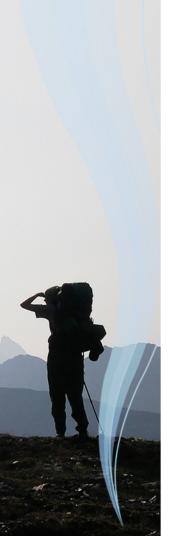
Foundation of Normal Conducting Accelerator

The 2nd International School on Beam Dynamics and Accelerator Technology (ISBA19)







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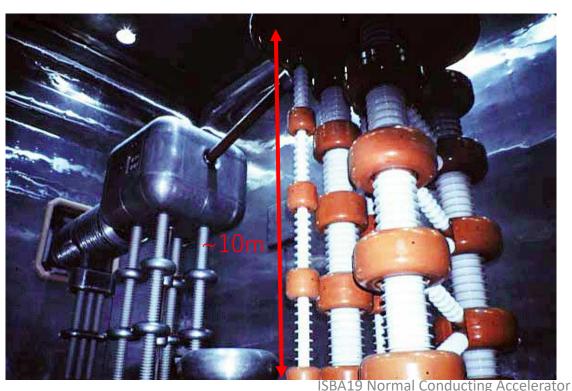
- The greatest Invention in Accelerator science, RF accelerator.
- Foundation of RF acceleration.
- Accelerator Structures.
- Beam loading control.

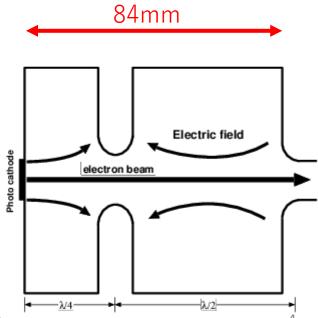




Static and RF fields

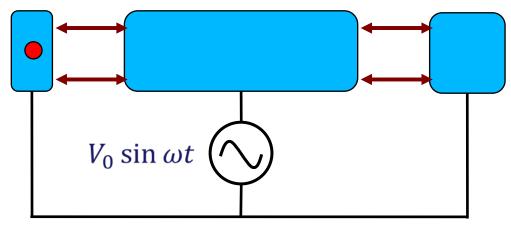
- Cockcrof Walton (KEK-PS Injector) : $700kV/15 \times 15 \times 15 \text{ (m}^3\text{)} = 2.1e+2 \text{ V/m}^3.$
- RF electron gun : $4MV/0.2x0.2x0.2 (m^3)=5.0e+8 V/m^3$.

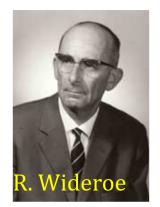




The greatest Invention

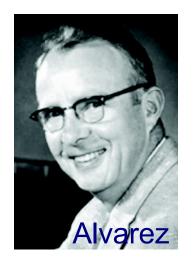
- R. Wideroe invented principle of AC acceleration (RF acceleration). This is the greatest invention in the accelerator science.
- By the repetitive acceleration employing temporally varied EM field, unlimited acceleration becomes possible.

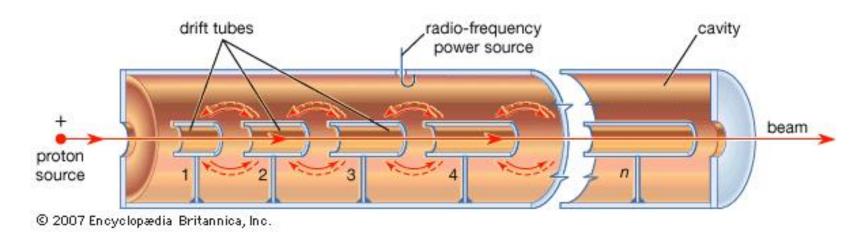




Alvarez-Linac: The first resonant cavity.

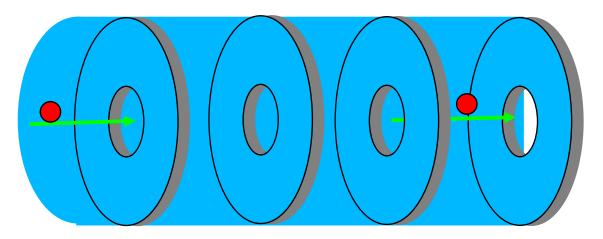
- Resonant structure surrounded by metal wall.
- No limitation on the acceleration frequency.
- Alvarez type, DTL(Drift Tube Linac) is still used in the hadron accelerator.





Disk-loaded Linac

- In a cylindrical wave guide, disks with a hole is placed.
- The particle passes through the hole.
- Phase velocity and group velocity are controlled by the disk shape and interval.

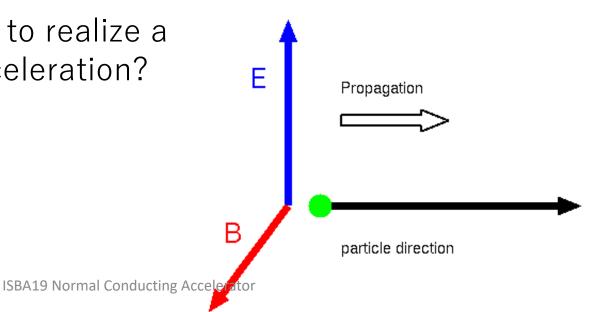




Foundation of RF Acceleration

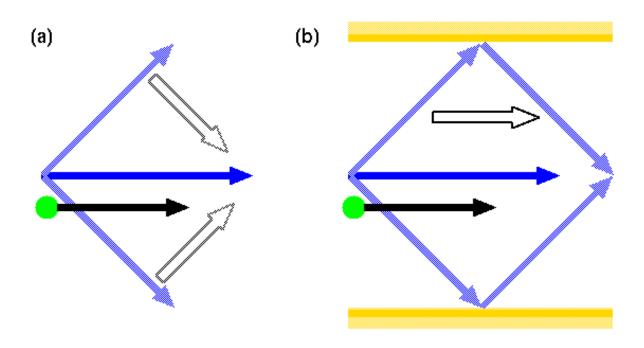
Question

- All of modern high energy accelerator is based on RF acceleration.
- Radio Frequency field is composed from plane wave.
- Plane wave propagation direction is perpendicular to Electric field.
- Question: How to realize a continuous acceleration?



Answer

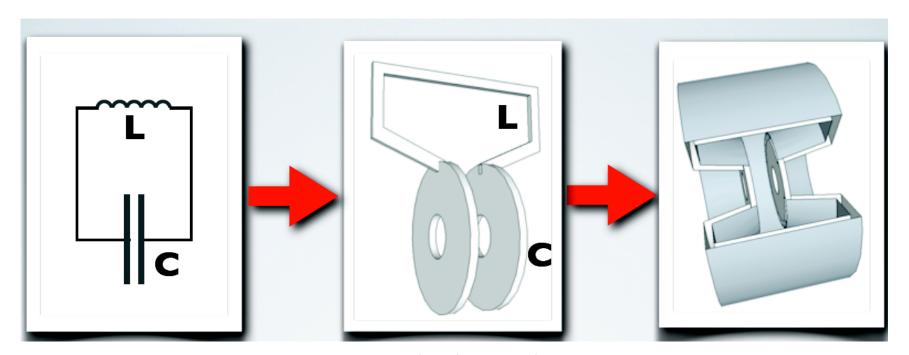
- Superposition of plane waves + reflection: E-field and propagation becomes same direction.
- Reflection is by metal wall -> cavity structure.



