau_{trans}}$$



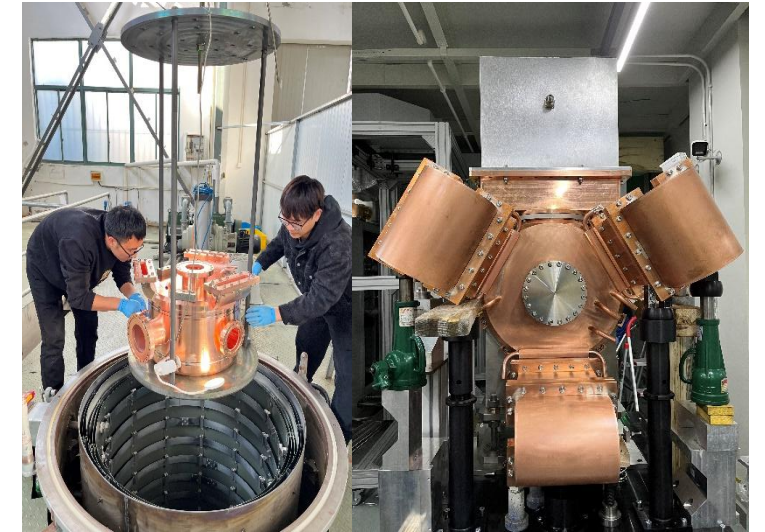
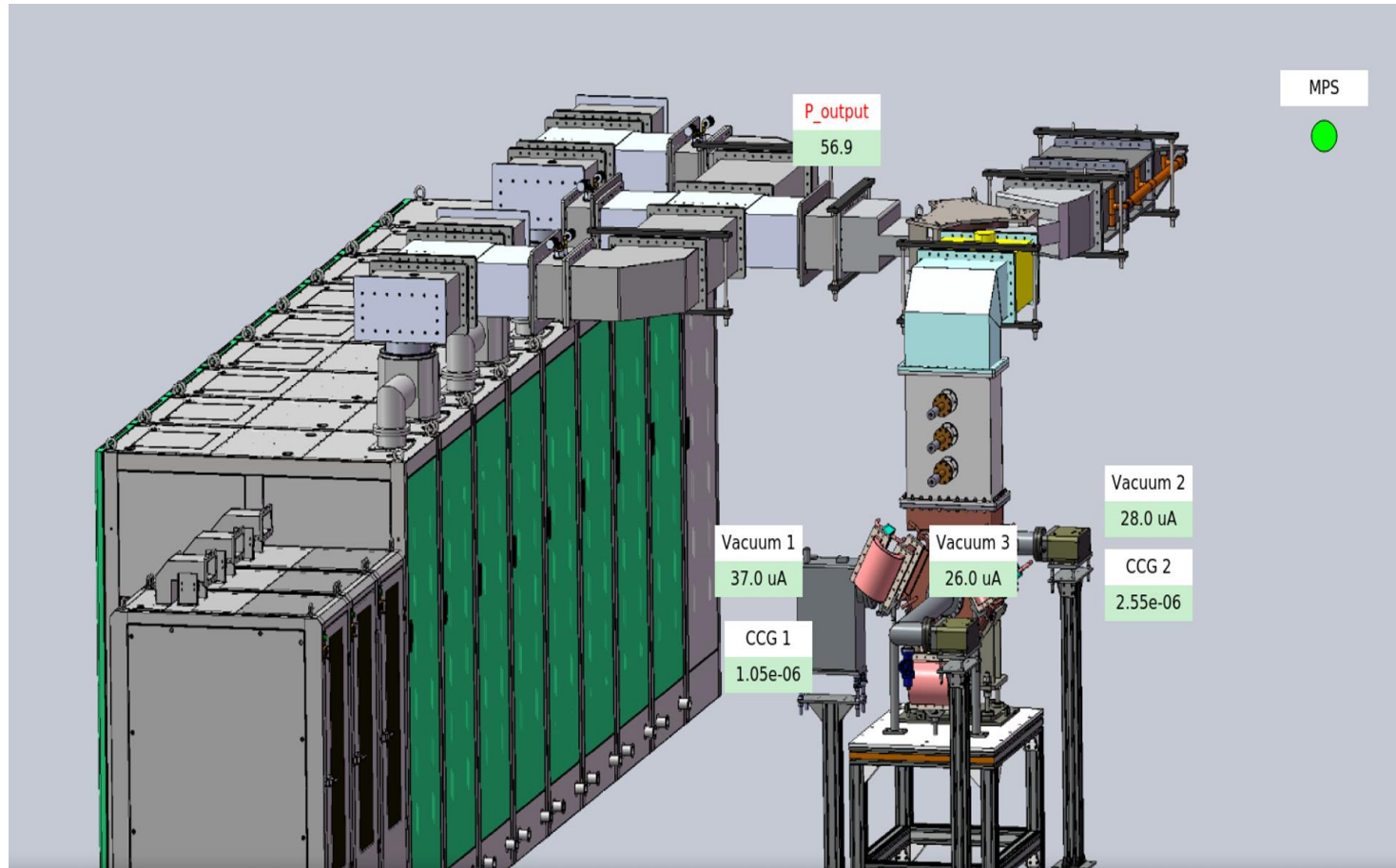
Mode frequency	R/Q	Q	R(MΩ)
754MHz	37.79	32500	0.613
1305MHz	9.14	63700	0.291
1420MHz	20.99	35700	0.375
1768MHz	5.51	41100	0.113
1821MHz	5.48	58600	0.161
2023MHz	3.21	89700	0.144
2127MHz	1.29	60600	0.0392

Multiphysics analyzation



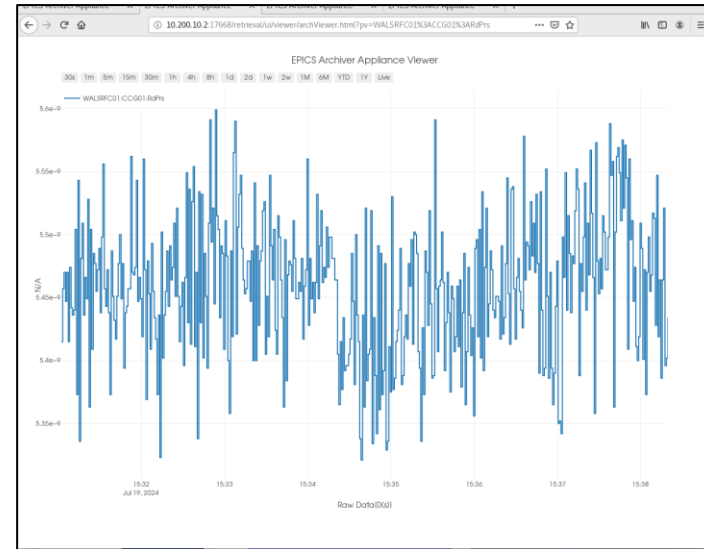
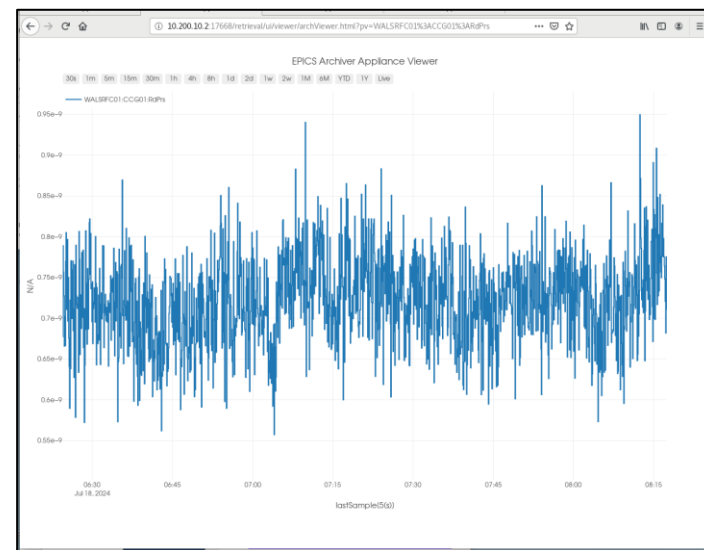
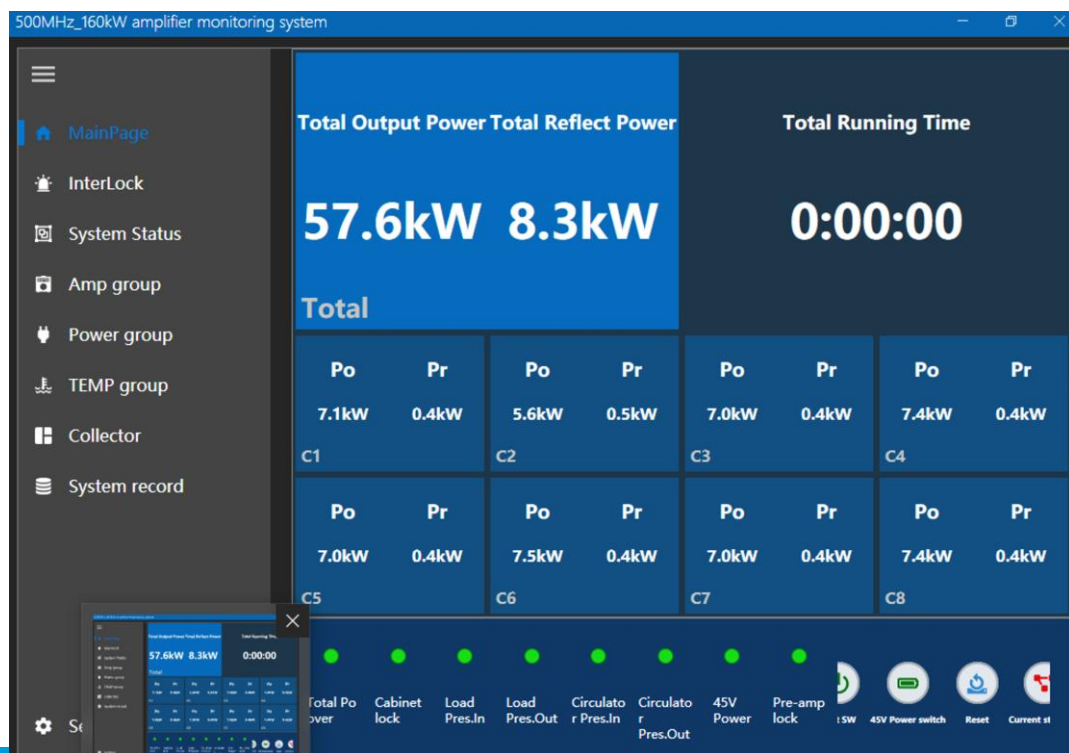
- Highest temperature rise: **44K**, at the corner of the HOM load
- Deformation caused by the temperature rise will result in the frequency shift by **282kHz**
- The frequency shift will be corrected by the tuner