Pill Box Cavity

- The simplest accelerator structure.
- It is equivalent with a LC circuit.

- $\omega = \frac{1}{\sqrt{C}}$
- Frequency is determined by Geometry.



Starting from Maxwell Equation

$$\nabla \times \boldsymbol{H} = \boldsymbol{J} + \frac{\partial \boldsymbol{D}}{\partial t}$$
$$\nabla \times \boldsymbol{E} = -\frac{\partial \boldsymbol{D}}{\partial t}$$

$$\nabla \cdot \boldsymbol{D} = \rho$$

$$\nabla \cdot \boldsymbol{B} = 0$$

In vacuum,

$$\Delta \mathbf{E} - \frac{1}{c^2} \frac{\partial^2 \mathbf{E}}{\partial t^2} = 0,$$

$$\Delta \mathbf{B} - \frac{1}{c^2} \frac{\partial^2 \mathbf{E}}{\partial t^2} = 0,$$

with

$$\Delta = \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \frac{\partial^2}{\partial z^2}$$

$$\Delta = \frac{1}{r} \frac{\partial}{\partial r} \left(r \frac{\partial}{\partial r} \right) + \frac{1}{r^2} \frac{\partial^2}{\partial \theta^2} + \frac{\partial^2}{\partial z^2}$$

Helmholz Equation

Assume temporal oscillation,

$$\mathbf{E} = \mathbf{E}_{\mathbf{s}} e^{\imath \omega t}$$

$$\mathbf{B} = \mathbf{B}_{\mathbf{s}} e^{i\omega t}$$

Equation is simplified as

$$\Delta \boldsymbol{E}_{s} + k^{2} \boldsymbol{E}_{s} = 0,$$

$$\Delta \boldsymbol{B}_{s} + k^{2} \boldsymbol{B}_{s} = 0,$$

with

$$k^2 \equiv \frac{\omega^2}{c^2} = \varepsilon_0 \mu_0 \omega^2$$

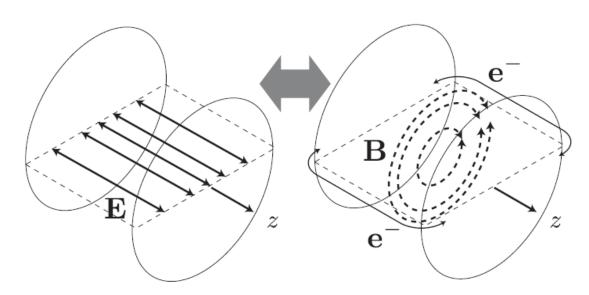
Solving Helmholz Equation

- Assuming TM mode
 (Only transverse B field)
- Other components are zero.

$$E_z = E_0 J_0(kr) \cos \omega t$$

$$H_{\phi} = -\frac{E_0}{Z_0} J_1(kr) \sin \omega t,$$

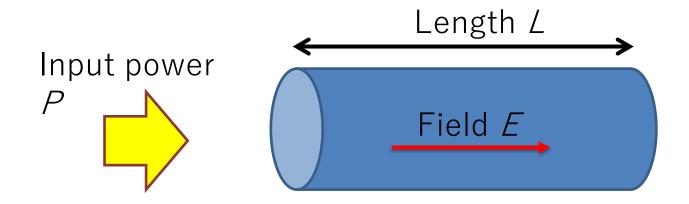
$$Z_0 \equiv \sqrt{\frac{\mu_0}{\varepsilon_0}} = 377\Omega, k = 2\pi/\lambda = p_{01}/b, p_{01}$$
 は Bessel 関数の根で 2.405



How much gain?

- Shunt impedance $r(\Omega/m)$ gives the gain,
 - *E*: field
 - P: dissipated power
 - L: length

$$r = \frac{(EL)^2}{PL} = \frac{E^2}{P/L}$$



Transit Time Factor

Time variation of RF field gives less acceleration than EL

$$V = \int_{L/2}^{L/2} E_0 \cos \frac{\omega s}{\lambda} ds = LE_0 \frac{\sin \frac{\omega L}{2\lambda}}{\frac{\omega L}{2\lambda}} = LE_0 T$$

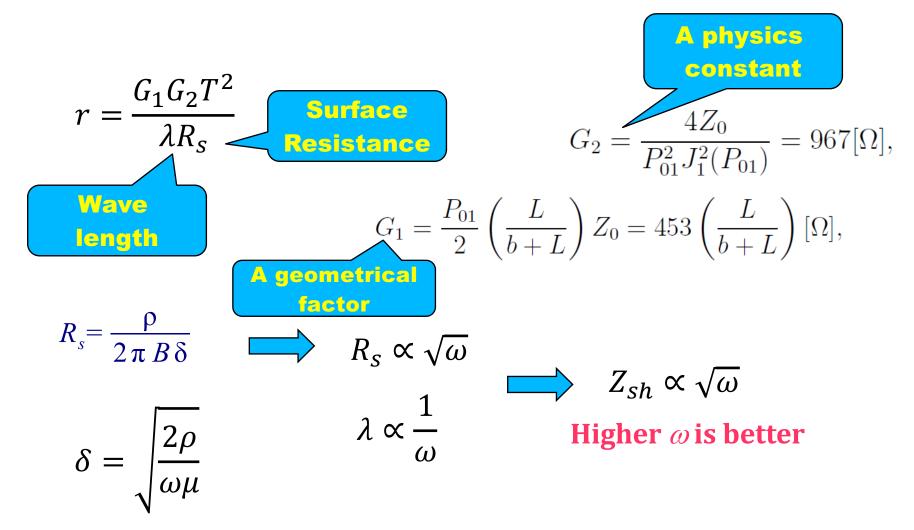
where T is Transit time factor

$$T \equiv \frac{\sin \frac{\omega L}{2\lambda}}{\frac{\omega L}{2\lambda}} < 1$$

The effective shunt impedance is defined with T as

$$r = \frac{(LE_0T)^2}{PL} = \frac{(E_0T)^2}{P/L}$$

Shunt Impedance of Pill Box Cavity



Q value

Q-value :numbers of oscillation to damp the stored energy W.

$$Q = \frac{\omega W}{P}$$

$$= \frac{\omega(=ck = cP_{01}/b)Z_0\varepsilon(=\sqrt{\mu\varepsilon} = 1/c)Z_0bL}{2R_s(b+L)}$$

$$= \frac{P_{01}Z_0L}{2R_s(b+L)}$$

$$= \frac{G_1}{R_s}$$
Surface
resistance

What is r/Q?

- r/Q is the ratio of the shunt impedance and Q value.
- The shunt impedance is product of r/Q and Q. So what???

$$\left(\frac{r}{Q}\right) = \frac{r}{Q}$$

$$r = Q\left(\frac{r}{Q}\right)$$

This is r/Q.

$$\frac{r}{Q} = \frac{V^2}{\omega WL} \qquad \text{(r/Q) is determined} \\ = \frac{2(E_0TL)^2}{\omega \pi \varepsilon b^2 L^2 E_0^2 J_1^2(P_{01})} \\ = \frac{2T^2 Z_0}{\omega (=cP_{01}/b)\pi (=P_{01}\lambda/(2b))Z_0\varepsilon (=1/c)b^2 J_1^2(P_{01})} \\ = \frac{4T^2 Z_0}{P_{01}^2 J_1^2(P_{01})\lambda} \\ = \frac{G_2T^2}{\lambda}, \qquad \text{Transit time} \\ \frac{G_2T^2}{\lambda}, \qquad \text{Transit time} \\ \frac{G_2T^2}{\lambda} \end{cases}$$

Shunt impedance again

• r is composed from two parts : (r/Q) and Q.