Fabrication and high power test



The results of high power test

- ❑ Cavity voltage: 600 kV (only 40% duration, 20kW)
- ❑ RF stability: better than 0.1%, 0.1 degree (rms)
- ❑ Vaccum: 7×10^{-10} torr (static), 5×10^{-9} torr (power on)



New dual-mode cavity design for compact SR

- Compact light source needs both main cavity and 3rd harmonic cavity
- Dual-mode only needs one straight section

Superposition field of main RF field and harmonic RF field

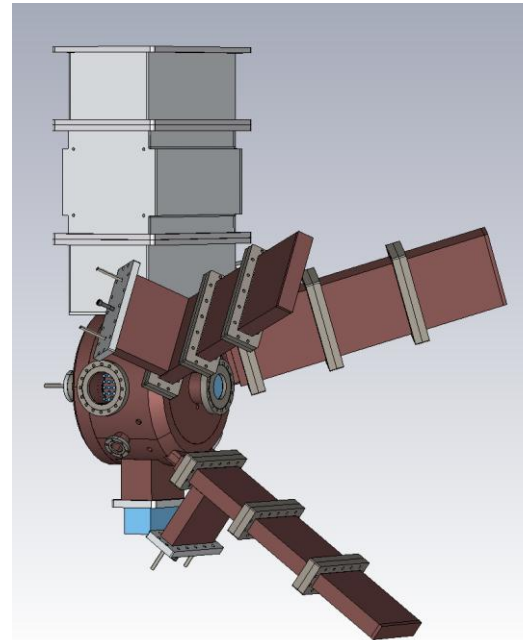
RF field seen by the bunch for main RF and higher harmonic cavities

$$E_z = E_1 \cos(kz) \cdot \sin(\omega t + \phi_0) + E_n \cos(nkz) \sin(n\omega t + \phi_n)$$

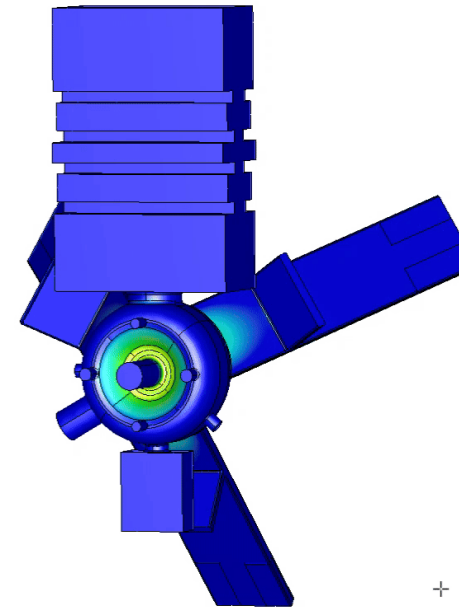
RF voltage seen by the bunch for main RF and higher harmonic cavities

$$V(\phi) = V_1 \cos(\omega t + \phi_s) + V_n \cos(n\omega t + n\phi_n)$$

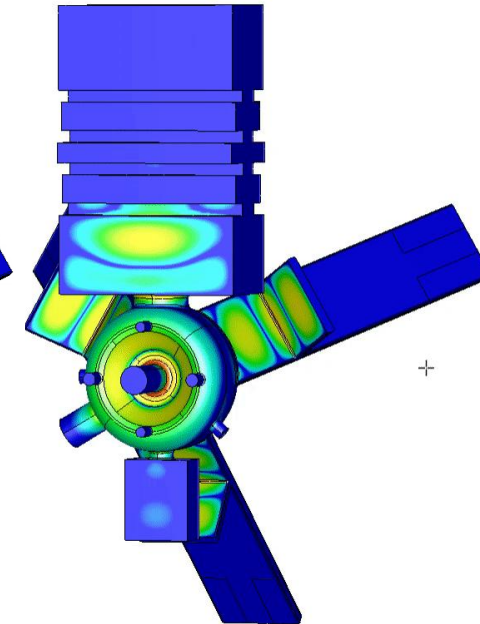
TM010 + TM020 cavity



TM010 500MHz



TM020 1.5GHz



Thanks for your attention!