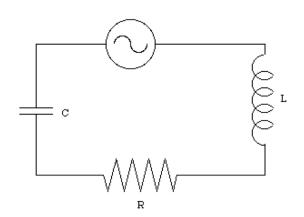
$$r = \left(\frac{r}{Q}\right)Q = \left(\frac{G_2T^2}{\lambda}\right)\left(\frac{G_1}{R_S}\right)$$
Genmetry
Material property

 r and Q values are measurable. The cavity shape and electrical quality can be evaluated with (r/Q) and Q values, respectively.

Equivalent Circuit : Pillbox

- Let us start with a simple L, C, R circuit.
- Impedance is

$$Z = R + \jmath(\omega L - \frac{1}{\omega C}),$$



It can be rewritten as

$$Z = R \left[1 + \jmath \sqrt{\frac{L}{C}} \frac{1}{R} \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right) \right], \qquad \omega_0 = \frac{1}{\sqrt{LC}}$$

Equivalent Circuit : Pillbox

We define Q as

$$Q = \sqrt{\frac{L}{C}} \frac{1}{R} = \frac{\omega_0 L}{R} = \frac{1}{\omega_0} CR,$$

With Q, Z is rewritten as

$$Z = R \left[1 + jQ \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right) \right],$$

$$Z \sim R \left(1 + \jmath Q \frac{2\delta f}{f_0} \right),$$

$$\delta f = \frac{\omega - \omega_0}{2\pi}$$

Equivalent Circuit: Pillbox

• On resonance $(\delta_f=0)$, the impedance is

$$Z = R \left(1 + jQ \frac{2\delta_f}{f_0} \right) = R$$

• Off resonance $(\delta_f = f_0/(2Q))$, the impedance is

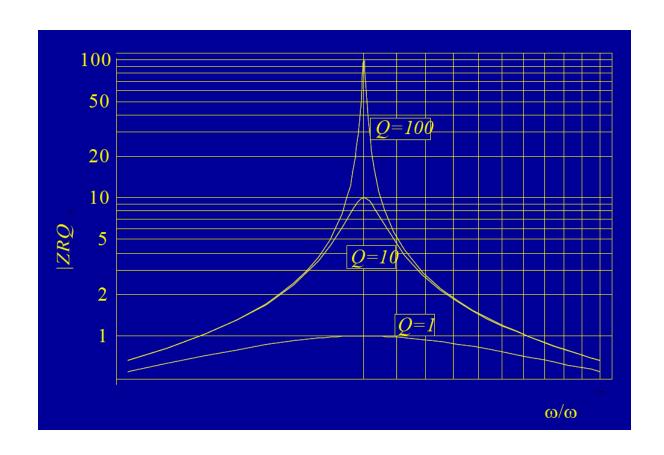
$$Z = R\left(1 + jQ\frac{2\delta_f}{f_0}\right) = R(1+j)$$

	Tuned	Detuned
Impedance	R	$\sqrt{2}R$
Field	E	$E/\sqrt{2}$
Stored Energy	W	W/2

Resonance Curve

Find δ_f where the resonance curve becomes ½ of the peak.

$$Q = \frac{f_0}{2\delta_f}$$



Other Definition of Q-value

$$W_C = \frac{1}{2}CV_C^2 = \frac{C}{2}\left(\frac{I}{j\omega_0 C}\right)^2 = \frac{1}{2}LI^2,$$

$$P = \frac{1}{2}RI^2,$$

$$Q = \omega_0 \frac{L}{R} = \omega_0 \frac{1/2LI^2}{1/2RI^2} = \omega_0 \frac{W}{P},$$

The last definition of Q-value

Q is a time constant of decay curve of W.

$$\frac{dW}{dt} = -P = -\frac{\omega_0}{Q}W$$
 because $Q = \omega_0 \frac{W}{P}$. It leads
$$W(t) = W_0 e^{-\frac{\omega_0}{Q}t}$$

Multi-cell cavity

The multi-cell cavity is LCR circuits with mutual inductance L_c

nth cell

$$\left(\frac{1}{\jmath\omega C} + R + \jmath\omega L\right)i_n + \jmath\omega L_c(i_n - i_{n+1}) + \jmath\omega L_c(i_n - i_{n-1}) = 0,$$

we assume a solution in a form of $i_n = i_0 e^{j(\omega t + n\phi)}$

which gives

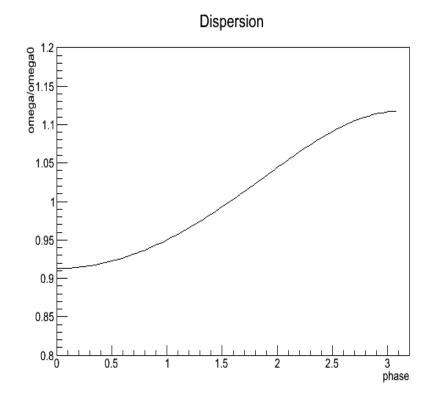
$$\left(\frac{1}{j\omega C} + R + j\omega L\right)i_n + j\omega 2L_c(1 - \cos\phi)i_n = 0,$$

Dispersion Relation

$$\frac{1}{j\omega C} + R + j\omega \left[L + 2L_c(1 - \cos\phi)\right] = 0,$$

$$\omega_0 = 1/\sqrt{C(L + 2L_c)}$$

$$\omega = \omega_0 \left[1 - \frac{2L_c}{L} \cos \phi \right]^{-1/2}$$



EM field in a cylinder

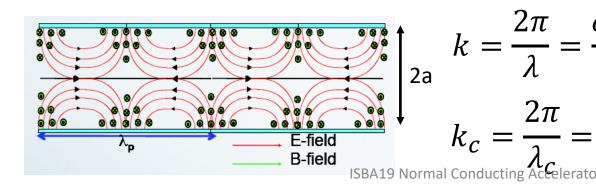


TM₀₁₀ solution

$$E_z = E_0 J_0(k_c r) e^{-jk_z z} e^{j\omega t}$$

$$E_r = j \frac{k_z}{k_c} E_0 J_1(k_c r) e^{-jk_z z} e^{j\omega t}$$

$$H_{\phi} = j \frac{k}{Z_0 k_c} E_0 J_1(k_c r) e^{-jk_z z} e^{j\omega t}$$



$$k = \frac{2\pi}{\lambda} = \frac{\omega}{c}$$

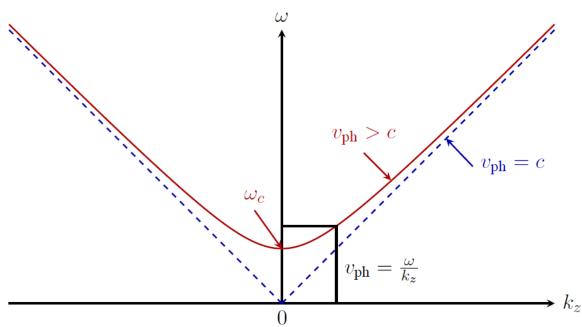
$$k_{za} = \frac{2\pi}{\lambda} = \frac{\omega}{c} \qquad k_{z}^{2} = k^{2} - k^{2}$$

$$k_{c} = \frac{2\pi}{\lambda_{c}} = \frac{\omega_{c}}{c} \qquad \lambda_{c} = 2.61a$$

$$k_z^2 = k^2 - k_c^2$$

$$\lambda_c = 2.61a$$

Dispersion Relation



$$\omega^2 = \omega_c^2 + c^2 k_z^2$$

$$v_{ph} = \frac{\omega}{k_z} = \sqrt{\frac{\omega^2}{k_z^2} + c^2} > c$$

 ν_{ph} is always faster than c.

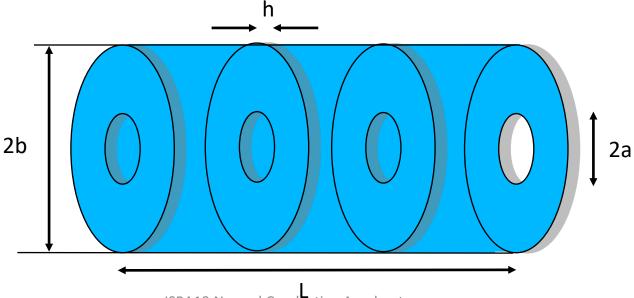
No physical particle can synchronize with the EM field in a cylinder.

ISBA19 Normal Conducting Accelerator

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Disk-loaded Structure

- Periodic boundary condition by disks.
- Dispersion relation is modified by the periodic condition.
- A part of forward wave is reflected by the disks and a backward wave is induced.

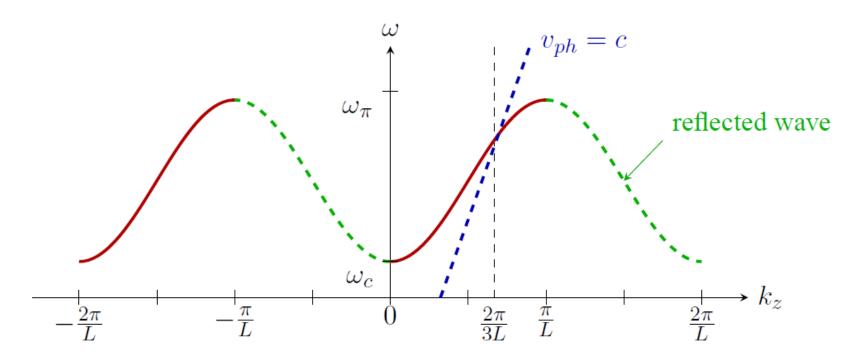


Dispersion Relation of Disk-loaded structure

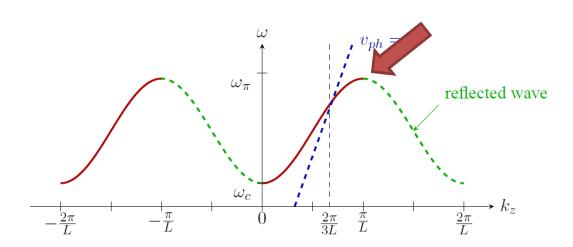
$$\omega = \frac{2.405c}{b} \sqrt{1 + \kappa (1 - \cos(k_z L)e^{-\alpha h})}$$

$$\kappa = \frac{4a^3}{3\pi J_1^2 (2.405)b^2 L} \ll 1$$

$$\alpha = \frac{2.405}{a}$$

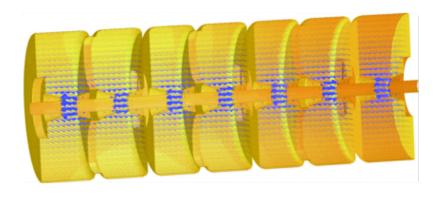


Standing Wave Linac

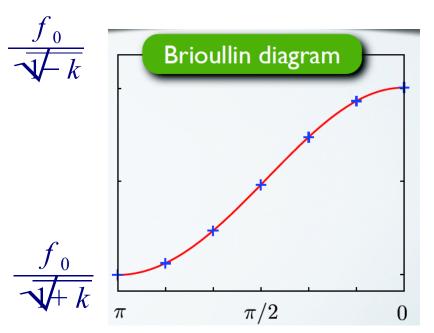


$$v_g = \frac{\partial \omega}{\partial k_z} = 0$$

Standing Wave Linac



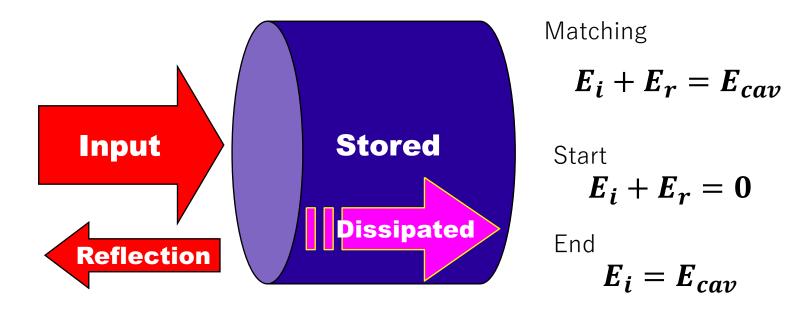
- Operated in Standing Wave.
- Phase advance per cell is



$$\phi_n = \frac{n\pi}{N}, n = 1, 2, \dots, N$$

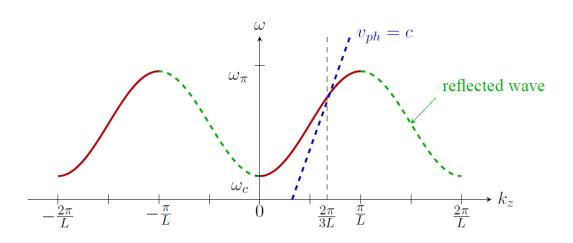
$$f_n = \frac{f_0}{\sqrt{k\cos(n\pi/N)}}$$

SW Cavity Power Balance

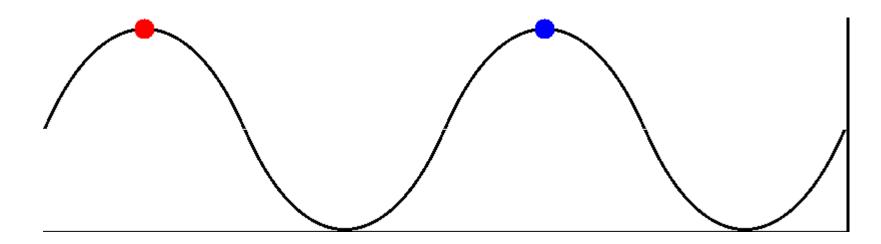


- The field should be matched at the coupling window.
- Starting from 100% reflection, the cavity power is grown up to 0% reflection.

Traveling Wave Cavity

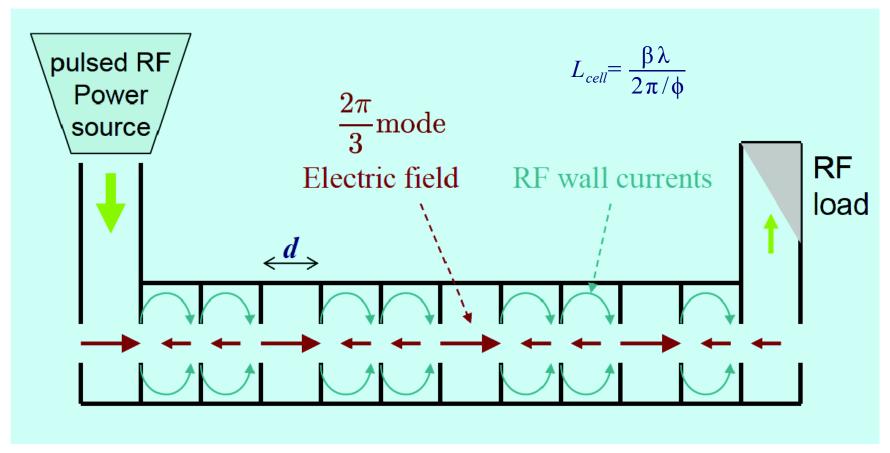


$$v_g = \frac{\partial \omega}{\partial k_z} \neq 0$$

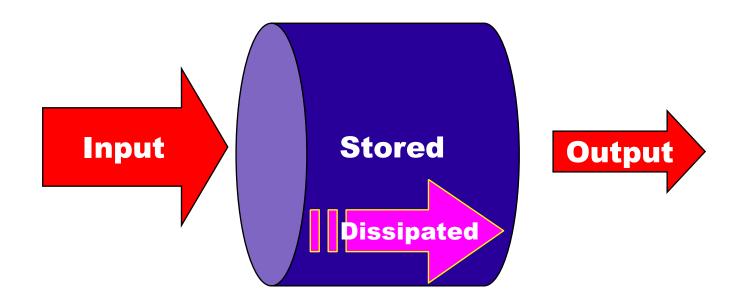


Traveling Wave Linac

EM power flows through accelerator



TW Cavity Power Balance



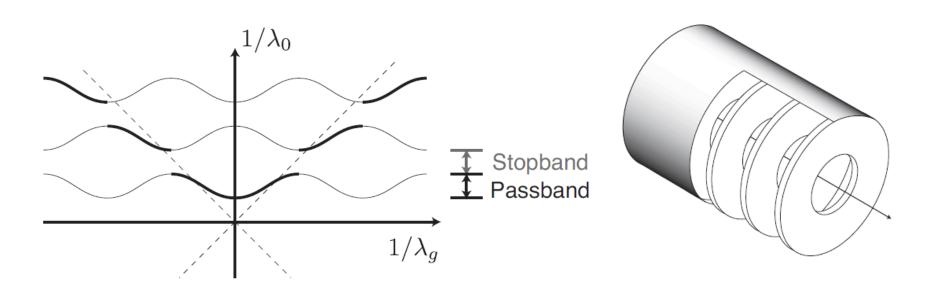
- Basically no reflection.
- Power flows through cavities.
- Rest of power is extracted and dumped.

Property of TW structure



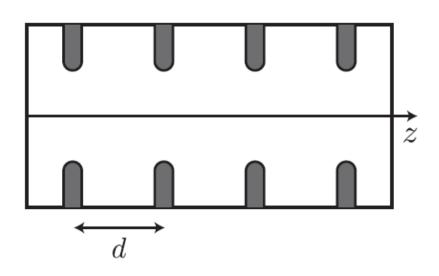
Disk-Loaded TW Linac

- By inserting disks periodically to the cylindrical wave guide, the phase velocity can be decreased.
- Synchronous frequency is determined from the dispersion relation.

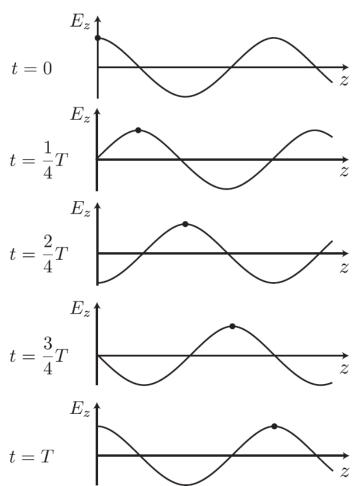


Disk-Loaded TW Linac

By operating it with the synchronous frequency, the particle sits always on a same phase along the structure.

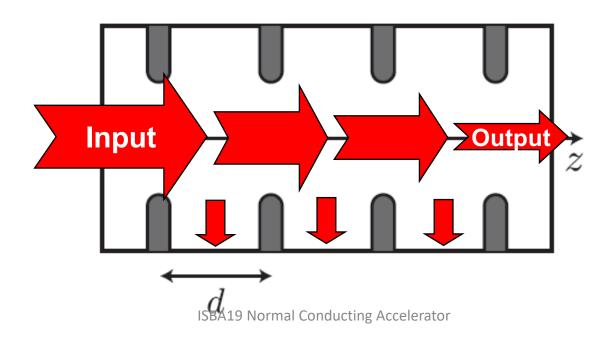


$$d = \frac{\varphi}{2\pi}\lambda$$



Disk-Loaded TW Linac

- The structure can be parametrized with a similar way to the pill box.
- The biggest difference from the SW cavity is that the power flows through the TW structure.



Parameters

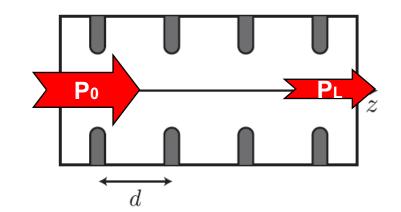
Shunt Impedance:
$$r=-\frac{E^2}{dP/dz}$$
 Q vaule: $Q=-\frac{\omega W}{dP/dz}$ $r/Q: Q=-\frac{E^2}{\omega W}$ Group velocity: $v_g=\frac{P}{W}=\frac{\omega P}{QdP/dz}$ Attenuation parameter $\alpha:\alpha=-\frac{1}{2P}\frac{dP}{dz}$ Filling time t_f : $t_f=\int_0^L\frac{dz}{v_g}$

Attenuation constant

$$\frac{dP}{dz} = -2\alpha P$$

P(z) is

$$P(z) = P_0 e^{\left(-2 \int_0^z \alpha(z') dz'\right)}$$



Output power

$$P(L) = P_0 e^{\left(-2 \int_0^L \alpha(z') dz'\right)} = P_0 e^{\left(-2\tau\right)}$$

With attenuation constant τ as

$$\tau = \int_0^L \alpha(z')dz'$$

Which can be measured as

$$\tau = -\frac{1}{2} ln \left(\frac{P_L}{P_0} \right)$$

Constant Impedance Structure

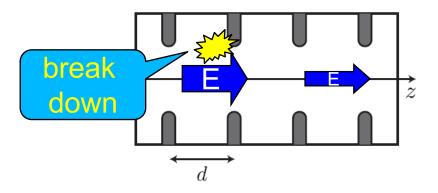
Constant v_g and r along the structure.

$$\tau = \alpha L$$

$$E(z) = E_0 e^{-\alpha z}$$

$$P(z) = P_0 e^{-2\alpha z}$$

Field is not constant along z. Because the maximum field is higher than the average, frequent discharge.



Constant Gradient

- In the CG structure, α is determined to give a constant field along the structure.
- α is adjusted by changing the iris size (aperture, a).
 - r depends not strongly on a.
 - dP(z)/dz should be constant.

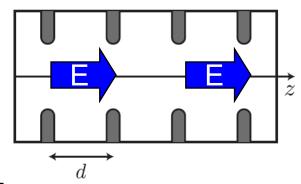
$$P(z) = P_0 - \frac{P_0 - P_L}{L} z = P_0 \left(1 - \frac{1 - e^{-2\tau t}}{L} z \right)$$
$$\frac{dP}{dz} = -\frac{P_0 - P_L}{L} = -P_0 \frac{1 - e^{-2\tau t}}{L}$$

The gradient of CG structure is a constant as expected.

$$E^2 = -r\frac{dP}{dz} = rP_0 \frac{1 - e^{-2\tau t}}{L}$$

 v_g and t_f are

$$v_g = \frac{\omega L - (1 - e^{-2\tau t})}{1 - e^{-2\tau t}} z$$
$$t_f = \frac{2Q\tau}{\omega}$$

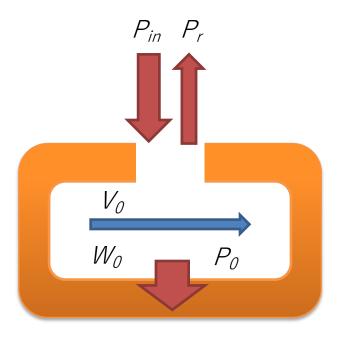


Because the maximum field is same as the average, discharge rate is much suppressed.

Transient Property and Beam Loading Compensation



Standing Wave Accelerator



Power balance

$$\frac{dW_0}{dt} = P_{in} - P_r - P_0$$

Express in V_0

$$\frac{dW_0}{dt} = \frac{Q_0}{\omega} \frac{dP_0}{dt} = \frac{Q_0}{\omega} 2G_0 V_0 \frac{dV_0}{dt}$$

Admittance G_0

$$P_0 = G_0 V_0^2$$

With the coupling beta β , admittance of wage guide is

$$G_{wg} = \beta G_0$$

$$P_{in} = \beta G_0 V_{in}^2 \qquad P_r = \beta G_0 V_r^2$$

Matching condition at the coupling window

$$V_{in} + V_r = V_0$$

$$\frac{Q_0}{\omega} 2G_0 V_0 \frac{dV_0}{dt} = P_{in} - P_r - P_0 = -\beta G_0 V_0^2 + 2\beta G_0 V_0 V_{in} - G_0 V_0^2$$

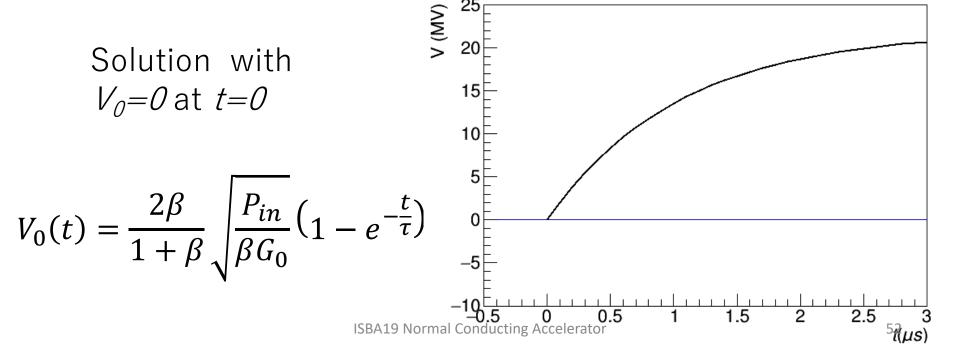
The equation can be summarized with $Q_0 = (1 + \beta)Q_L$

$$\frac{2Q_L}{\omega}\frac{dV_0}{dt} = -V_0 + \frac{2\beta}{1+\beta}V_{in}$$

Differential equation of SW accelerator

$$\tau \frac{dV_0}{dt} = -V_0 + \alpha V_{in}$$

$$\alpha \equiv \frac{2\beta}{1+\beta} \qquad \tau \equiv \frac{2Q_L}{\omega}$$



Introduce Beam

- Cavity voltage is determined by not only Input power, but also beam current.
- Beam consume the cavity stored energy as

$$P_{beam} = \frac{1}{T} \int_0^T dt \, IV_0 \cos(\omega t)$$

If the beam is delta function on crest,

$$P_{beam} = IV_0$$

Power balance is

$$\frac{dW_0}{dt} = P_{in} - P_r - P_0 - IV_0$$

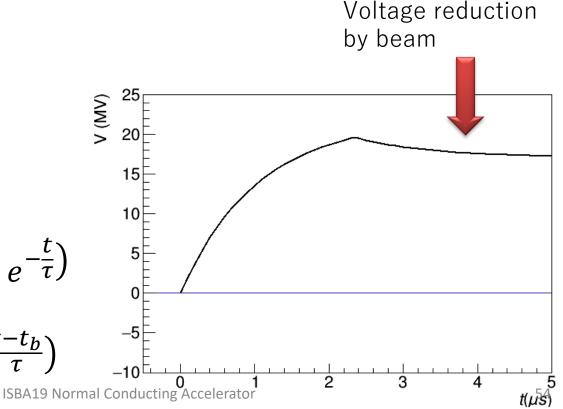


Differential equation of SW accelerator with Beam

$$\tau \frac{dV_0}{dt} = -V_0 + \alpha V_{in} - \frac{I}{(1+\beta)G_0}$$

Solution with $V_0=0$ at t=0, Beam start at $t=t_h$

$$V_{0}(t) = \frac{2\beta}{1+\beta} \sqrt{\frac{P_{in}}{\beta G_{0}}} \left(1 - e^{-\frac{t}{\tau}}\right)$$
$$-\frac{I}{(1+\beta)G_{0}} \left(1 - e^{-\frac{t-t_{b}}{\tau}}\right)$$

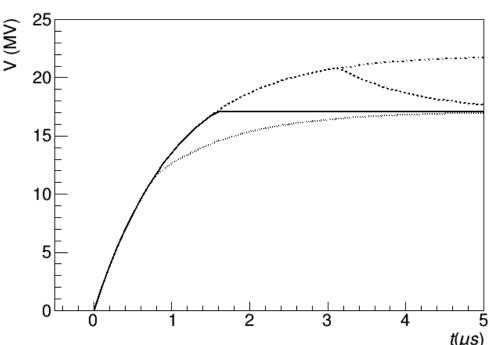


Beam Loading Compensation

Since RF and Beam term have the same time constant τ , V_0 can be constant by adjusting amplitude and timing.

$$V_{0}(t) = \frac{2\beta}{1+\beta} \sqrt{\frac{P_{in}}{\beta G_{0}}} \left(1 - e^{-\frac{t}{\tau}}\right) - \frac{I}{(1+\beta)G_{0}} \left(1 - e^{-\frac{t-t_{b}}{\tau}}\right)$$

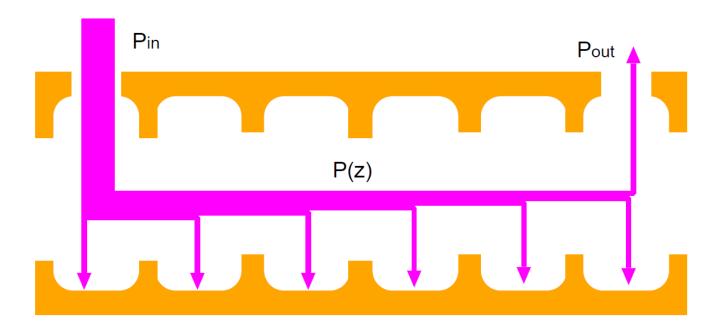
$$t_b = -\tau ln \left(\frac{I}{2} \sqrt{\frac{rL}{\beta P_{in}}} \right)$$



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- Accelerator with a finite group velocity.
- The power flows through the accelerator.
- The beam loading can be suppressed with an active method.





Power Balance

Assume quasi-constant power flow in TW accelerator.

$$\frac{dP}{dz} = \left(\frac{dP}{dz}\right)_{wall} + \left(\frac{dP}{dz}\right)_{beam} = -2\alpha(z)P(z,t) - IE(z,t)$$

Expansion with partial derivative

$$\frac{dP}{dz} = \frac{\partial P}{\partial z} + \frac{\partial P}{\partial t} \frac{dt}{dz} = \frac{\partial P}{\partial z} + \frac{1}{v_g} \frac{\partial P}{\partial t}$$

$$\frac{\partial P}{\partial z} + \frac{1}{v_g} \frac{\partial P}{\partial t} - 2\alpha(z)P(z,t) - IE(z,t) = 0$$

Expression in field,

$$\frac{\partial E}{\partial z} - \frac{E}{2\alpha} \frac{d\alpha}{dz} + \frac{1}{v_g} \frac{\partial E}{\partial t} + \alpha(z)E(z,t) + I\alpha r = 0$$

$$P = \frac{E^2}{2\alpha r}$$

r. shunt impedance

Constant Gradient condition $\frac{d\alpha}{dz} = -2\alpha$

$$\frac{\partial E(z,t)}{\partial z} + \frac{1}{v_g} \frac{\partial E(z,t)}{\partial t} + I(t)\alpha r = 0$$

Laplace Transformation

$$\mathcal{L}[f(t)] \equiv F(s) = \int_0^\infty dt f(t)e^{-st}$$

In s-domain, differential and integral operation becomes algebraic operation.

$$\mathcal{L}\left[\frac{df}{dt}\right] = sF(s) - f(0)$$

$$\mathcal{L}\left[\int_{0}^{t} f(u)du\right] = \frac{1}{s}F(s)$$

Differential Equation f(t)

Laplace transformation

Differential Equation F(s)

Algebra

Solution f(t)

Inverse Laplace transformation

Solution F(s)

$$\frac{\partial E(z,t)}{\partial z} + \frac{1}{v_g} \frac{\partial E(z,t)}{\partial t} + I(t)\alpha r = 0$$

Laplace transformation

$$\frac{\partial E(z,s)}{\partial z} + \frac{s}{v_g} \frac{\partial E(z,s)}{\partial t} + I(s)\alpha r = 0$$

By integrating with z,

$$E(z,s) - E(0,s)e^{st_z} + I(s)r \int_0^z \alpha dz = 0$$

By integrating with z again,

$$V(s) = \frac{\omega L}{Q(1 - e^{-2\tau})} \frac{1}{s + \frac{\omega}{Q}} E(s) \left[1 - e^{-\left(s + \frac{\omega}{Q}\right)t_f} \right]$$

$$-\frac{\omega r L I(s)}{2Q(1-e^{-2\tau})s} \left[1 - e^{-\frac{\omega}{Q}t_f} - \frac{\omega (1-e^{-st_f-2\tau})}{Q(s+\omega/Q)} \right]$$

Input RF E(t) and Beam current,

$$E(t) = E_0 u(t), I(t) = I_0 u(t - t_f)$$

Giving

$$E(s) = \frac{E_0}{s}, \qquad I(s) = \frac{I_0}{s} e^{-st_f}$$

$$V(s) = \frac{\omega L}{Q(1 - e^{-2\tau})} \frac{1}{s + \frac{\omega}{Q}} \frac{E_0}{s} \left[1 - e^{-\left(s + \frac{\omega}{Q}\right)t_f} \right]$$

$$-\frac{\omega r L I(s)}{2O(1 - e^{-2\tau})} \frac{I_0 e^{-st_f}}{s^2} \left[1 - e^{-\frac{\omega}{Q}t_f} - \frac{\omega \left(1 - e^{-st_f - 2\tau}\right)}{O(s + \omega/Q)} \right]$$

In t-domain,

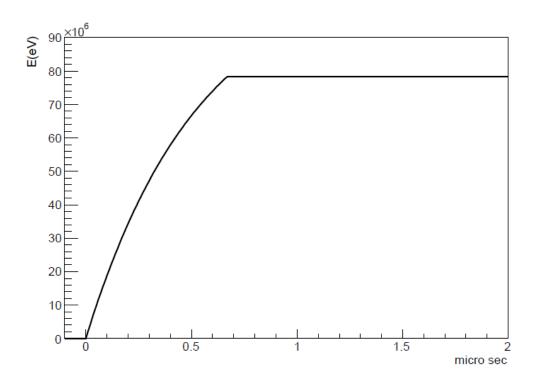
$$\begin{split} V(t) &= \frac{LE_0}{1 - e^{-2\tau}} \bigg[\Big(1 - e^{-\frac{\omega}{Q}t} \Big) u(t) - e^{-2\tau} \Big(1 - e^{-\frac{\omega}{Q}(t - t_f)} \Big) u(t - t_f) \bigg] \\ &- \frac{rLI_0}{2} \bigg[-\frac{\omega}{Q} \frac{e^{-2\tau}}{1 - e^{-2\tau}} \Big(t - t_f \Big) + \frac{1 - e^{-\frac{\omega}{Q}(t - t_f)}}{1 - e^{-2\tau}} \bigg] u(t - t_f) \end{split}$$

$$0 < t < t_f$$

$$V(t) = \frac{LE_0}{1 - e^{-2\tau}} \left(1 - e^{-\frac{\omega}{Q}t} \right)$$

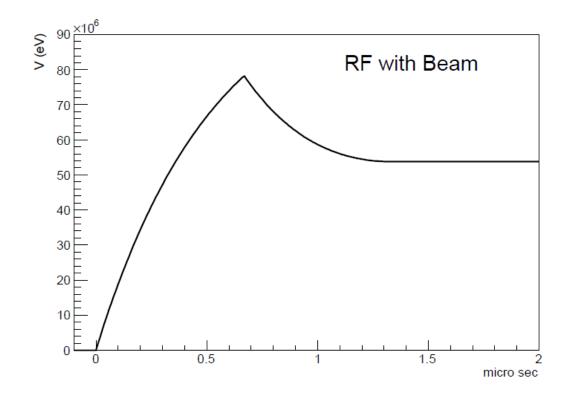
 $t > t_f$ and $I_0 = 0$,

$$V(t) = \frac{LE_0}{1 - e^{-2\tau}} \left(1 - e^{-\frac{\omega}{Q}t_f} \right) = LE_0$$



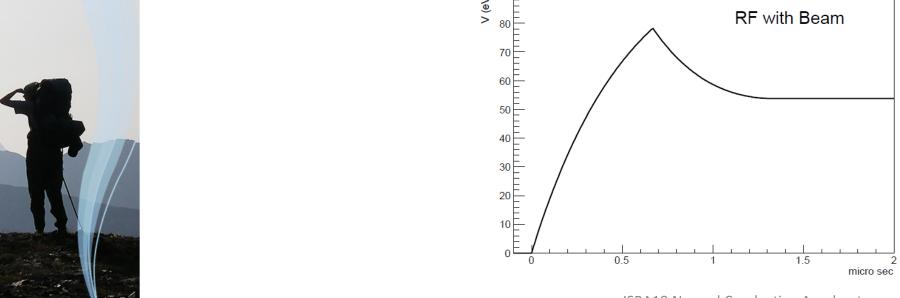
$$t > t_f \ and \ I_0 \neq 0,$$

$$V(t) = LE_0 - \frac{rLI_0}{2} \left[-\frac{\omega}{Q} \frac{e^{-2\tau}}{1 - e^{-2\tau}} (t - t_f) + \frac{1 - e^{-\frac{\omega}{Q}(t - t_f)}}{1 - e^{-2\tau}} \right],$$



Beam Loading Compensation

- Acceleration field is varied when we start the beam acceleration.
- We consider its compensation for uniform acceleration.



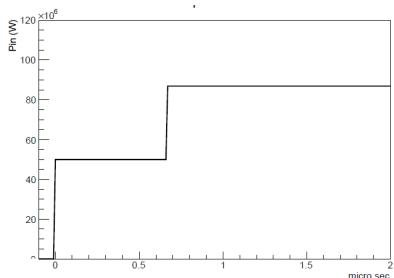


Amplitude Modulation

Amplitude modulation on E(t) for compensation,

$$E(t) = E_0 u(t) + E_1 u(t - t_f)$$

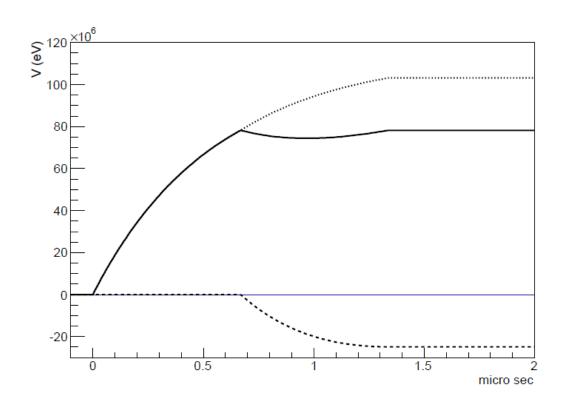
$$E(s) = \frac{E_0}{s} + \frac{E_1}{s} e^{-st_f}$$



$$V(t) = E_0 L + \frac{LE_1}{1 - e^{-2\tau}} \left(1 - e^{-\frac{\omega}{Q}(t - t_f)} \right)$$

$$-\frac{r_0 L I_0}{2(1 - e^{-2\tau})} \left[-\frac{\omega}{Q} e^{-2\tau} (t - t_f) + 1 - e^{-\frac{\omega}{Q}(t - t_f)} \right],$$

$$E_1 = \frac{r_0 I_0}{2} \left(\frac{2\tau e^{-2\tau}}{1 - e^{-2\tau}} - 1 \right),$$



The variation is much compensated, but not perfect.

Two components AM

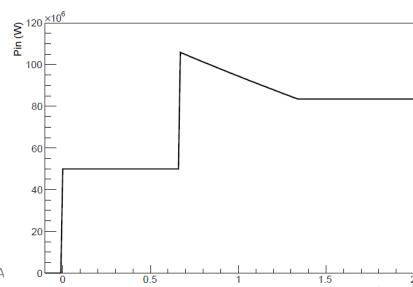
$$V(t) = E_0 L - \frac{r_0 L I_0}{2(1 - e^{-2\tau})} \left[-\frac{\omega}{Q} e^{-2\tau} (t - t_f) + 1 - e^{2\tau - \frac{\omega}{Q}t} \right].$$

The beam loading term has two components. AM should have also two components for a perfect compensation.

$$E(t) = E_0 u(t) + E_1 u(t - t_f)$$

$$+ E_2 (t - t_f) u(t - t_f)$$

$$- E_2 (t - t_f) u(t - 2t_f)$$



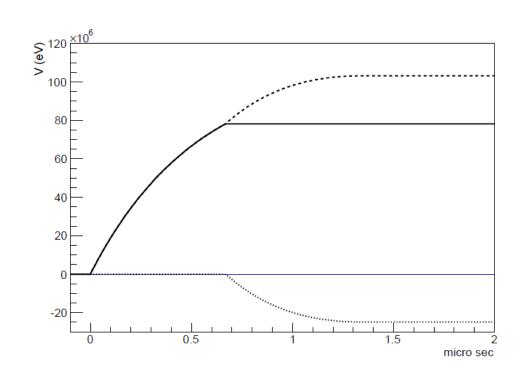
$$V(t) = E_0 L + \frac{L}{1 - e^{-2\tau}} \left(E_1 - \frac{Q}{\omega} E_2 \right) \left(1 - e^{-\frac{\omega}{Q}(t - t_f)} \right) + \frac{L e^{-2\tau}}{1 - e^{-2\tau}} E_2(t - t_f)$$

$$- \frac{r_0 L I_0}{2(1 - e^{-2\tau})} \left[-\frac{\omega}{Q} e^{-2\tau} (t - t_f) + 1 - e^{-\frac{\omega}{Q}(t - t_f)} \right],$$

$$V(t) = E_0 L + \frac{L}{1 - e^{-2\tau}} \left(E_1 - \frac{Q}{\omega} E_2 \right) - \frac{r_0 L I_0}{2(1 - e^{-2\tau})} = E_0 L,$$

$$E_1 = \frac{r_0 I_0}{2} (1 - e^{-2\tau}),$$

$$E_2 = -\frac{r_0 I_0}{2} \frac{\omega}{Q} e^{-2\tau},$$



Summary

- RF acceleration is a great invention realizing high energy acceleration beyond the vacuum discharge.
- For acceleration, appropriate boundary condition and synchronization are essential.
- Two types of accelerator: Standing wave and Travelling wave.
- Beam loading compensation is mandatory for multi-bunch acceleration.

Stored Energy

$$W = \frac{\varepsilon_0}{2} \int_V E_z^2 dV$$

$$= \frac{\varepsilon_0}{2} \int_V [E_0 J_0(kr)]^2 dV$$

$$= \frac{\varepsilon_0 E_0^2}{2} \int_z \int_r J_0^2(kr) 2\pi r dr dz$$

$$= \frac{\varepsilon_0 E_0^2 L \pi}{2} \int_r J_0^2(kr) 2r dr$$

$$= \frac{\varepsilon_0 E_0^2 L \pi}{2} \left[r^2 [J_1^2(kr) - J_0^2(kr)] \right]_0^b$$

$$= \frac{\varepsilon_0 E_0^2}{2} \pi b^2 L J_1^2(p_{01}),$$

Dissipated Power

$$P = \frac{R_s}{2} \int_A H_\phi^2 dA$$

$$= \frac{R_s E_0^2}{2Z_0^2} \int_A J_1^2(kr) dA$$

$$= \frac{R_s E_0^2}{2Z_0^2} \left[\int J_1^2(kr) 2\pi r dr + 2\pi b \int J_1^2(p_{01}) dz \right]$$

$$= \frac{R_s E_0^2}{2Z_0^2} \left(2\pi \left[r^2 [J_1^2(kr) - J_0(kr) J_2(kr)] \right]_0^b + 2\pi b L J_1^2(p_{01}) \right)$$

$$= \frac{R_s E_0^2 \pi b}{Z_0^2} (b + L) J_1^2(p_{01})$$